

1. (a)  $X(e^{j0})$

$$X(e^{j0}) = \sum_{n=-\infty}^{\infty} x[n] = 0$$

(b)  $\arg\{X(e^{j\Omega})\}$

$x[n]$  is a real and odd function shifted by 2 to the right, i.e.  
 $x[n] = x_o[n - 2]$ .

Since  $x_o[n]$  is real and odd,  $X_o(e^{j\Omega})$  is purely imaginary,  
thus

$$X(e^{j\Omega}) = |X_o(e^{j\Omega})|e^{-j2\Omega}e^{j\frac{\pi}{2}}$$

which means,

$$\arg\{X(e^{j\Omega})\} = \frac{\pi}{2} - 2\Omega$$

(c)

$$\int_{-\pi}^{\pi} |X(e^{j\Omega})|^2 d\Omega = 2\pi \sum_{n=-\infty}^{\infty} |x[n]|^2 = 28\pi$$

(d)

$$\int_{-\pi}^{\pi} X(e^{j\Omega}) e^{j3\Omega} d\Omega = 2\pi x(3) = -2\pi$$

(e)

$$y[n] \leftrightarrow \operatorname{Re}\{e^{j2\Omega} X(e^{j\Omega})\} \text{ by DTFT.}$$

2.

$$x_s(t) = x(t)p(t) = x(t) \sum_{n=-\infty}^{\infty} \delta(t - nT_s)$$

But since  $p(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT_s)$  is periodic, it can be represented as a Fourier Series:

$$p(t) = \sum_{n=-\infty}^{\infty} C_n e^{jn2\pi f_s t},$$

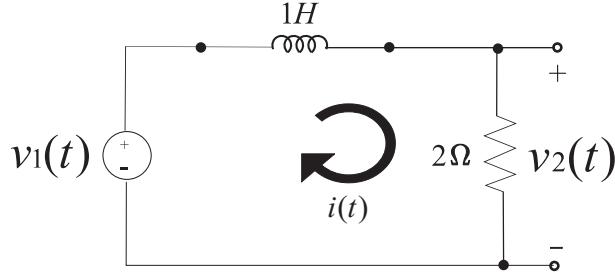
where

$$\begin{aligned} C_n &= \frac{1}{T_s} \int_{-T_s/2}^{T_s/2} p(t) e^{-jn2\pi f_s t} dt \\ &= \frac{1}{T_s} \int_{-T_s/2}^{T_s/2} \delta(t) e^{-jn2\pi f_s t} dt = \frac{1}{T_s} = f_s \end{aligned}$$

$$\therefore p(t) = \sum_{n=-\infty}^{\infty} f_s e^{jn2\pi f_s t} \text{ and } x_s(t) = x(t) \sum_{n=-\infty}^{\infty} f_s e^{jn2\pi f_s t}$$

Then

$$\begin{aligned} X_s(f) &= \int_{-\infty}^{\infty} x_s(t) e^{-j2\pi f t} dt = \int_{-\infty}^{\infty} f_s \cdot x(t) \sum_{n=-\infty}^{\infty} f_s e^{jn2\pi f_s t} e^{-j2\pi f t} dt \\ &= \sum_{n=-\infty}^{\infty} f_s \int_{-\infty}^{\infty} x(t) e^{-j2\pi(f-nf_s)t} dt \\ &= \sum_{n=-\infty}^{\infty} f_s X(f - nf_s) \end{aligned}$$



3. (a)

$$L \frac{di}{dt} + Ri = v_1 \Rightarrow 1 \cdot \frac{di}{dt} + 2 \cdot i = v_1$$

$$sI(s) + 2I(s) = V_1(s) \quad (\text{assuming } i(0) = 0)$$

$$I(s) = \frac{1}{s+2} V_1(s)$$

$$V_2(s) = 2I(s) = \frac{2}{s+2} V_1(s)$$

$$\begin{aligned} H_a(s) &= \frac{V_2(s)}{V_1(s)}|_{i(0)=0} \\ &= \frac{2}{s+2} = \frac{1}{1+\frac{s}{2}} \end{aligned}$$

(b)

$$H_a(j\omega) = \frac{1}{1 + \frac{j\omega}{2}}$$

$\therefore \omega_{3dB} = 2$  is the 3-dB dropping frequency.

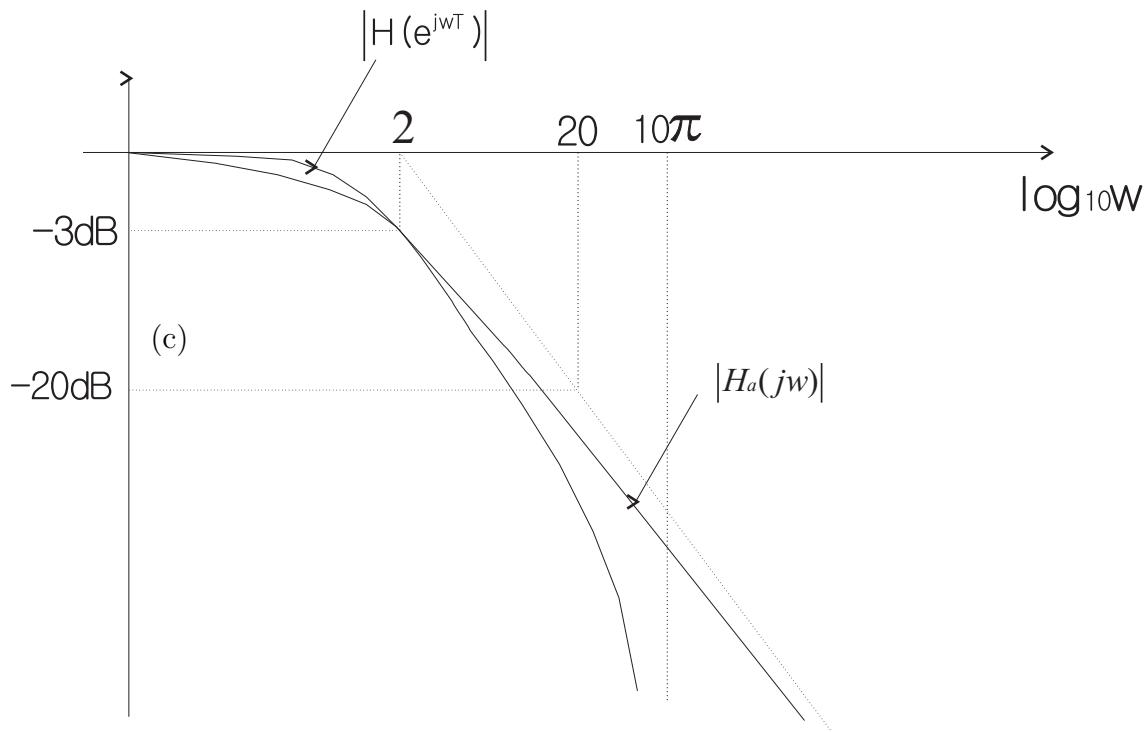
On the other hand, for bilinear z-transform,  
we set

$$s = C \cdot \frac{1 - z^{-1}}{1 + z^{-1}} ,$$

where C is to be determined to have the same 3-dB bandwidth.

$$\therefore c = \omega_r \cot\left(\frac{\omega_r T}{2}\right) = 2 \cot\left(\frac{2 \cdot \left(\frac{1}{10}\right)}{2}\right) \simeq 2\left(\frac{1}{0.1}\right) = 20$$

$$\begin{aligned} \therefore H(z) &= \frac{1}{1 + \frac{1}{2}(20 \frac{1-z^{-1}}{1+z^{-1}})} \\ &= \frac{1}{1 + (10) \frac{1-z^{-1}}{1+z^{-1}}} \\ &= \frac{1+z^{-1}}{(1+z^{-1}) + (10)(1-z^{-1})} \\ &= \frac{1+z^{-1}}{11 - 9z^{-1}} \end{aligned}$$



(d)

$$\begin{aligned}
 H(z) &= \mathcal{L}(\mathcal{L}^{-1}\{\frac{1-e^{-sT_s}}{s}H_a(s)\}|_{t=nT_s}) \\
 &= (1-z^{-1})\mathcal{L}(\mathcal{L}^{-1}(\frac{H_a(s)}{s})|_{t=nT_s}) \\
 &= (1-z^{-1})\mathcal{L}(\mathcal{L}^{-1}(\frac{2}{s(s+2)})|_{t=nT_s}) \\
 &= (1-z^{-1})\mathcal{L}(\mathcal{L}^{-1}(\frac{1}{s}-\frac{1}{s+2})|_{t=nT_s}) \\
 &= (1-z^{-1})\mathcal{L}((u(t)-e^{-2t}u(t))|_{t=nT_s}) \\
 &= (1-z^{-1})\mathcal{L}(u(nT_s)-e^{-2nT_s}u(nT_s)) \\
 &= (1-z^{-1})(\frac{1}{1-z^{-1}}-\frac{1}{1-e^{-2T_s}z^{-1}}) \\
 &= 1 - \frac{1-z^{-1}}{1-e^{-2T_s}z^{-1}}
 \end{aligned}$$

4. (a)

$$\begin{aligned}
 y(t) &= \int_{t-T_s}^t x(\tau)d\tau \quad (\text{by inspection}) \\
 &= \int_{-\infty}^{\infty} x(\tau)h(t-\tau)d\tau \\
 &= x(t) * h(t) \\
 \text{choose } h(t) &= \begin{cases} 1 & 0 \leq t \leq T_s \\ 0 & \text{otherwise} \end{cases} \\
 h(t-\tau) &= \begin{cases} 1 & t-T_s \leq \tau \leq T_s \\ 0 & \text{otherwise} \end{cases} \\
 h(t) &= u(t) - u(t-T_s)
 \end{aligned}$$

(b)

$$\begin{aligned}
 Y(j\omega) &= X(j\omega)H(j\omega) \\
 y(nT_s) &\leftrightarrow \frac{1}{T_s} \sum_{k=-\infty}^{\infty} Y(j(\omega - k\frac{2\pi}{T_s})) \quad (\text{by } FT) \\
 FT\{x[n]\} &= \frac{1}{T_s} \sum_{k=-\infty}^{\infty} X(j(\omega - k\frac{2\pi}{T_s}))H(j(\omega - k\frac{2\pi}{T_s}))
 \end{aligned}$$