

Key



BIRZEIT UNIVERSITY

Faculty of Information Technology

Computer Systems Engineering

Digital Signal Processing – ENCS431

Second Exam

Fall Semester 30/1/2014

Key

Time allowed: 90 minutes

Name:

ID:

section: 9:30-11:00, 12:30 -2:00

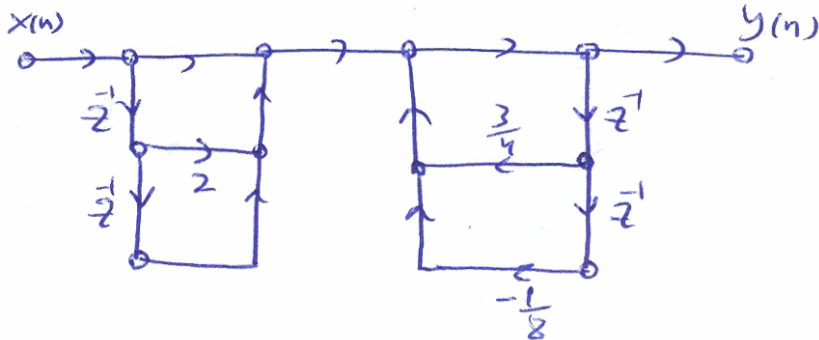
Question 1: [35 marks] Structures for DT systems

(a) Consider a causal LTI system whose system function is

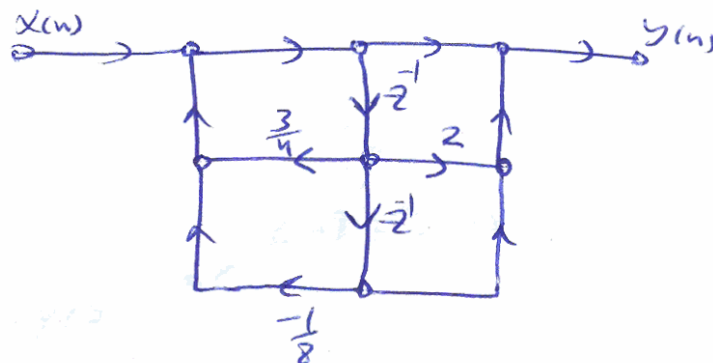
$$H(z) = \frac{1 + 2z^{-1} + z^{-2}}{1 - \frac{3}{4}z^{-1} + \frac{1}{8}z^{-2}}$$

Draw the signal flow graph for implementations of the system in each of the following forms: [5pts each]

(i) Direct form I.



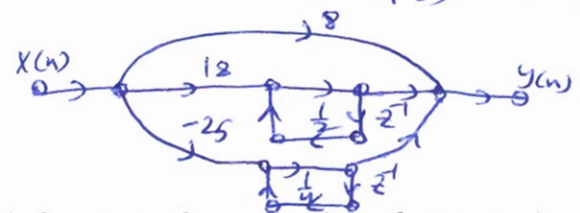
(ii) Direct form II (Canonical form)



(iii) Parallel form using first order sections by long division $H(z)$ can expressed as

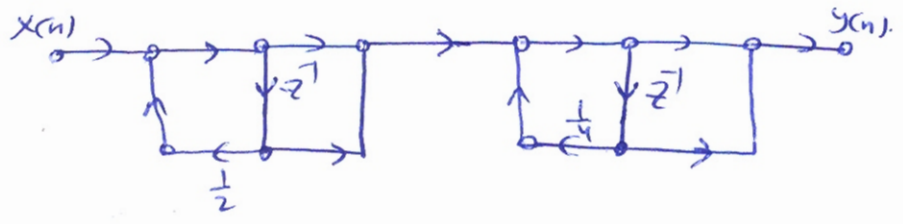
$H(z) = 8 + \frac{8z^{-1} - 7}{1 - \frac{3}{4}z^{-1} + \frac{1}{8}z^{-2}}$ by using partial fraction expansion

$$\frac{8z^{-1} - 7}{1 - \frac{3}{4}z^{-1} + \frac{1}{8}z^{-2}} = \frac{A_1}{(1 - \frac{1}{2}z^{-1})} + \frac{A_2}{(1 - \frac{1}{4}z^{-1})} = \frac{18}{1 - \frac{1}{2}z^{-1}} - \frac{25}{1 - \frac{1}{4}z^{-1}}$$

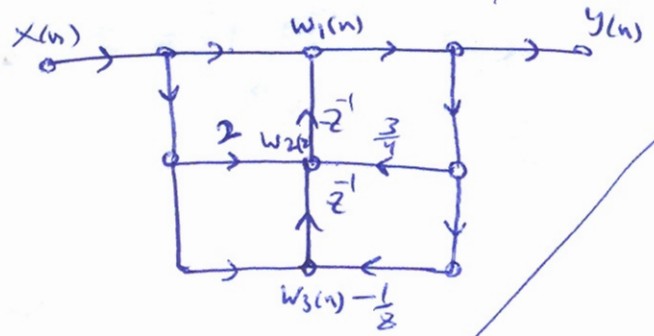


(iv) Cascade form using first-order direct form II sections

$$H(z) = \frac{(1 + z^{-1})^2}{(1 - \frac{1}{2}z^{-1})(1 - \frac{1}{4}z^{-1})} = \left(\frac{1 + z^{-1}}{1 - \frac{1}{2}z^{-1}} \right) \left(\frac{1 + z^{-1}}{1 - \frac{1}{4}z^{-1}} \right) = H_1 H_2$$



(v) Transposed direct form II



From this graph we write diff eqn

$$\begin{aligned} w_1 &= X(n) + w_2(n-1) \quad \text{--- (1)} \\ w_2 &= 2X(n) + w_3(n-1) + \frac{3}{4}Y(n) \quad \text{--- (2)} \\ w_3 &= X(n) - \frac{1}{8}Y(n) \quad \text{--- (3)} \\ Y(n) &= w_1 \quad \text{--- (4)} \end{aligned}$$

$$\begin{aligned} z^{-1}W_2 &= Y(z) - X(z) \\ W_2 &= 2X(z) + z^{-1}(X(z) - \frac{1}{8}Y(z)) + \frac{3}{4}Y(z) \\ \cancel{2X(z)} - \cancel{2X(z)} &= X(z) + \end{aligned}$$

(vi) Write the difference equation for the flow graph of (v) above and show that this system has the correct system function? [10 pts]

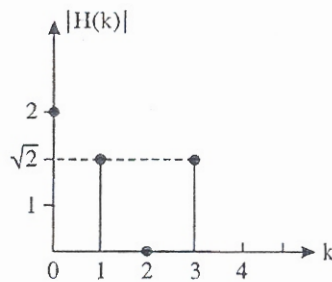
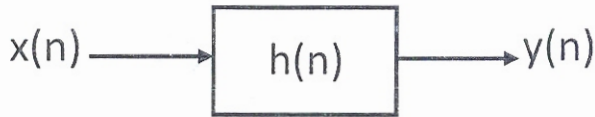
$$\begin{aligned} zY(z) - zX(z) &= 2X(z) + z^{-1}X(z) - \frac{1}{8}z^{-1}Y(z) + \frac{3}{4}Y(z) \\ \left(z + \frac{1}{8}z^{-1} - \frac{3}{4} \right) Y(z) &= (z + 2 + z^{-1})X(z) \\ \left(1 - \frac{3}{4}z^{-1} + \frac{1}{8}z^{-2} \right) Y(z) &= (1 + 2z^{-1} + z^{-2})X(z) \end{aligned}$$

$$\Rightarrow \frac{Y(z)}{X(z)} = \frac{1 + 2z^{-1} + z^{-2}}{1 - \frac{3}{4}z^{-1} + \frac{1}{8}z^{-2}} \text{ which is same of } H(z) \text{ given}$$

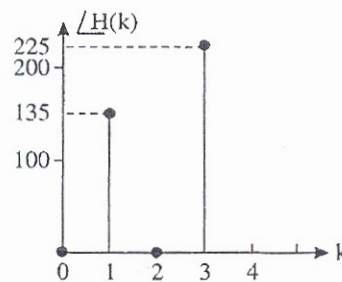
Question 2: [30 marks] DFT

(a) The figure below shows the block diagram of an LTI system, where $x(n)$ is the input sequence, $h(n)$ is the impulse response of the system and $y(n)$ is the output sequence. The magnitude and phase (in degrees) of the Discrete Fourier Transform $H(k)$ of $h(n)$ is shown in figures (a) and (b), respectively.

Assume the input sequence $x(n) = \{1, 0, 2, 3\}$.



(a)



(b)

(i) Determine the impulse response of this system, $h(n)$? [5pts]

2pts

$$H(0) = 2e^{j0} = 2$$

$$H(1) = \sqrt{2}e^{j135} = \sqrt{2}(\cos 135 + j\sin 135) = -1 + j$$

$$H(2) = 0$$

$$H(3) = \sqrt{2}e^{j225} = \sqrt{2}(\cos 225 + j\sin 225) = -1 - j$$

3pts

$$h(n) = \frac{1}{4} \sum_{k=0}^3 H(k) e^{j\frac{2\pi}{4}kn}$$

$$h(0) = \frac{1}{4} [2e^{j0} + (-1+j)e^{j\frac{\pi}{2}} + (-1-j)e^{j\frac{3\pi}{2}}] = \frac{1}{4} [2 - j - 1 + j - 1] = 0$$

$$h(1) = 1$$

$$h(2) = 1$$

$$h(3) = 1$$

$$h(n) = \{0, 0, 1, 1\}$$

$$h(n) = \frac{1}{4} \sum_{k=0}^3 H(k) W_4^{-nk}$$

$$h(0) = \frac{1}{4} \sum_{k=0}^3 H(k) e^{j\frac{2\pi}{4}k \cdot 0} = \frac{1}{4} (2 - 1 + j - 1 - j) = 0$$

(ii) Determine the output sequence $y(n)$ by computing the 4-point circular convolution? [5pts]

Using tabular method $y(n) = x(n) \circledast h(n)$

n:	0	1	2	3
$x(n)$	1	0	2	3
$h(n)$	0	0	1	1
	0	0	0	0
	0	0	0	0
	2	3	4	0
	0	2	3	1

$\Rightarrow y(n) = \{2, 5, 4, 1\}, 0 \leq n \leq 3$

(ii) Determine the output sequence, $y(n)$ using DFT? [10 pts]

5 pts $X(k) = \text{DFT of } x(n)$

$$\begin{bmatrix} X(0) \\ X(1) \\ X(2) \\ X(3) \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix} \begin{bmatrix} 1 \\ 0 \\ 2 \\ 3 \end{bmatrix} = \begin{bmatrix} 6 \\ -1+3j \\ -1 \\ -1-3j \end{bmatrix} \Rightarrow Y(k) = X(k) \cdot H(k) = \begin{bmatrix} 6 \\ -1+3j \\ 0 \\ -1-3j \end{bmatrix} \begin{bmatrix} 2 \\ -1+j \\ 0 \\ -1-j \end{bmatrix} = \begin{bmatrix} 12 \\ -2-4j \\ 0 \\ -2+4j \end{bmatrix}$$

5 pts $y(n) = \text{IDFT of } X(k)$

$$\begin{bmatrix} y(0) \\ y(1) \\ y(2) \\ y(3) \end{bmatrix} = \frac{1}{4} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & -1 & j \end{bmatrix} \begin{bmatrix} 12 \\ -2-4j \\ 0 \\ -2+4j \end{bmatrix} = \begin{bmatrix} 2 \\ 5 \\ 4 \\ 1 \end{bmatrix}$$

(b) Let 10-length finite sequence $x(n) = \{1, 2, 3, a, 1, 3, 2, 1, 2, 3\}$ for $0 \leq n \leq 9$. Assume $x(n)$ is zero outside the interval $0 \leq n \leq 9$. Let $X(e^{j\omega})$ be the DTFT of $x(n)$ and $X_1(k)$ be samples of $X(e^{j\omega})$ every $\frac{\pi}{3}$; i.e.,

$$X_1(k) = X(e^{j\omega}) \Big|_{\omega=(\pi/3)k}, \quad 0 \leq k \leq 5$$

The 6-point sequence $x_1(n)$ that results from taking the 6-point inverse DFT of $X_1(k)$ is

$x_1(n) = \{3, 3, 5, 7, 1, 3\}$ for $0 \leq n \leq 5$. Based on this sequence, x_1 , determine the value of the fourth sample of $x(n)$, i.e. $x(3)$ or a : is value of a unique? If so justify your answer. If not, find another choice of a consistent with the given information? [10 pts]

$X_1(k)$ is 6-point DFT of $x(n)$.

Since length of $x(n)$ is greater than 6 \Rightarrow There will be time aliasing by 4 points. \Rightarrow i.e. $x_1(n) = x(n)_8$

So

$$\begin{aligned} x_1(0) &= x(0) + x(6) = 3 \\ x_1(1) &= x(1) + x(7) = 3 \\ x_1(2) &= x(2) + x(8) = 5 \\ x_1(3) &= x(3) + x(9) = 7 \end{aligned}$$

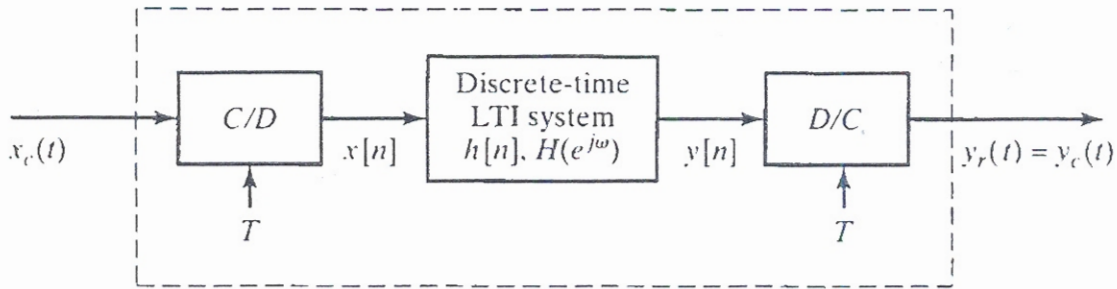
$$\Rightarrow a + 3 = 7 \Rightarrow a = 4$$

This value is unique.

It's clear from this relationship between $x(n)$ and $x_1(n)$ that a is unique.

Question 3: [35 marks] Sampling and reconstruction

Consider the system below:



- (a) Let the continuous-time input signal $x_c(t) = \cos(25\pi t) + \cos(120\pi t) + \cos(50\pi t)$, what is the range of sampling frequency, in Hz, that will avoid aliasing in the C/D converter? [5 pts]

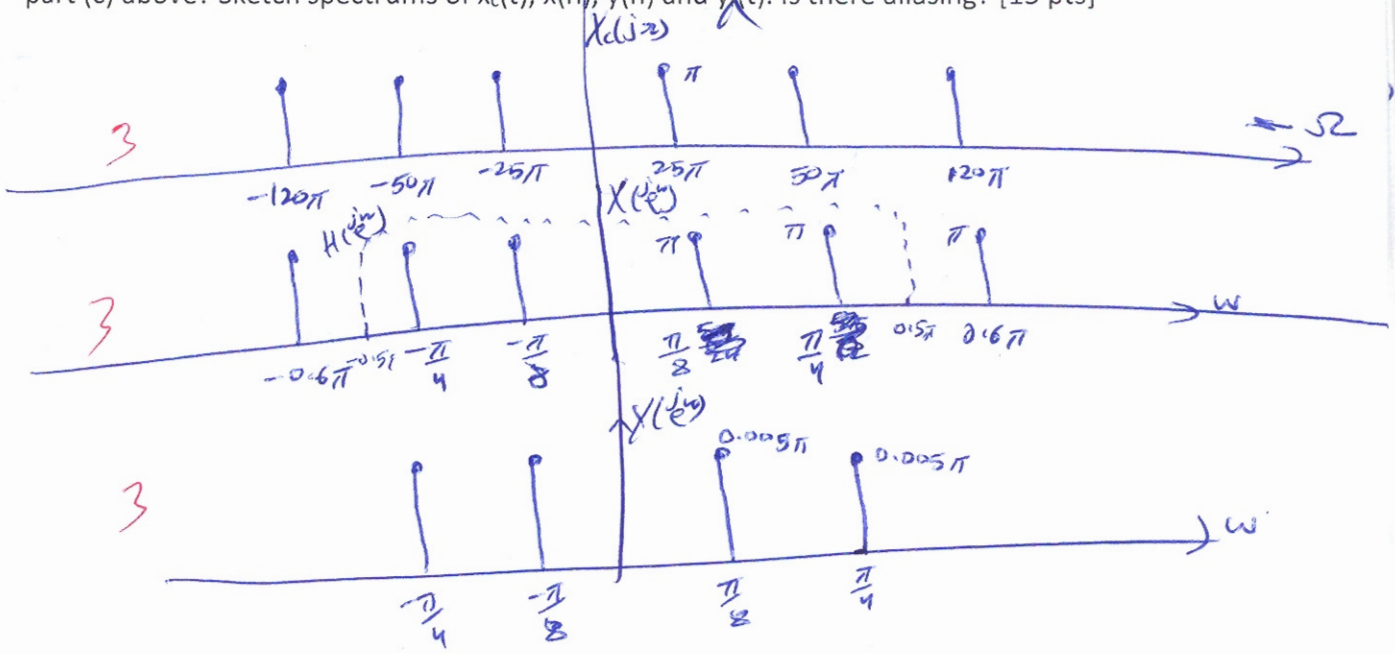
max. freq. of $x_c(t) \rightarrow \omega_{max} = 120\pi$
 $\omega_s \geq 2\omega_{max} \Rightarrow \omega_s \geq 240\pi$ } 2
 $2\pi f_s \geq 240\pi \Rightarrow f_s \geq 120 \text{ Hz}$ } 3

- (b) Now, assume the sampling frequency of C/D converter $f_{s1} = 200 \text{ Hz}$ and the discrete system is an ideal low-pass filter with frequency response as follow:

$$H(e^{j\omega}) = \begin{cases} 0.005 & |\omega| \leq 0.5\pi \\ 0 & \text{otherwise} \end{cases}$$

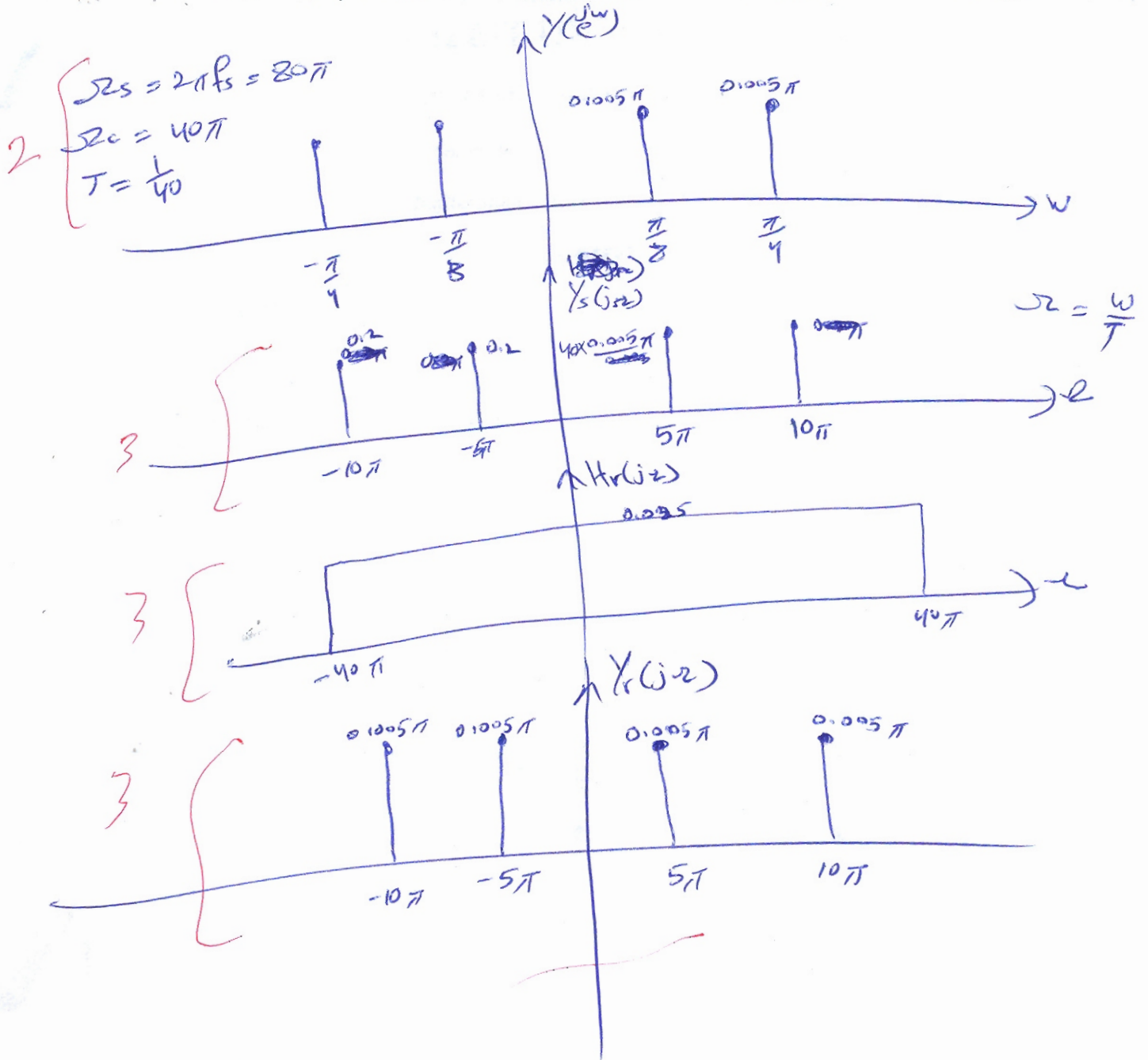
$\omega_s = 2\pi f_s = 2\pi(200) = 400\pi > 240\pi$
 2 pts } So, there is no aliasing
 2 pts }

For $-\pi \leq \omega \leq \pi$, Find the discrete output $y(n)$, if the input signal $x_c(t)$ is the same as specified in part (c) above? Sketch spectrums of $x_c(t)$, $x(n)$, $y(n)$ and $y_r(t)$. Is there aliasing? [15 pts]



2 [$y(n) = 0.005 \cos(\frac{\pi}{8}n) + 0.005 \cos(\frac{\pi}{4}n)$

(c) If the sampling frequency of D/C converter $F_{S2} = 40\text{Hz}$, Find the reconstructed continuous-time signal if the analogue reconstruction filter in the D/C converter, $H_r(j\Omega)$, has a gain of (0.025) and cut-off frequency, Ω_c , half of the sampling frequency. Sketch spectrums of all involved signals [15pts]



4

So, output $y_r(t) = 0.005 \left[\cos(5\pi t) + \cos(10\pi t) \right]$