

Text book
1.3

Phasor Representation of a Sinusoids

- The basic representation of a real sinusoids is written :-

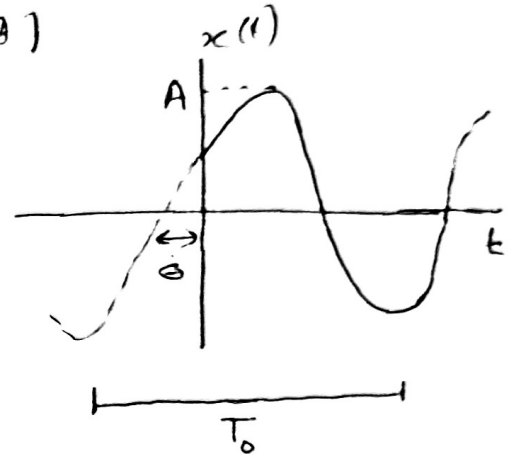
$$x(t) = A \cos(2\pi f_0 t + \theta)$$

$$= A \cos\left(\frac{2\pi}{T_0} t + \theta\right)$$

A: Amplitude

T_0 : Fundamental period

θ : phase shift (angle)



- It is often mathematically convenient to represent real signals in terms of complex quantities
- From Euler's identity
$$e^{jz} = \cos z + j \sin z$$

↑
complex exponential signal
(rotating phasor)
- $x(t)$ can be represented by phasors (or complex exp.)

$$x(t) = \text{Real} \left\{ A e^{j(2\pi f_0 t + \theta)} \right\} = \text{Re} [x_p(t)]$$

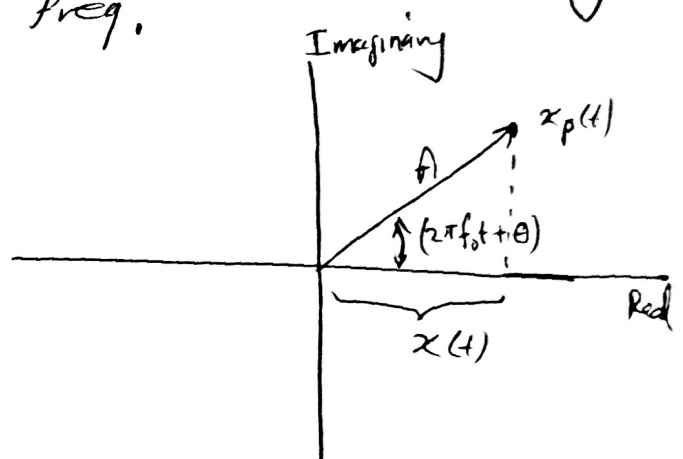
$x_p(t)$ is a complex exponential ~~ph~~ signal and it is periodic

$$x_p(t) = \underbrace{A e^{j\theta}}_{\text{Fixed part}} \underbrace{e^{j2\pi f_0 t}}_{\text{rotating part}}$$

Hence, $x_p(t) = A e^{j\theta} e^{j2\pi f_0 t} = X e^{j2\pi f_0 t}$

$X = A e^{j\theta}$ is called phasor represents and specify $x(t)$ excepts for the frequency information

- phasor representation can be used in combining signals with same freq.



- $X = A e^{j\theta} = A \angle \theta$

- example, let us add 2 currents in parallel

$$i_1 = 6.5 \cos(2\pi \cdot 60 \cdot t - \pi/6)$$

$$i_2 = 2 \cos(2\pi \cdot 60 \cdot t + \pi/40)$$

the phasor representation $I_1 = 6.5 \angle -\pi/6 = 6.5 e^{-j\pi/6}$

$$I_2 = 2 \angle \pi/40 = 2 e^{j\pi/40}$$

also, in rectangular format $I_1 = 5.63 - j3.25$

$$I_2 = 1.99 + j0.16$$

$$\Rightarrow I_{total} = I_1 + I_2 = 7.62 - j3.1 = 8.23 \angle -0.38$$

$$\Rightarrow i(t) = 8.23 \cos(2\pi \cdot 60 \cdot t - 0.38) \quad \#$$

- An alternative phasor representation can be obtained by

$$x(t) = A \cos(2\pi f_0 t + \theta)$$

$$= A \left[\frac{e^{j(2\pi f_0 t + \theta)} + e^{-j(2\pi f_0 t + \theta)}}{2} \right]$$

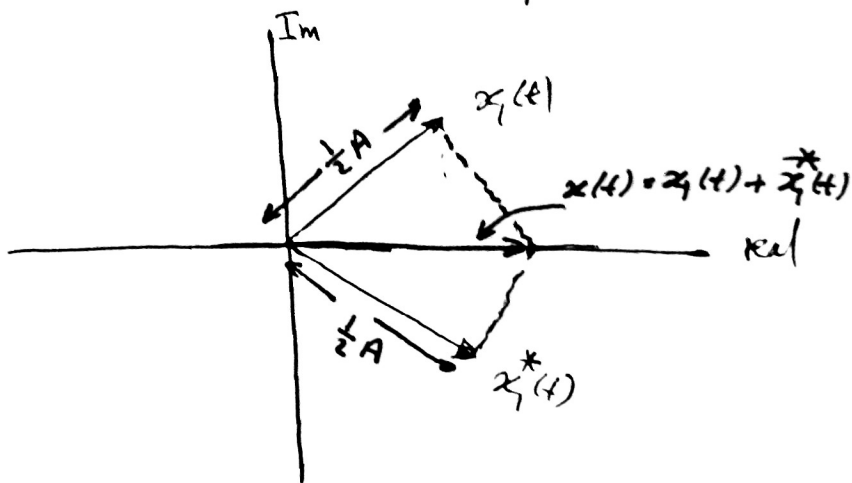
we used Euler again

$$\cos z = \frac{e^{jz} + e^{-jz}}{2}$$

$$\Rightarrow x(t) = \frac{A e^{j\theta}}{2} e^{j2\pi f_0 t} + \frac{A e^{-j\theta}}{2} e^{-j2\pi f_0 t}$$

$$= \frac{X}{2} e^{j2\pi f_0 t} + \frac{X^*}{2} e^{-j2\pi f_0 t} = x_1(t) + x_1^*(t)$$

\uparrow at $+f$ \uparrow at $-f$



Summary : two ways to express $x(t)$ =
 in terms of phasor and complex
 exponential signal. (will be useful)
 in freq. domain

Spectra

- Alternative representation for $x(t)$ which is provided in the freq. domain, whereby, the amplitude and the phase of the signal is plotted ~~is~~ with respect to the freq.

- $x(t) = A \cos(2\pi f_0 t + \theta) = \text{Re} \{ A e^{j(2\pi f_0 t + \theta)} \}$
 $x(t)$ is completely specified by A and θ for a given f_0

also $x_p(t) = A e^{j(2\pi f_0 t + \theta)}$ is completely specified by A and θ for a given f_0 .

However, dealing with complex quantities ~~some~~ most of the time is preferable.

- Frequency domain Representation can take the form of two plots :-
 - ① A as a function of f_0 .
 - ② θ as a function of f_0 .

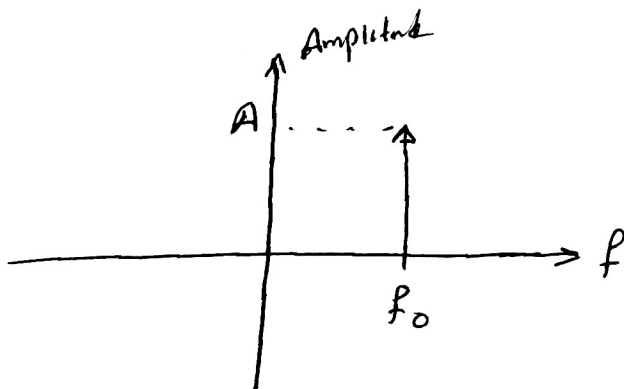
• Based on the two ways discussed earlier to express $x(t) = A \cos(2\pi f_0 t + \theta)$ in terms of complex exponential and phasor, we ~~show~~ such that

$$x(t) = \operatorname{Re} \left\{ A e^{j\theta} e^{j2\pi f_0 t} \right\}$$

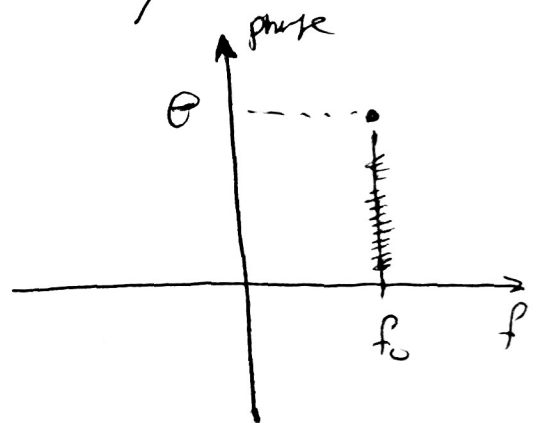
or $x(t) = \frac{A}{2} e^{j\theta} e^{j2\pi f_0 t} + \frac{A}{2} e^{-j\theta} e^{-j2\pi f_0 t}$

we classify the frequency representation into two forms

(a) "Single-sided" spectra which considers only +ve freq.

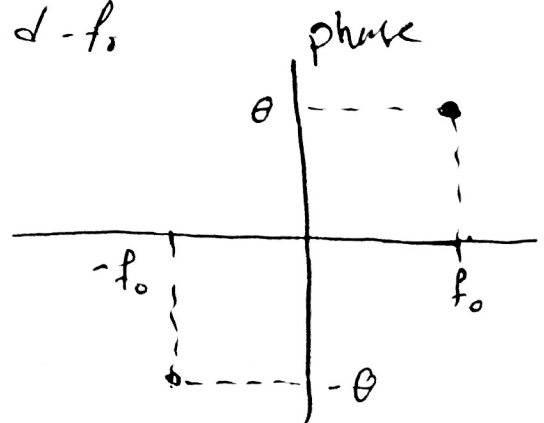
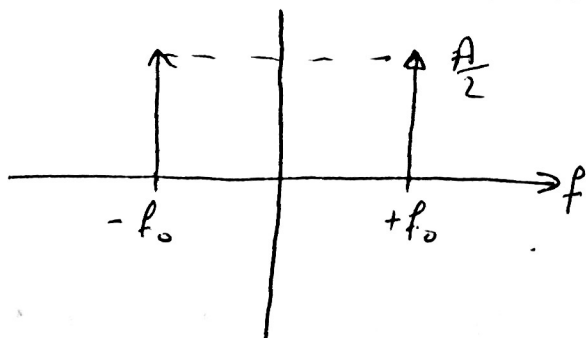


single-sided Amplitude spectra



single-sided phase spectra

(b) "Double-sided" spectra which considers both $+f_0$ and $-f_0$



Example: Given the signal

$$x(t) = 6 \cos(20\pi t - \frac{\pi}{3}) + 4 \sin^2(30\pi t - \frac{\pi}{6})$$

- ① sketch its single-sided amplitude and phase spectra
 ② = = Double-sided :

Sol.

$$x(t) = 6 \cos(20\pi t - \frac{\pi}{3}) + 4 \left[\frac{1}{2} - \frac{1}{2} \cos 2(30\pi t - \frac{\pi}{6}) \right]$$

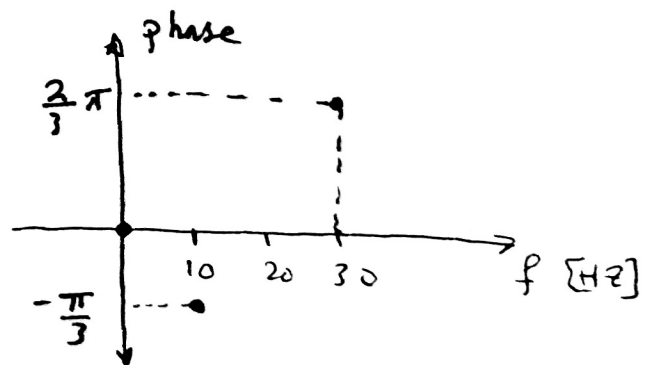
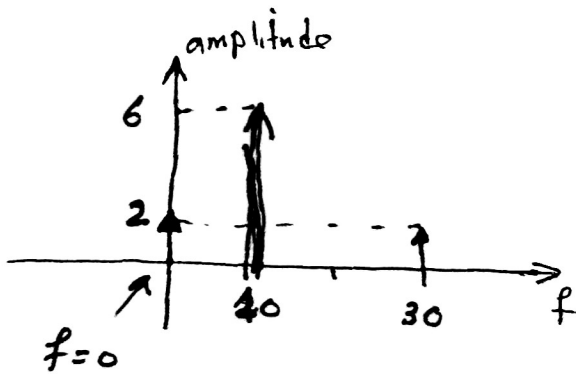
$$= 6 \cos(20\pi t - \frac{\pi}{3}) + 2 - 2 \cos(60\pi t - \frac{\pi}{3})$$

$$= \operatorname{Re} \left\{ 6 e^{-j\frac{\pi}{3}} e^{j20\pi t} \right\} + 2 + \operatorname{Re} \left\{ -2 e^{-j\frac{\pi}{3}} e^{j60\pi t} \right\}$$

$$= \operatorname{Re} \left\{ 6 e^{-j\frac{\pi}{3}} e^{j20\pi t} \right\} + 2 + \operatorname{Re} \left\{ 2 e^{-j\frac{\pi}{3} + \pi} e^{j60\pi t} \right\}$$

note that $-1 = e^{j\pi} = \cos \pi + j \sin \pi$

$$= \operatorname{Re} \left\{ 6 e^{-j\frac{\pi}{3}} e^{j20\pi t} \right\} + 2 + \operatorname{Re} \left\{ 2 e^{j\frac{2}{3}\pi} e^{j60\pi t} \right\}$$



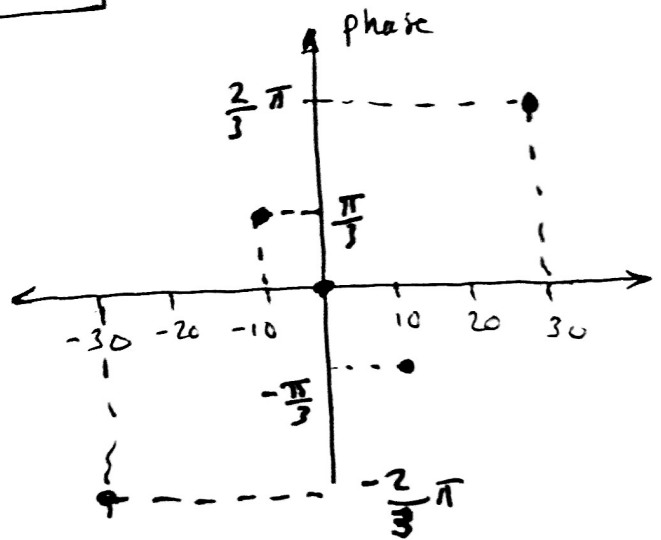
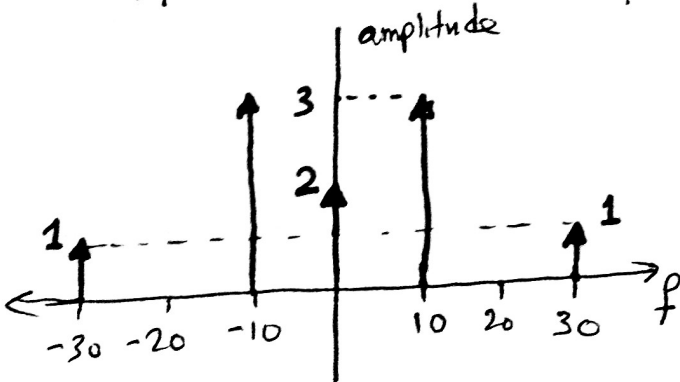
at	Amplitude	phase
$f = 0$	2	0.
$f = 10$	6	$-\frac{\pi}{3}$
$f = 30$	2	$\frac{2}{3}\pi$

② to sketch the double sided:

$$x(t) = 6 \left[\frac{e^{-j\frac{\pi}{3} 20\pi t} e^{j20\pi t} + e^{j\frac{\pi}{3} 20\pi t} e^{-j20\pi t}}{2} \right] + 2$$

$-1 = e^{j\pi}$ \rightarrow $+2e^{j\pi} \left[\frac{e^{-j\frac{\pi}{3} 60\pi t} e^{j60\pi t} + e^{j\frac{\pi}{3} 60\pi t} e^{-j60\pi t}}{2} \right]$

$f = 0$	2	0
$f = 10$	6/2	$-\pi/3$
$f = -10$	6/2	$\pi/3$
$f = 30$	2/2	$2/3 \pi$
$f = -30$	2/2	$-2/3 \pi$



Energy and Power Spectral Densities

- it is useful for some applications to define Functions of Frequency that when integrated over all frequencies give

total Energy if it is energy signal
Average power : : : power signal

- For Energy signal, a fun. of freq. when integrated that gives total energy is referred to as an "Energy spectral Density (ESD)"

denoting the ESD of $x(t)$ by $G(f)$

$$\Rightarrow E = \int_{-\infty}^{\infty} G(f) df$$

note: How to obtain
ESD will be
discussed in CH.4

- For power signal,
denoting the "Power spectral density (PSD)" of
a power signal $x(t)$ by $S(f)$

$$\Rightarrow P = \int_{-\infty}^{\infty} S(f) df$$

↑
average
power

Generalizing:-

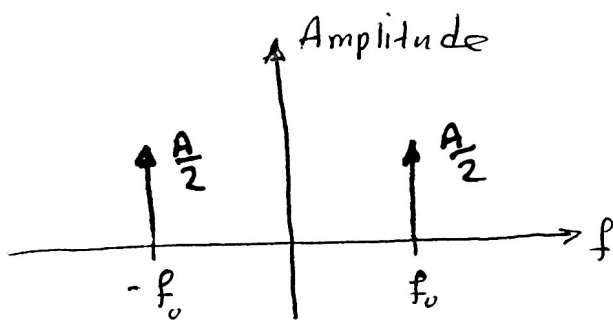
to obtain the PSD: for any signal possessing a two sided line (amplitude) spectrum, we obtain the corresponding PSD by:

- ① taking each line of the amplitude spectrum
- ② squaring the line value
- ③ Multiply the squared value by unit impulse located at that particular frequency.

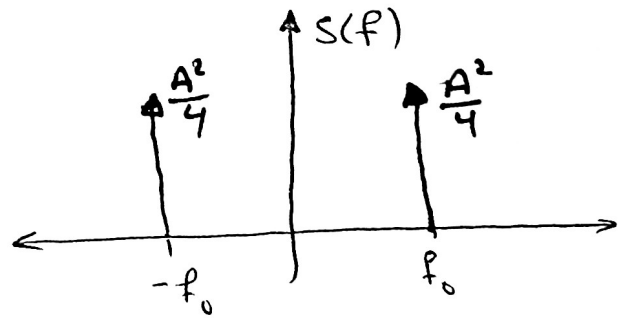
Example: The amplitude (Duple-sided spectrum) for periodic signal

$$x(t) = A \cos(2\pi f_0 t + \theta)$$

$$= \frac{A}{2} e^{j\theta} e^{j2\pi f_0 t} + \frac{A}{2} e^{-j\theta} e^{-j2\pi f_0 t}$$



\Rightarrow



One can show that

$$P = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T |x(t)|^2 dt = \left(\frac{A^2}{2} \right)$$

$$S(f) = \frac{A^2}{4} \delta(f+f_0) + \frac{A^2}{4} \delta(f-f_0)$$

Also

$$P = \int_{-\infty}^{\infty} S(f) df = \int \frac{A^2}{4} \delta(f+f_0) + \int \frac{A^2}{4} \delta(f-f_0) df$$

$$= \frac{A^2}{4} + \frac{A^2}{4} = \left(\frac{A^2}{2} \right)$$

Note that :-

- ① PSD of any signal is an even function of frequency
- ② PSD possesses no phase information about the signal.

Exercise : ① Find and plot the PSD for
 $x(t) = 6 \cos(20\pi t - \frac{\pi}{3}) + 4 \sin^2(30\pi t - \frac{\pi}{6})$

- ② Find the average power.