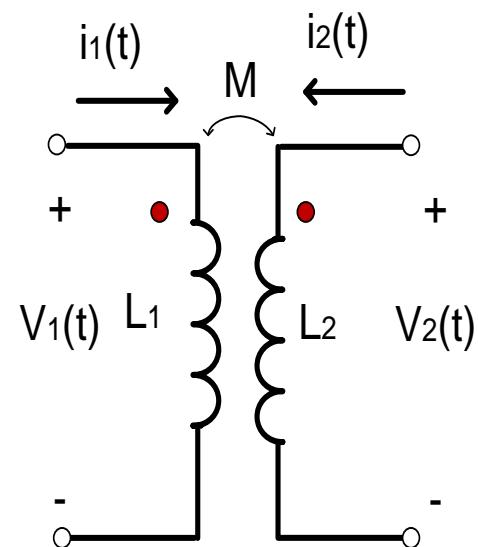


ENEE2301

Network Analysis 1



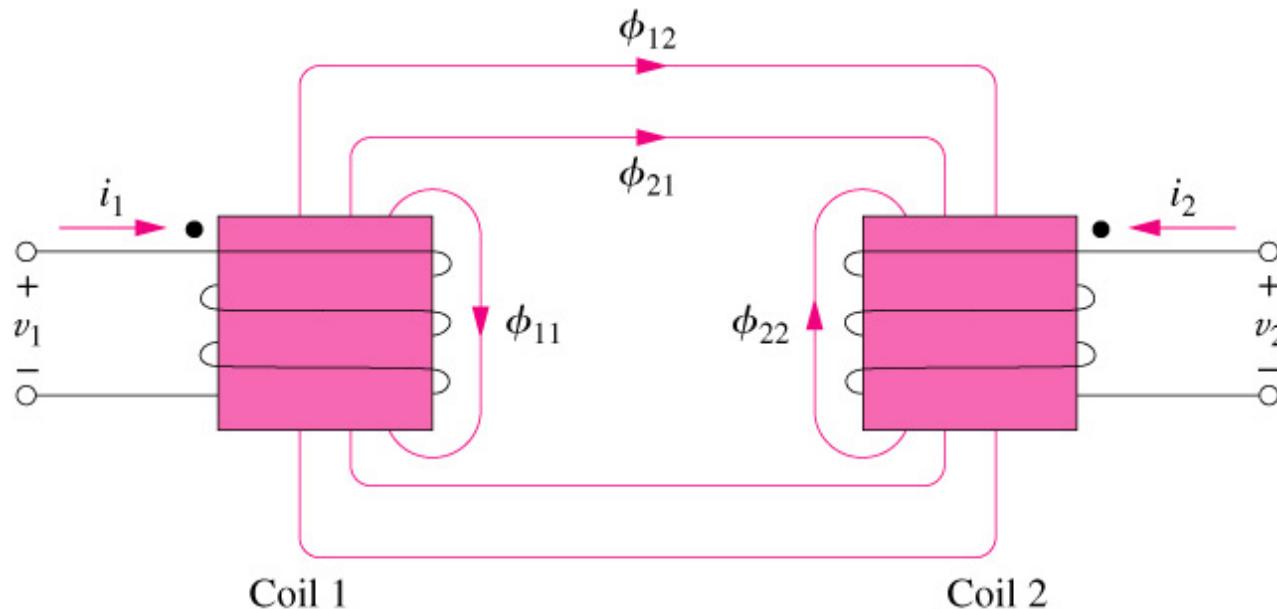
Transformers

Transformers

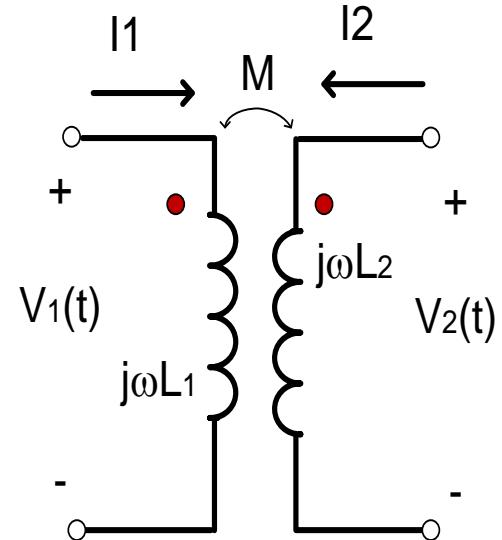
- Transformers are used in a wide variety of applications.
- In electric power transmission and distribution systems they step up the voltage at the sending end to reduce transmission losses
- and step down at the receiving end to make it safer and easier to utilize.
- Transformers change voltages and currents to any desired amplitude, large or small.
- They transform impedances and match load impedances to source impedances for maximum power transfer.

- **Transformers are constructed of two coils placed so that the Changing flux developed by one will link the other**
- **When two coils are placed close to each other, a changing flux in one coil will cause an induced voltage in the second coil.**
- **The coils are said to have mutual inductance M , which can either add or subtract from the total inductance depending on if the fields are aiding or opposing.**

- The coil to which the source is applied is called the primary coil
- The coil to which the load is applied is called the secondary coil



The Ideal Transformer



$$V_1 = j\omega L_1 I_1 + j\omega M I_2 \quad \text{----- (1)}$$

$$V_2 = j\omega M I_1 + j\omega L_2 I_2 \quad \text{----- (2)}$$

From (1)

$$I_1 = \frac{V_1 - j\omega M I_2}{j\omega L_1} \quad \text{----- (3)}$$

Substitute (3) in to (2)

$$V_2 = j\omega L_2 I_2 + \frac{M V_1}{L_1} - \frac{j\omega M^2 I_2}{L_1}$$

$$M = K\sqrt{L_1 L_2}$$

For ideal transformer k=1

$$M = \sqrt{L_1 L_2}$$

$$V_2 = j\omega L_2 I_2 + \frac{MV_1}{L_1} - \frac{j\omega M^2 I_2}{L_1}$$

$$V_2 = j\omega L_2 I_2 + \frac{\sqrt{L_1 L_2}}{L_1} V_1 - \frac{j\omega L_1 L_2 I_2}{L_1}$$

$$V_2 = \sqrt{\frac{L_2}{L_1}} V_1 = \frac{N_2}{N_1} V_1$$

Remember $L_1 = K N_1^2$

$$L_2 = K N_2^2$$

$$V_2 = n V_1$$

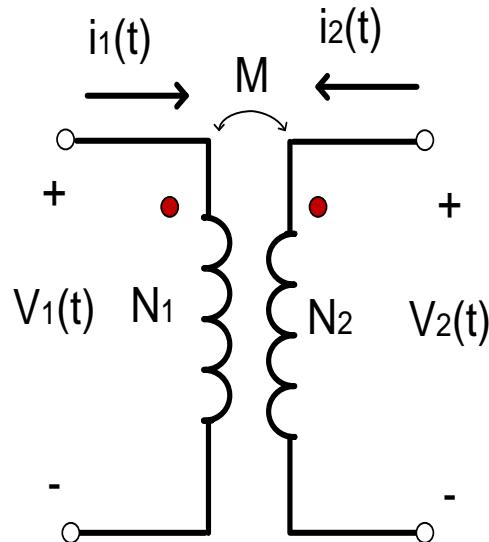
$$n = \frac{N_2}{N_1} \equiv \text{Turns ratio}$$

The ideal transformer

$$V_1(t) = N_1 \frac{d\Phi}{dt}$$

$$V_2(t) = N_2 \frac{d\Phi}{dt}$$

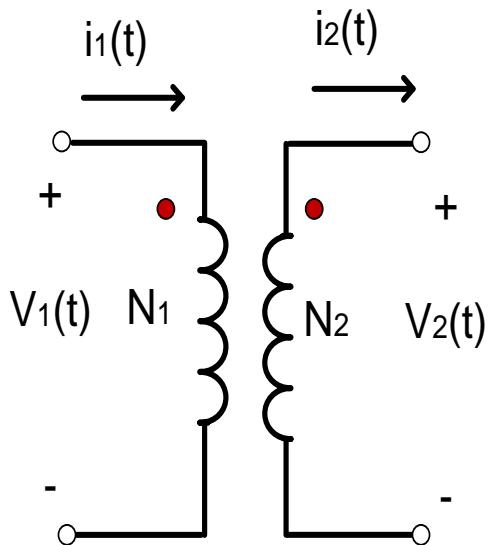
$$\frac{V_1(t)}{V_2(t)} = \frac{N_1}{N_2}$$



In an ideal transformer all flux generated by coil 1 is seen by coil 2 and there is no leakage flux

There is some techniques in transformer winding and in core geometry that allows the real transformers to approach the ideal characteristics

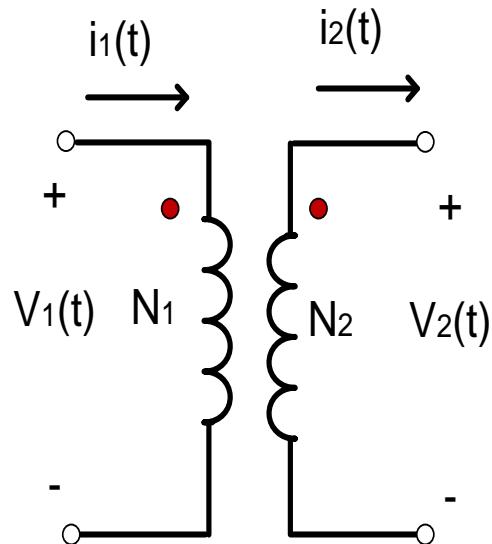
The ideal transformer



$$V_1(t) = N_1 \frac{d\Phi}{dt}$$

$$V_2(t) = N_2 \frac{d\Phi}{dt}$$

$$\frac{V_1(t)}{V_2(t)} = \frac{N_1}{N_2}$$



$$V_1(t) = N_1 \frac{d\Phi}{dt}$$

$$V_2(t) = N_2 \frac{d\Phi}{dt}$$

$$\frac{V_1(t)}{V_2(t)} = \frac{N_1}{N_2}$$

Assuming $P_1=P_2$ (100% efficiency- no losses)

$$V_1(t)i_1(t) = V_2(t)i_2(t)$$

$$\therefore \frac{i_1(t)}{i_2(t)} = \frac{V_2(t)}{V_1(t)} = \frac{N_2}{N_1}$$

Ideal Transformer

$$\frac{i_1(t)}{i_2(t)} = \frac{N_2}{N_1} \quad \xrightarrow{\hspace{1cm}} \quad N_1 : N_2 \xrightarrow{\hspace{1cm}}$$

Since the equations are algebraic ,
they are unchanged for phasors

$$\frac{I_1}{I_2} = \frac{V_2}{V_1} = \frac{N_2}{N_1}$$

Impedance Reflection

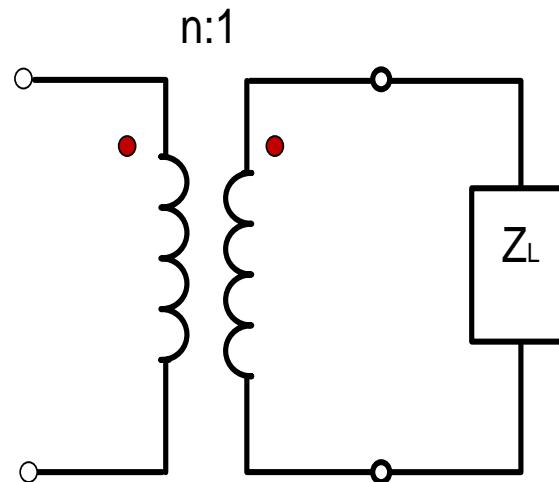
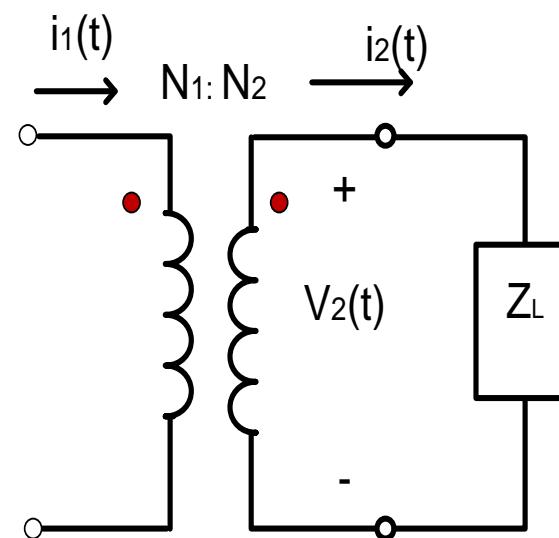
$$Z_{TH} = \frac{V_1}{I_1} = \frac{\frac{N_1}{N_2} V_2}{\frac{N_2}{N_1} I_2}$$

$$Z_{TH} = \frac{N_1^2}{N_2^2} \frac{V_2}{I_2}$$

$$Z_{TH} = \frac{N_1^2}{N_2^2} Z_L$$

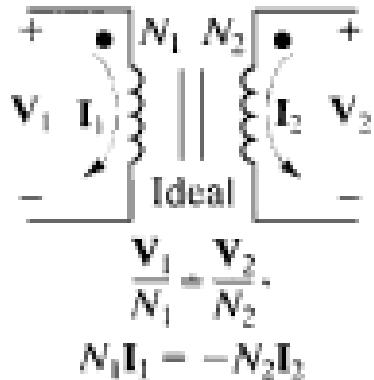
Let $\frac{N_1}{N_2} = a$

$$Z_{TH} = a^2 Z_L$$

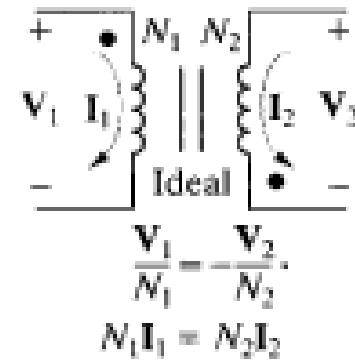


Important

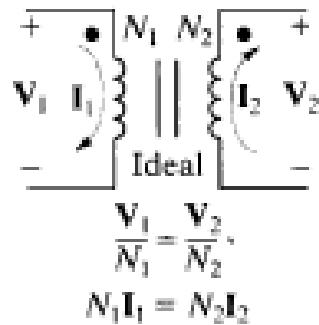
Dot Convention for Ideal Transformer



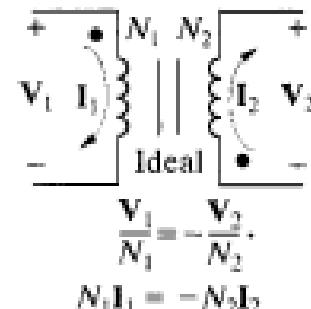
(a)



(b)

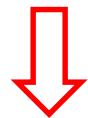
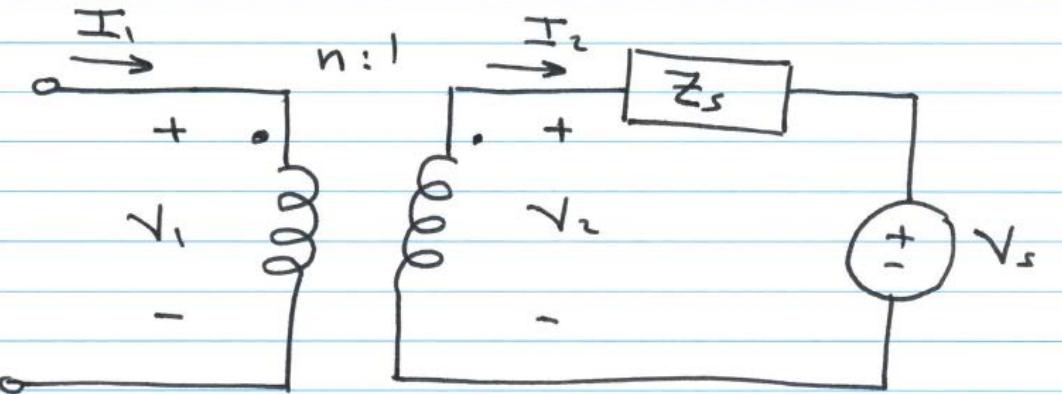


(c)

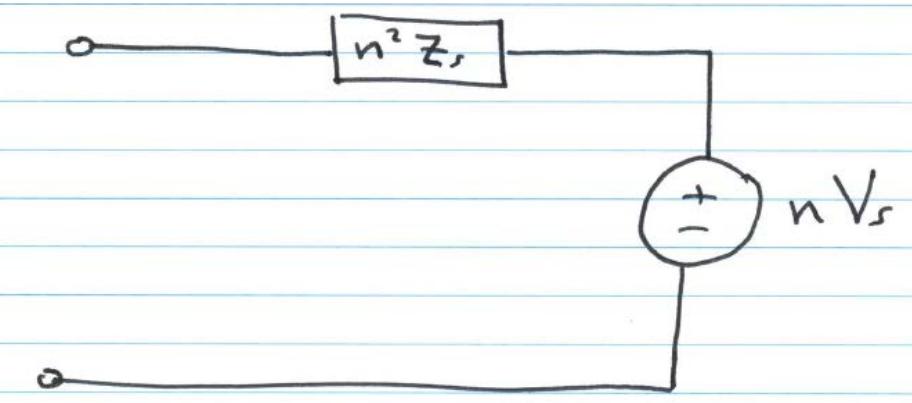


(d)

Example

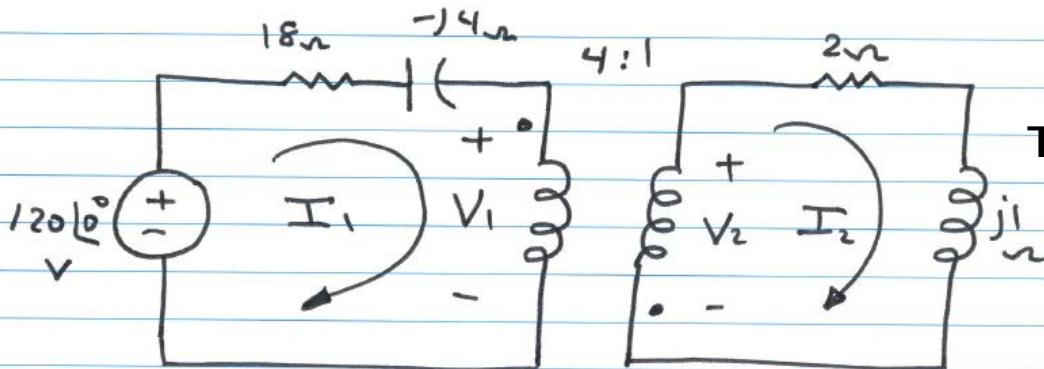


$$V_{TH} = V_1 = nV_2 = nV_s$$

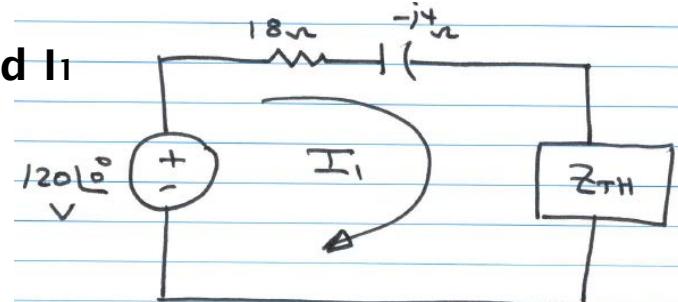


$$Z_{TH} = n^2 Z_s$$

Find I_1 and I_2



To Find I_1

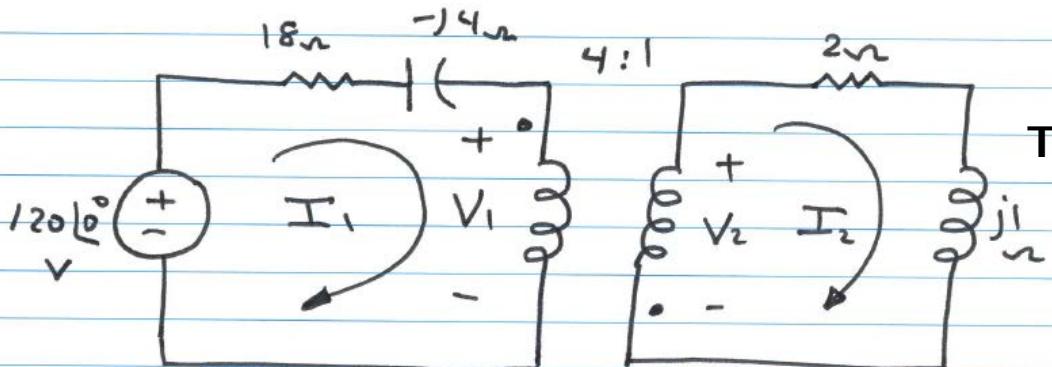


$$Z_{TH} = \frac{N_1^2}{N_2^2} Z_L = 4^2 (2 + j1) = 32 + j16$$

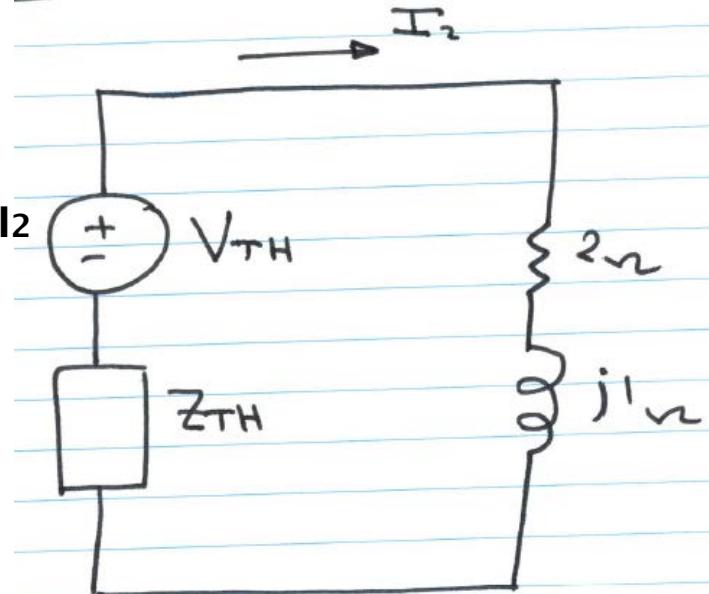
$$I_1 = \frac{120\langle 0 }{18 - j4 + 32 + j16} = \frac{120\langle 0 }{50 + j12}$$

$$= 2.33\langle -13.5^\circ \text{ A}$$

Find I_1 and I_2



To Find I_2



$$Z_{TH} = \frac{N_2^2}{N_1^2} Z_s = \frac{(18 - j4)}{4^2} = 1.125 - j0.25 \Omega$$

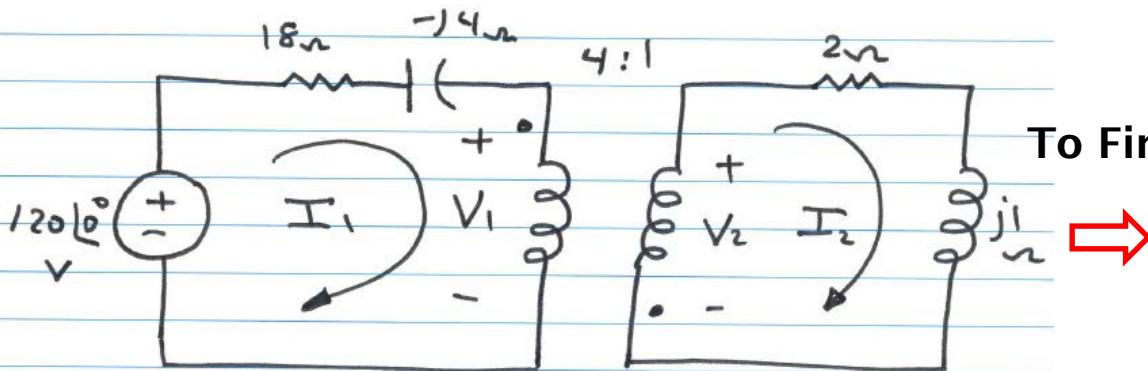
$$V_{TH} = -\frac{120 \angle 0}{4} = -30 \angle 0 \text{ V}$$

$$I_2 = \frac{V_{TH}}{Z_{TH} + 2 + j1}$$

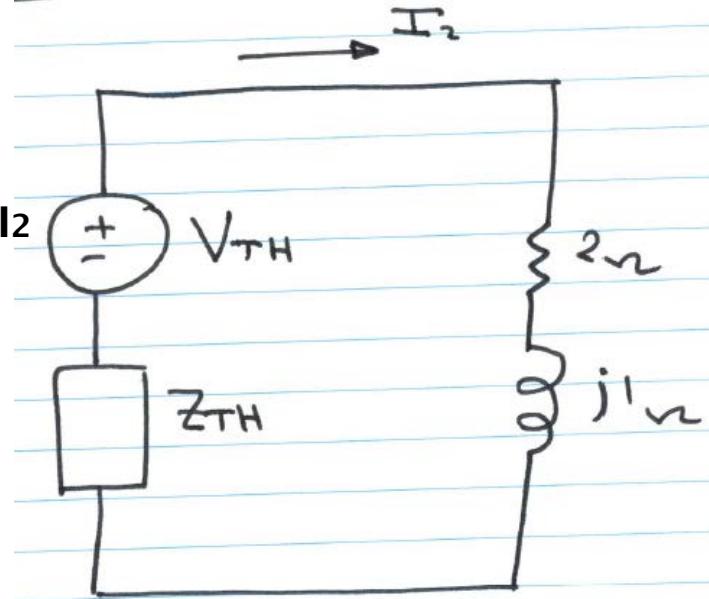
$$= 9.335 \angle 166.5^\circ \text{ A}$$

$$\begin{aligned} & \left. \begin{array}{c} + \\ V_1 \end{array} \right\} \frac{N_1}{N_2} \left| \begin{array}{c} N_1 \\ N_2 \end{array} \right\| \left| \begin{array}{c} I_2 \\ V_2 \end{array} \right\} \left. \begin{array}{c} + \\ V_2 \end{array} \right\} \\ & \text{Ideal} \\ & \frac{V_1}{N_1} = -\frac{V_2}{N_2}, \\ & N_1 I_1 = -N_2 I_2 \end{aligned} \right. \quad (d)$$

Find V_1 and V_2



To Find I_2



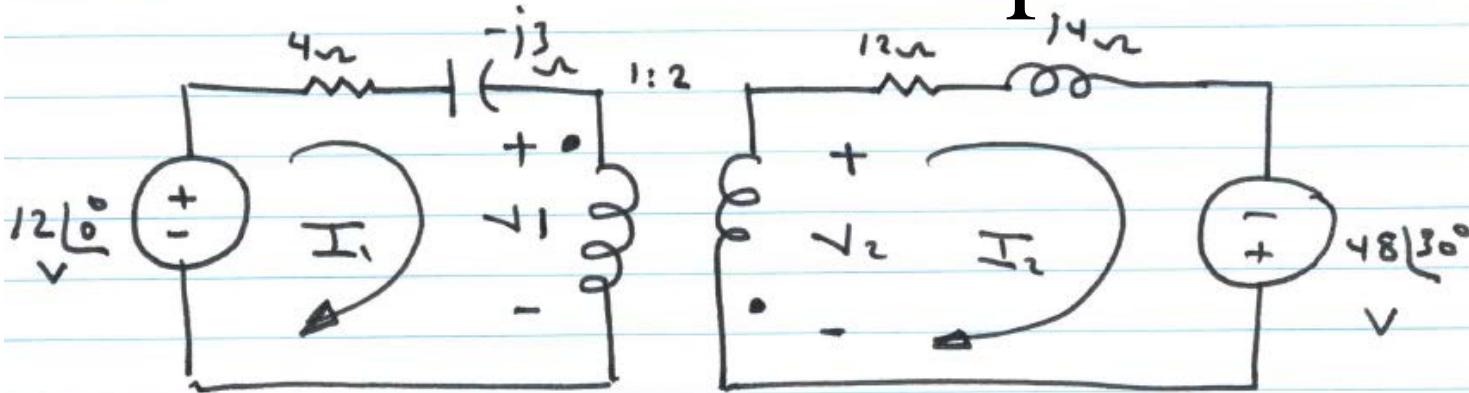
$$V_2 = I_2(2 + j1)$$

$$V_2 = 20.87 \angle 193.07^\circ \text{ V}$$

$$V_1 = -4V_2$$

$$= 83.49 \angle 3.07^\circ \text{ V}$$

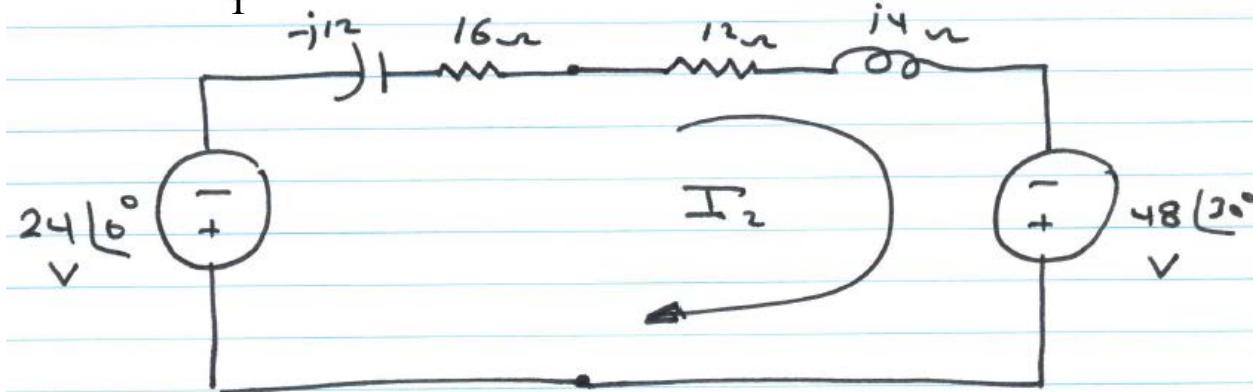
Example



↓ To Find I_2

$$V_{TH1} = -\frac{N_2}{N_1} V_1 = -2 (12 \angle 0^\circ) = -24 \angle 0^\circ$$

$$Z_{TH1} = \frac{N_2^2}{N_1^2} Z_s = 2^2 (4 - j3) = 16 - j12 \Omega$$



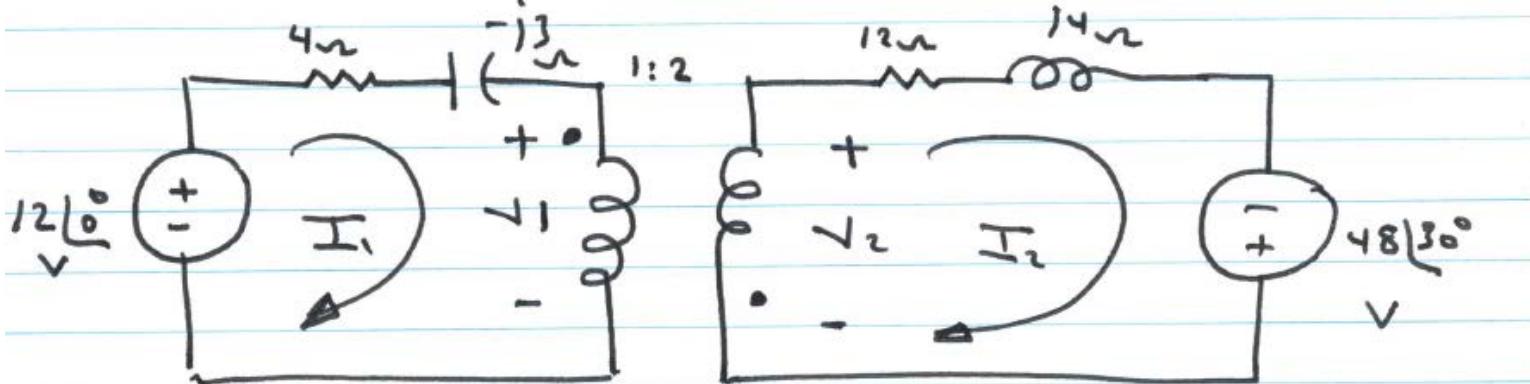
Ideal

$$\frac{V_1}{N_1} = -\frac{V_2}{N_2}$$

$$N_1 I_1 = -N_2 I_2$$

(d)

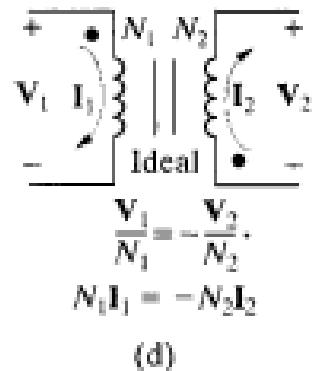
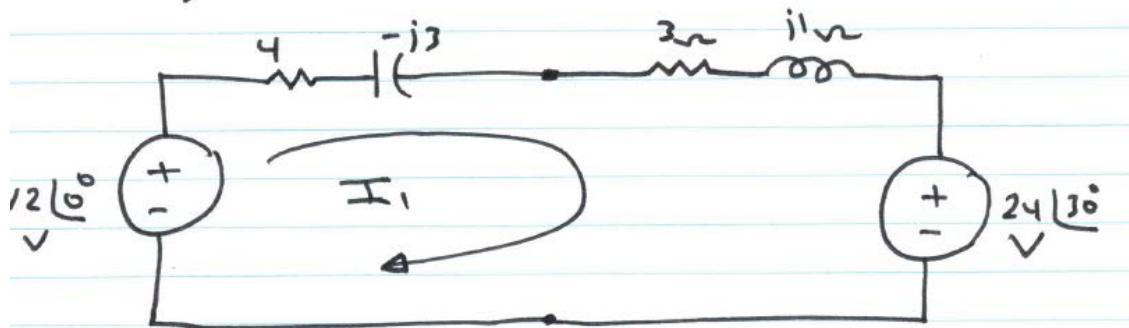
Example



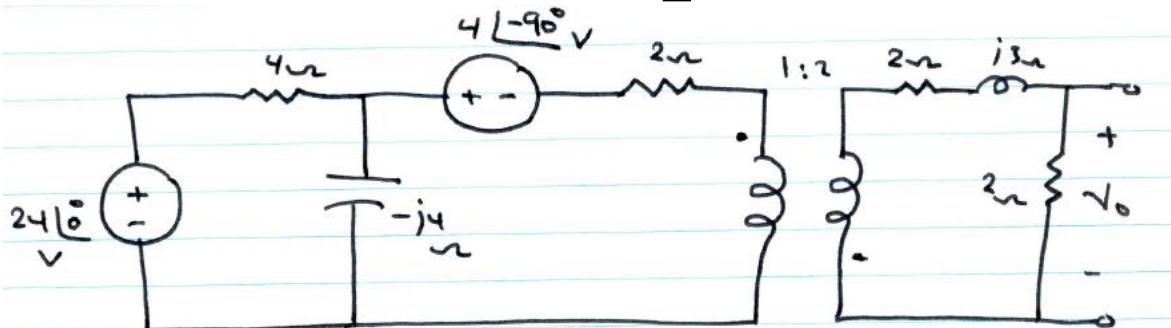
↓ To Find I_1

$$V_{TH2} = -\frac{N_1}{N_2} V_2 = -\frac{1}{2} (48 \angle 30^\circ) = -24 \angle 30^\circ$$

$$Z_{TH2} = \frac{N_1^2}{N_2^2} Z_L = \frac{(12 + j4)}{2^2} = 3 + j1 \Omega$$



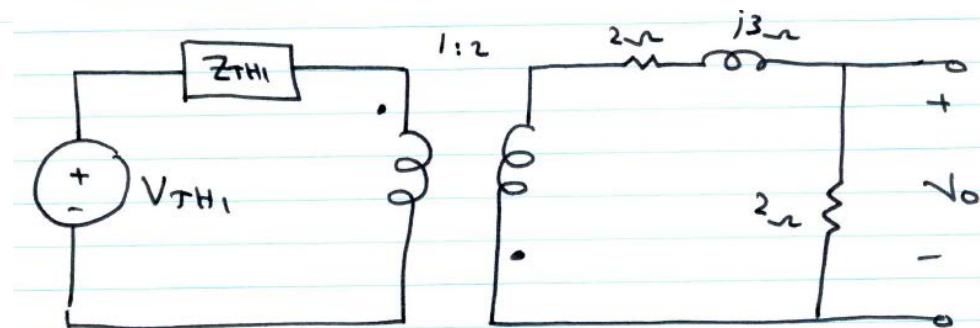
Example: Find V_o



$$Z_{TH1} = 2 + (4 // -j4)$$

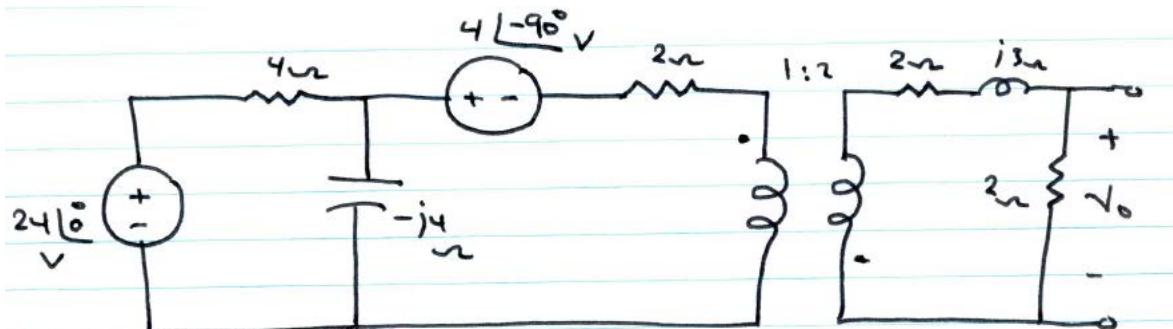
$$= 2 + \left(\frac{-j16}{4 - j4} \right) \left(\frac{4 + j4}{4 + j4} \right) = 2 + \left(\frac{-j64 + 64}{32} \right)$$

$$= 2 + (2 - j2) = 4 - j2 \Omega$$



Important

Example: Find V_o

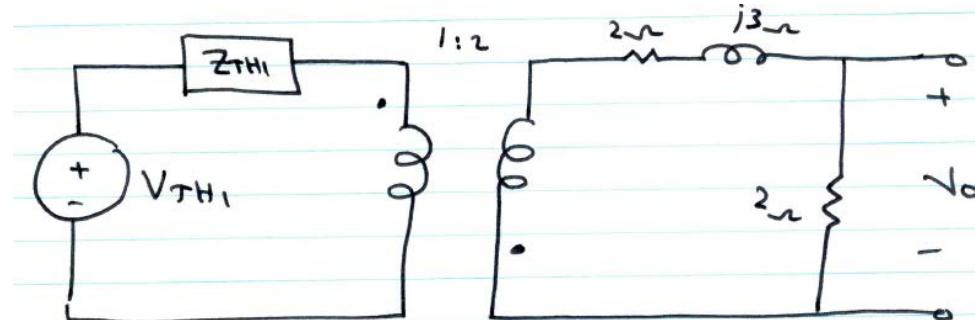


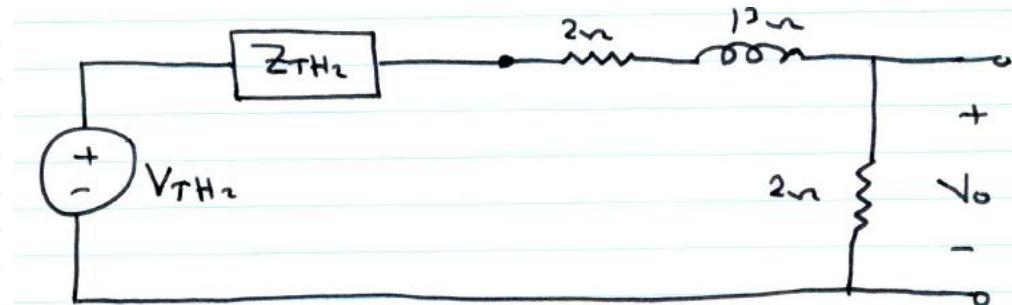
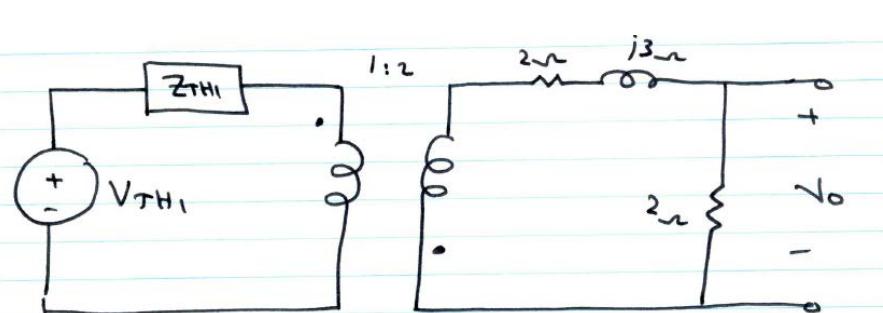
$$V_{TH} = \frac{-j4}{4 - j4} 24 \langle 0^\circ - 4 \langle -90^\circ$$

$$\frac{-j4}{4 - j4} = \frac{-j4}{4 - j4} \times \frac{4 + j4}{4 + j4} = \frac{-j16 + 16}{32} = \frac{1}{2} - j \frac{1}{2}$$

$$V_{TH} = \left(\frac{1}{2} - j \frac{1}{2} \right) 24 \langle 0^\circ - (-j4)$$
$$= 12 - j12 + j4 = 12 - j8$$

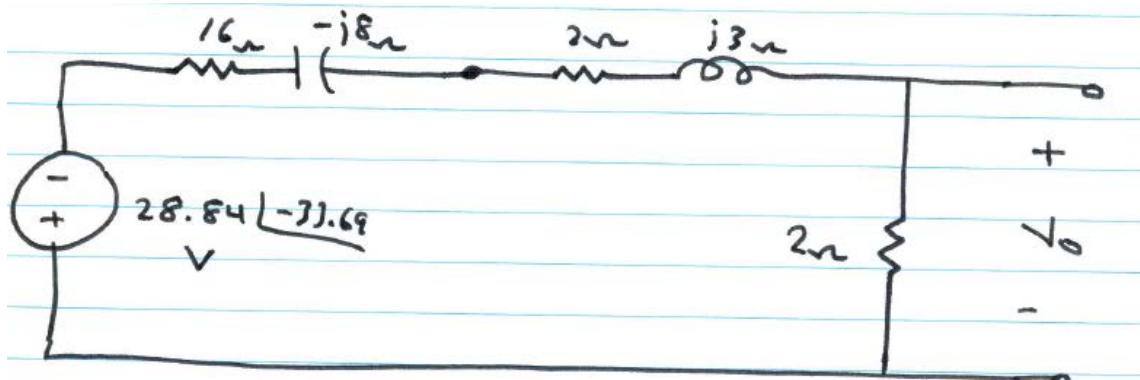
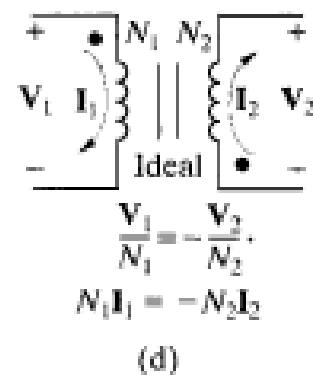
$$V_{TH1} = 14.42 \langle -33.69^\circ \text{ V}$$





$$Z_{TH2} = (2^2) Z_{TH1} = 4(4 - j2) = 16 - j8 \Omega$$

$$V_{TH2} = -2V_{TH1} = -28.84 \angle -33.69^\circ \text{ V}$$



$$V_o = \frac{2}{(2 + 2 + 16 + j3 - j8)} (-28.84 \angle -33.69^\circ)$$

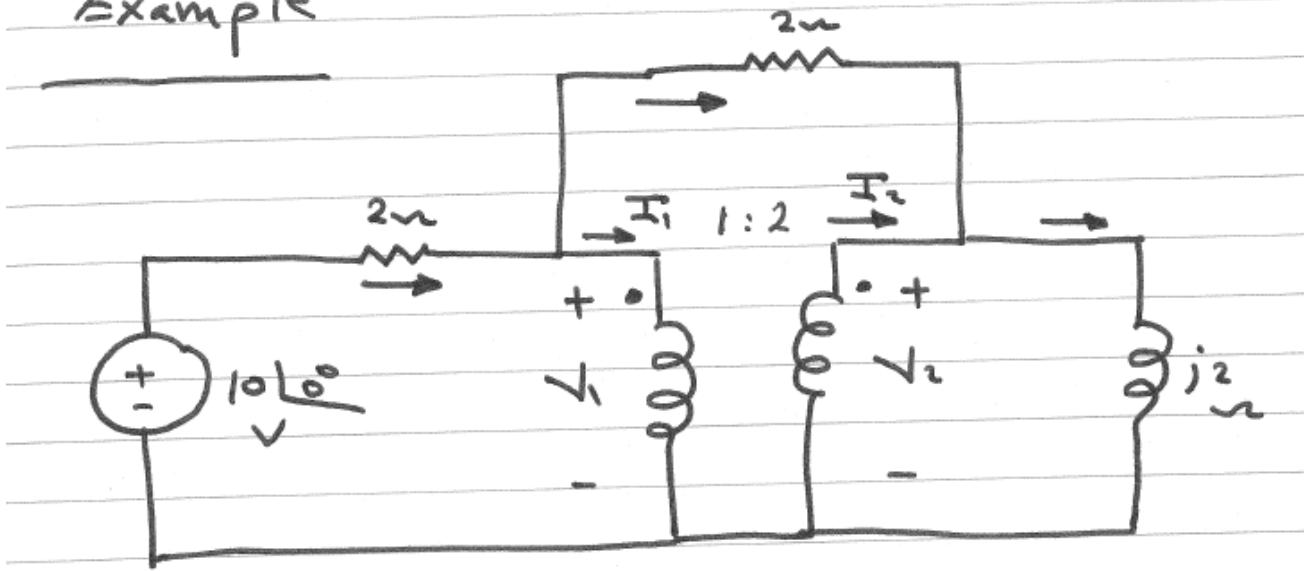
$$V_o = 2.8 \angle 160.35^\circ \text{ V}$$

$$\frac{2}{(2+2+16+j3-j8)} =$$

$$= \frac{2}{(20-j5)} \frac{(20+j5)}{(20+j5)} = \frac{40+j10}{425} = 0.09701 \langle 14.036^\circ$$

$$V_o = (0.09701 \langle 14.036) (-28.84 \langle -33.69^\circ) = 2.8 \langle 160.35^\circ$$

Example



KCL :

$$\frac{10 - V_1}{2} = \frac{V_1 - V_2}{2} + H_1$$

KCL :

$$I_2 + \frac{V_1 - V_2}{2} = \frac{V_2}{j2}$$

$$V_2 = 2V_1$$

$$I_1 = 2I_2$$

$$\therefore I_1 = 5 \angle 0^\circ A$$

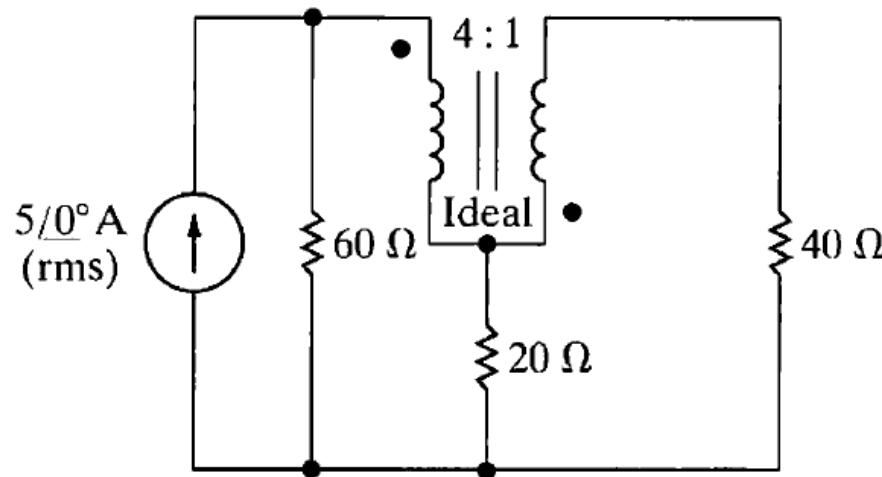
$$I_2 = 2.5 \angle 0^\circ A$$

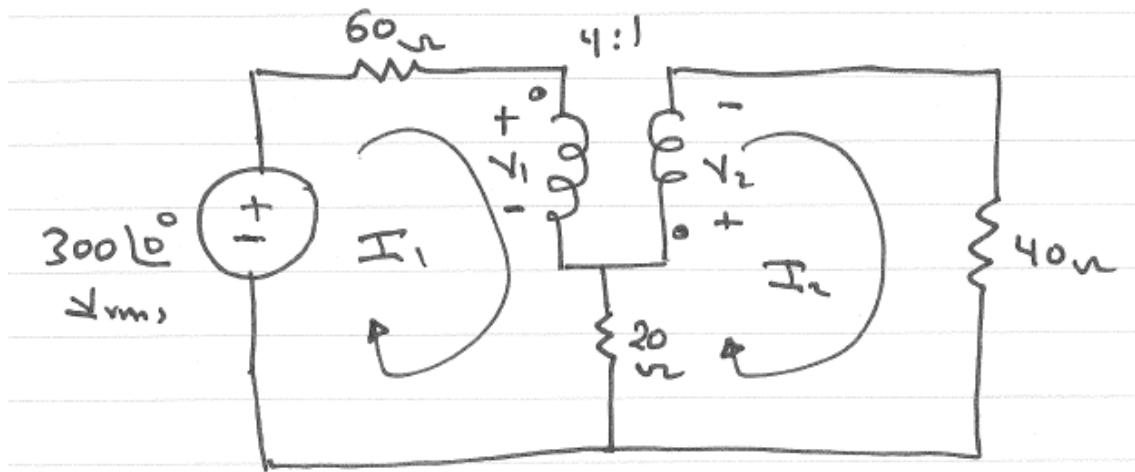
$$V_1 = \sqrt{5} \angle 63^\circ V$$

$$V_2 = 2\sqrt{5} \angle 63^\circ V$$

- 10.39** a) Find the average power delivered by the sinusoidal current source in the circuit of Fig. P10.39.
b) Find the average power delivered to the $20\ \Omega$ resistor.

Figure P10.39





$$300 \angle 0^\circ = 60 \vec{H}_1 + \vec{V}_1 + 20 (\vec{H}_1 - \vec{H}_2)$$

$$0 = 40 \vec{H}_2 + 20 (\vec{H}_2 - \vec{H}_1) + \vec{V}_2$$

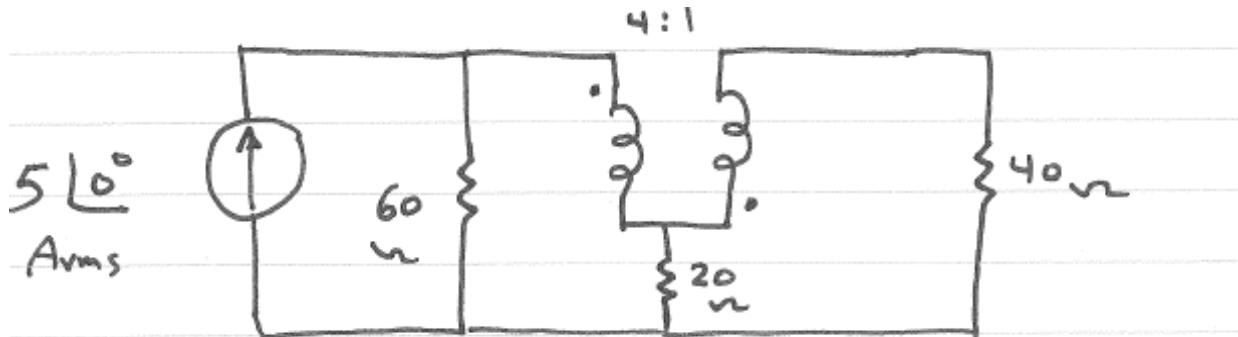
$$\vec{V}_2 = \frac{\vec{V}_1}{4} ; \quad \vec{H}_2 = -\frac{1}{4} \vec{H}_1$$

Solving:

$$\vec{V}_1 = 260 \angle 0^\circ \text{ V rms}$$

$$\vec{V}_2 = 65 \angle 0^\circ \text{ V rms}$$

$$\vec{H}_1 = 0.25 \angle 0^\circ \text{ A rms} ; \quad \vec{H}_2 = 1 \angle 180^\circ \text{ A rms}$$



$$\vec{V}_{5\angle 0^\circ A} = \vec{I}_1 + 20(\vec{I}_1 - \vec{I}_2) = 285\angle 0^\circ \text{ V rms}$$

$$P_{5\angle 0^\circ A} = (5)(285) = 1425 \text{ W}$$

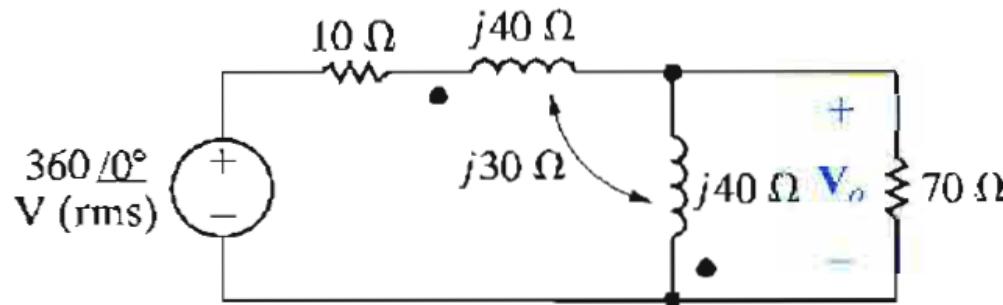
$$\vec{I}_{20\Omega} = \vec{I}_1 - \vec{I}_2 = 1.25\angle 0^\circ \text{ Arms}$$

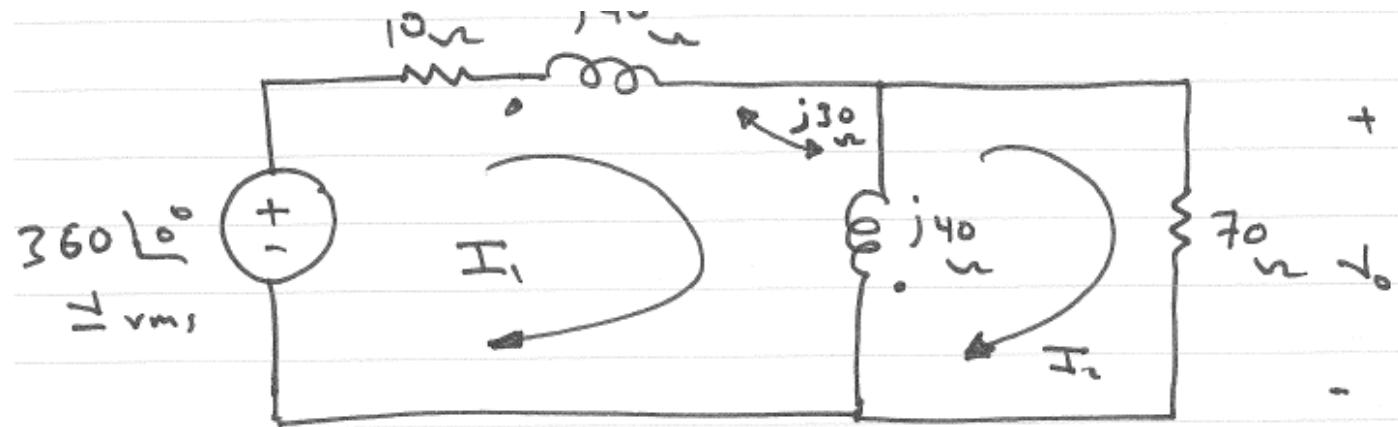
$$P_{20\Omega} = (1.25)^2 \cdot 20 = 31.25 \text{ W}$$

10.44 For the frequency-domain circuit in Fig. P10.44, calculate:

- the rms magnitude of \mathbf{V}_o .
- the average power dissipated in the 70Ω resistor.
- the percentage of the average power generated by the ideal voltage source that is delivered to the 70Ω load resistor.

Figure P10.44





$$360 \angle 0^\circ = 10 \vec{I}_1 + j40 \vec{I}_1 - j30 (\vec{I}_1 - \vec{I}_2) + j40 (\vec{I}_2 - \vec{I}_1) - j30 \vec{I}_2$$

$$360 \angle 0^\circ = (10 + j20) \vec{I}_1 - j10 \vec{I}_2$$

$$0 = 70 \vec{I}_2 + j40 (\vec{I}_2 - \vec{I}_1) + j30 \vec{I}_1$$

$$0 = -j10 \vec{I}_1 + (70 + j40) \vec{I}_2$$

solving :

$$\overrightarrow{I_1} = 16.1245 \angle -60.25^\circ \text{ A rms}$$

$$\overrightarrow{I_2} = 2 \angle 0^\circ \text{ A rms}$$

$$\overrightarrow{V_o} = 140 \angle 0^\circ \text{ V rms}$$

b) $P_{70} = (I_2)^2 \cdot 70 = 280 \text{ W}$

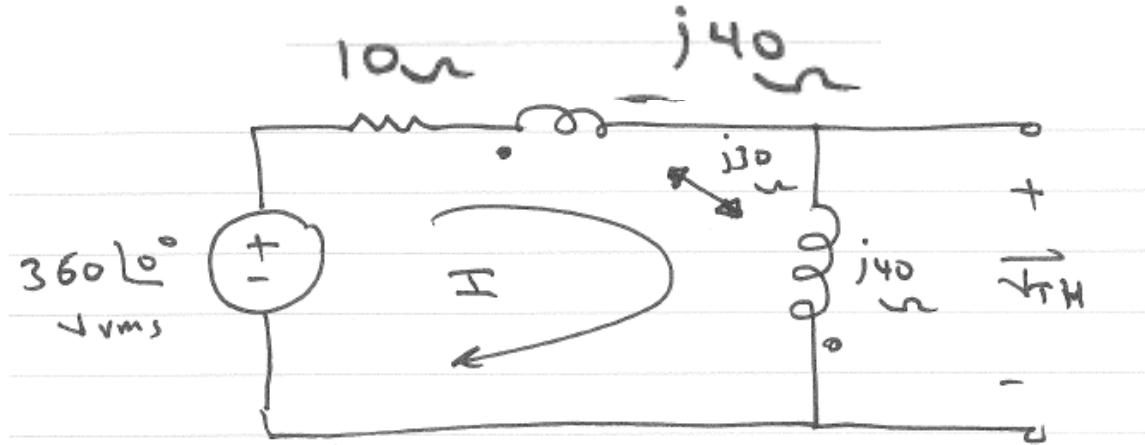
c) $P_{360 \angle 0^\circ} = \sqrt{I} \cos G = (360)(16.1245) \cos(0 + 60.25^\circ)$

$$P_{360 \angle 0^\circ} = 2880 \text{ W}$$

$$\therefore \text{delivered} = \frac{280}{2880} \times 100\% = 9.72\%$$

10.45 The 70Ω resistor in the circuit in Fig. P10.44 is replaced with a variable impedance Z_o . Assume Z_o is adjusted for maximum average power transfer to Z_o .

- a) What is the maximum average power that can be delivered to Z_o ?
- b) What is the average power developed by the ideal voltage source when maximum average power is delivered to Z_o ?



$$360\angle 0^\circ = 10\vec{I} + j40\vec{I} - j30\vec{I} + j40\vec{I} - j30\vec{I}$$

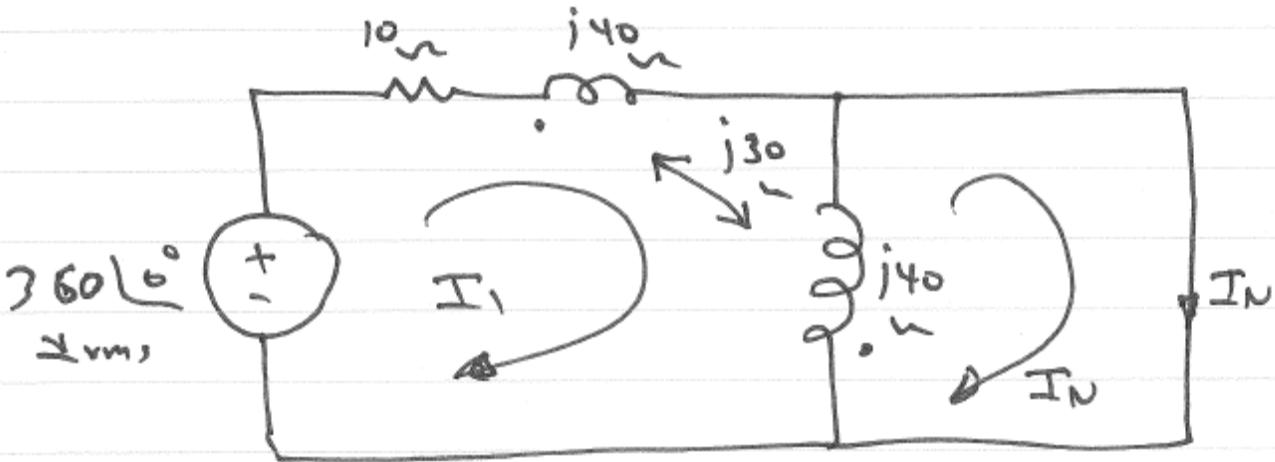
$$360\angle 0^\circ = (10 + j20)\vec{I}$$

$$\vec{I} = 7.2 - j14.4 \text{ A rms},$$

$$\vec{V}_{TH} = j40\vec{I} - j30\vec{I} = j10\vec{I}$$

$$\vec{V}_{TH} = 144 + j72 = 161\angle 26.56^\circ \text{ V rms},$$

To find I_N



$$360 \angle 0^\circ = (10 + j20) \vec{I}_1 - j10 \vec{I}_2$$

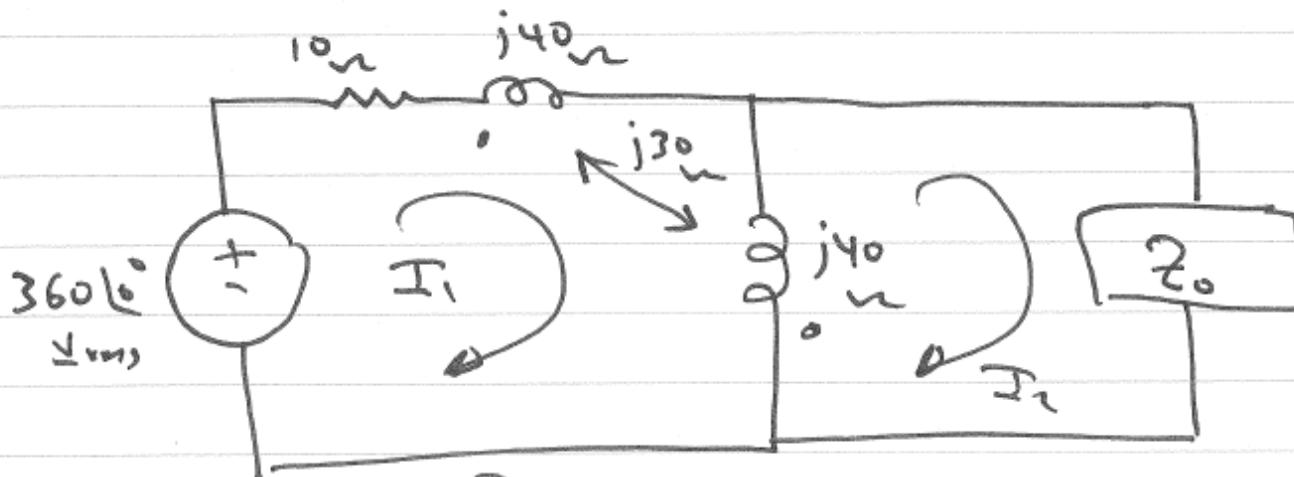
$$0 = -j10 \vec{I}_1 + j40 \vec{I}_N$$

$$\therefore \vec{I}_N = 2.215 - j2.877 \text{ A}_{\text{rms}}$$

$$Z_{TH} = \frac{\sqrt{r_H}}{\frac{1}{I_2}} = 2 + j36 \Omega$$

$$\therefore Z_0 = Z_{TH}^* = 2 - j36 \Omega$$

$$P_{L, \max} = \frac{(\sqrt{r_H})^2}{4 R_{TH}} = 3240 \text{ W}$$



$$I_2 = \frac{\sqrt{r_H}}{4} = 36 + j18 \text{ Arms}$$

$$360\angle 0^\circ = (10 + j20) \overline{I}_1 - j10 \overline{I}_2$$

$$\therefore \overline{I}_1 = 18\angle 0^\circ \text{ Arms}$$