

# Bipolar Junction Transistors

Instructor  
Nasser Ismail



# The First Transistor



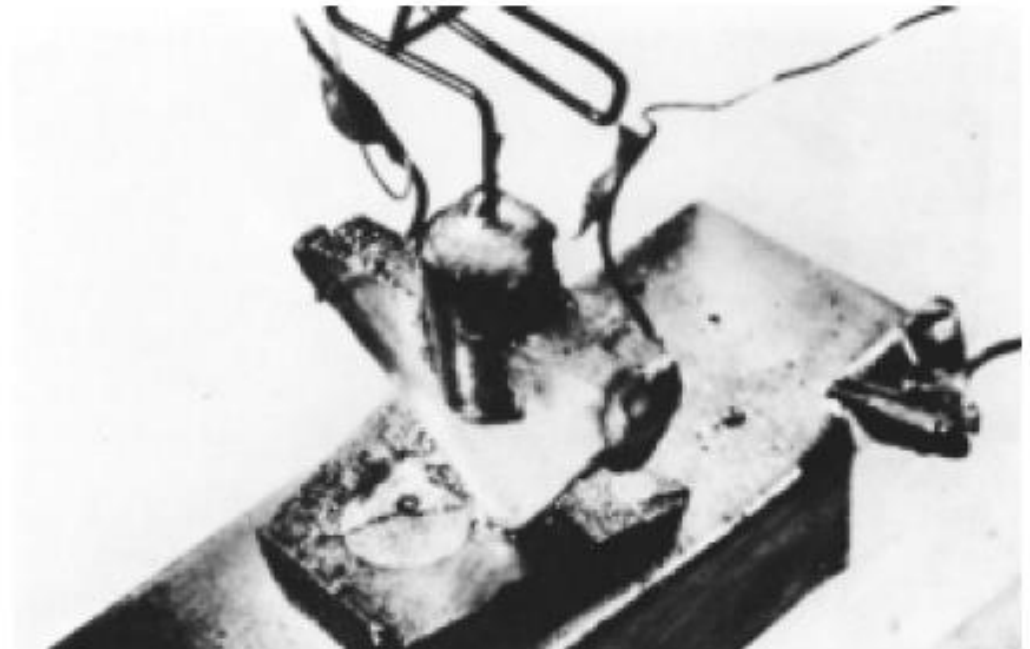
Co-inventors of the first transistor at Bell Laboratories: Dr. William Shockley (seated); Dr. John Bardeen (left); Dr. Walter H. Brattain. (Courtesy of AT&T Archives.)

**Dr. Shockley** Born: London, England, 1910  
PhD Harvard, 1936

**Dr. Bardeen** Born: Madison, Wisconsin, 1908  
PhD Princeton, 1936

**Dr. Brattain** Born: Amoy, China, 1902  
PhD University of Minnesota, 1928

All shared the Nobel Prize in 1956 for this contribution.

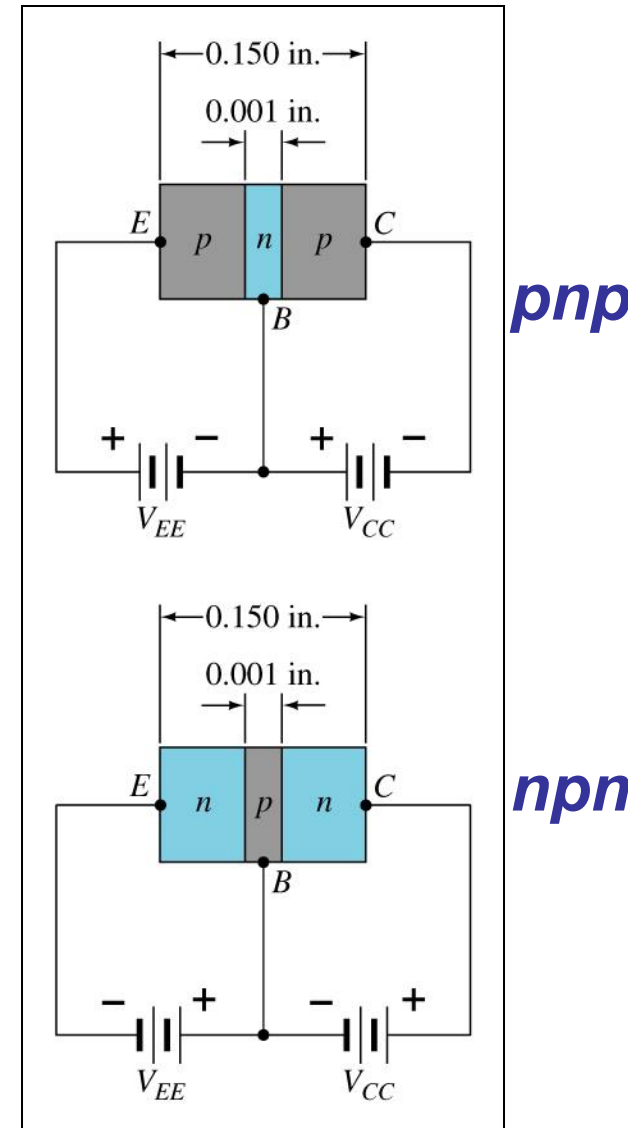


E

**Figure 3.1** The first transistor. (Courtesy Bell Telephone Laboratories.)

# Construction

- 3-Layer Semiconductor device
- 2 p- layers and one n-layer or vice versa
- pnp or npn types
- Two pn junctions , each of them can be either forward or reverse biased
- This results in 4 possible modes of operation



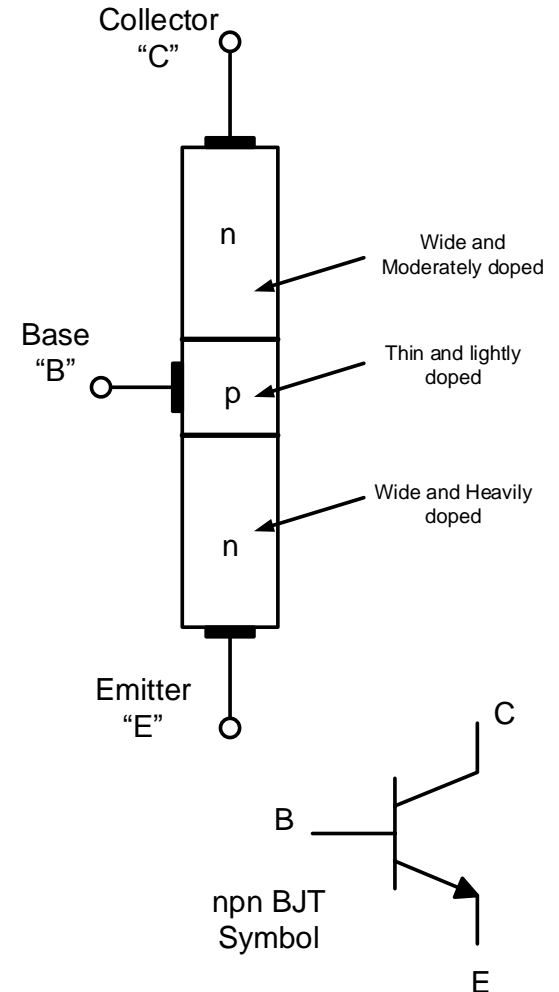
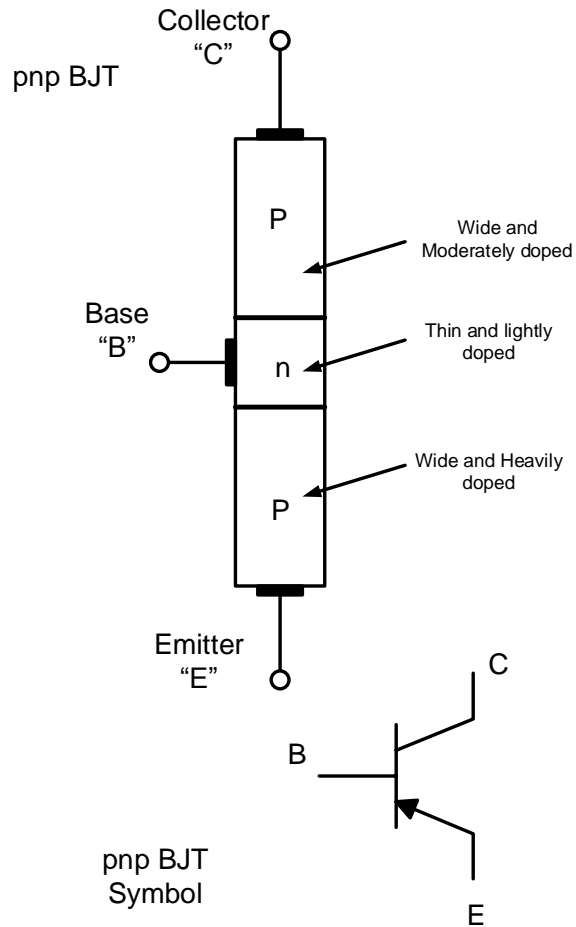
# npn transistor modes of operation

Junction/ Mode	BE	BC	Remarks
<b>Saturation Mode</b>	Forward	Forward	Equivalent to short circuit $I_c = I_c(\text{sat})$ $V_{ce} = V_{ce}(\text{sat}) \approx 0.2V$
<b>Active Mode (Linear Region)</b>	Forward	Reverse	$I_c$ proportional to $I_b$ $V_{ce}$ defined by circuit
<b>Cut-off Mode</b>	Reverse	Reverse	Equivalent to open circuit $I_c = I_b = 0$ $V_{ce}$ defined by circuit
Inverse Mode	Reverse	Forward	Rarely used and will not be discussed in this course

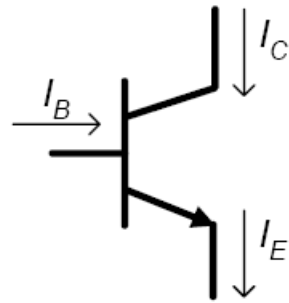
# pnp transistor modes of operation

Junction/ Mode	EB	CB	Remarks
<b>Saturation Mode</b>	Forward	Forward	Equivalent to short circuit $I_c = I_c(\text{sat})$ $V_{ce} = V_{ce}(\text{sat}) \approx 0.2\text{V}$
<b>Active Mode (Linear Region)</b>	Forward	Reverse	$I_c$ proportional to $I_b$ $V_{ce}$ defined by circuit
<b>Cut-off Mode</b>	Reverse	Reverse	Equivalent to open circuit $I_c = I_b = 0$ $V_{ce}$ defined by circuit
Inverse Mode	Reverse	Forward	Rarely used and will not be discussed in this course

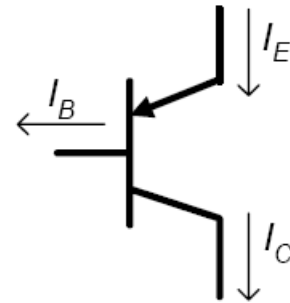
# Construction and Symbol



- Current Directions



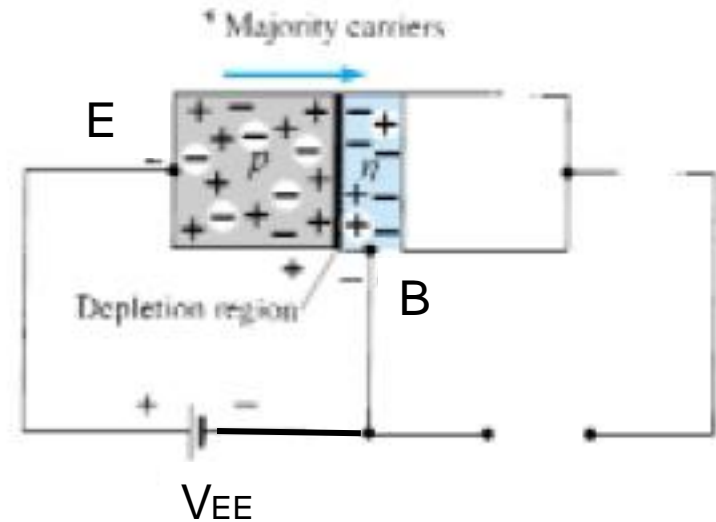
npn



pnp

# Transistor Operation in the Active Mode(pnp CB)

- EB is forward biased
- CB not biased
- This is similar to the forward biased diode
- The depletion region is reduced due to applied bias resulting in heavy flow of majority carriers (+) from p-layer to n-layer

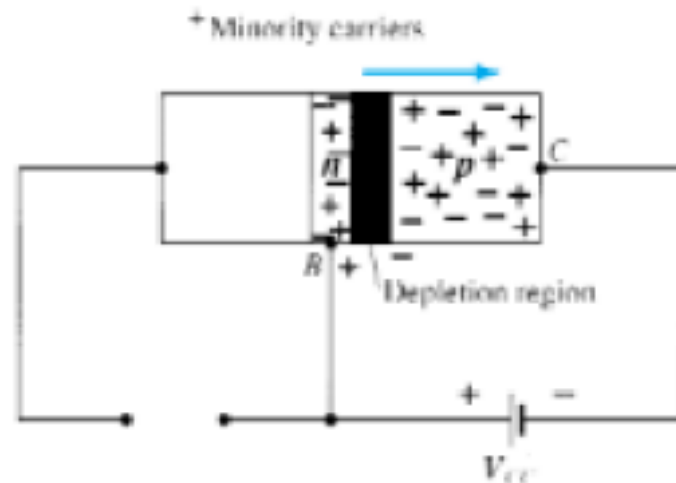


Forward Biased EB Junction of a pnp transistor



# Transistor Operation

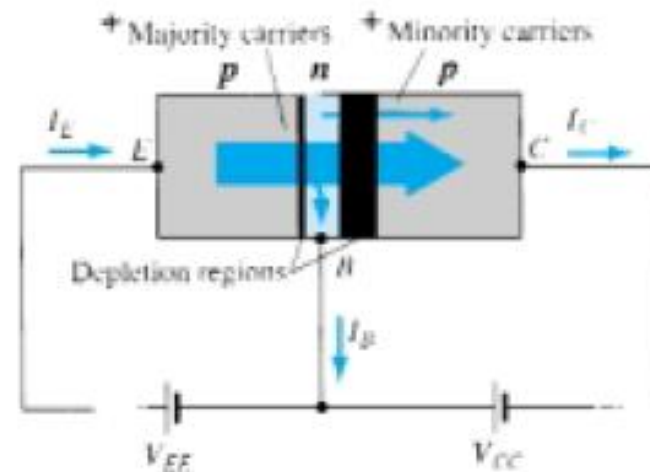
- 5) Remove EB bias and apply reverse bias on CB junction resulting in widening of the depletion region
- 6) Recall that the flow of majority carriers is zero, resulting in only a minority-carrier flow



**Figure 3.4** Reverse-biased junction of a *pnp* transistor.

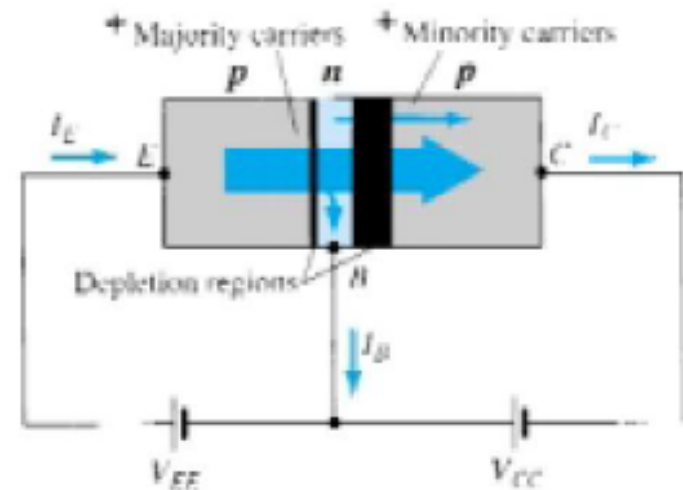
# Transistor Operation

- If the two previous sources were applied with one junction FB and the other reverse biased
- a large number of majority carriers will diffuse across the forward-biased  $p$ - $n$  junction into the  $n$ -type material.
- Since the base is very thin and has a low conductivity, a very small number of these carriers will take this path of high resistance to the base terminal
- The magnitude of the base current is typically on the order of microamperes as compared to milliamperes for the emitter and collector currents



- The larger number of these majority carriers will diffuse across the reverse-biased junction into the  $p$ -type material connected to the collector terminal as indicated in Fig. 3.5.
- The reason for the relative ease with which the majority carriers can cross the reverse-biased junction is easily understood if we consider that for the reverse-biased diode **the injected majority carriers will appear as minority carriers in the  $n$ -layer.**
- In other words, there has been an *injection* of minority carriers into the  $n$ -type base region material.

## Transistor Operation

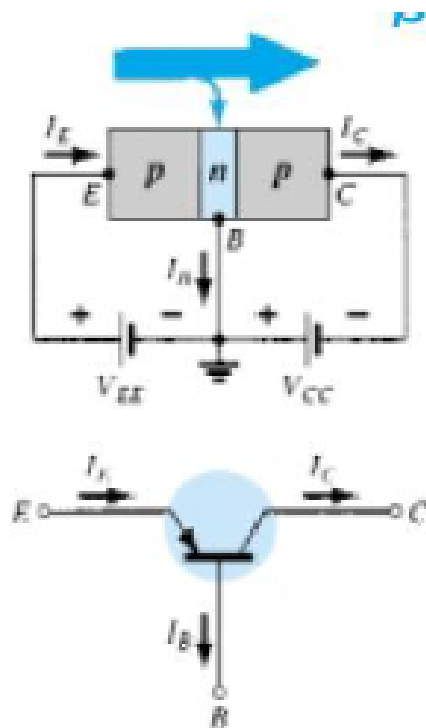


**Figure 3.5** Majority and minority carrier flow of a *pnp* transistor.

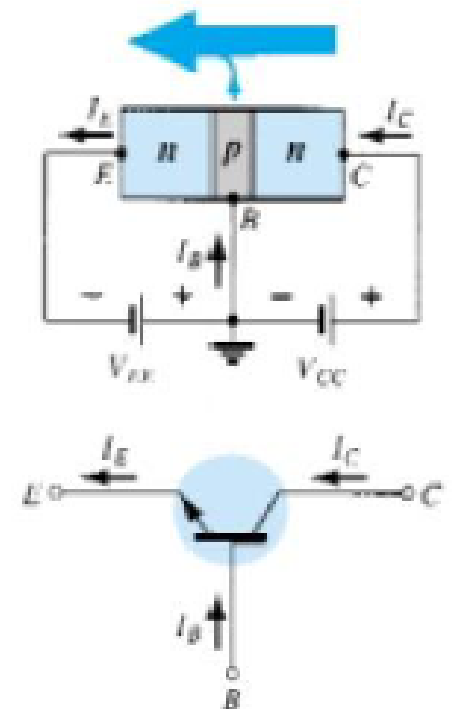
- Combining this with the fact that all the minority carriers in the depletion region will cross the reverse-biased junction of a diode accounts for the flow indicated

# Basic Current Equation

- Applying KCL to the transistor as if it were a single node, we obtain

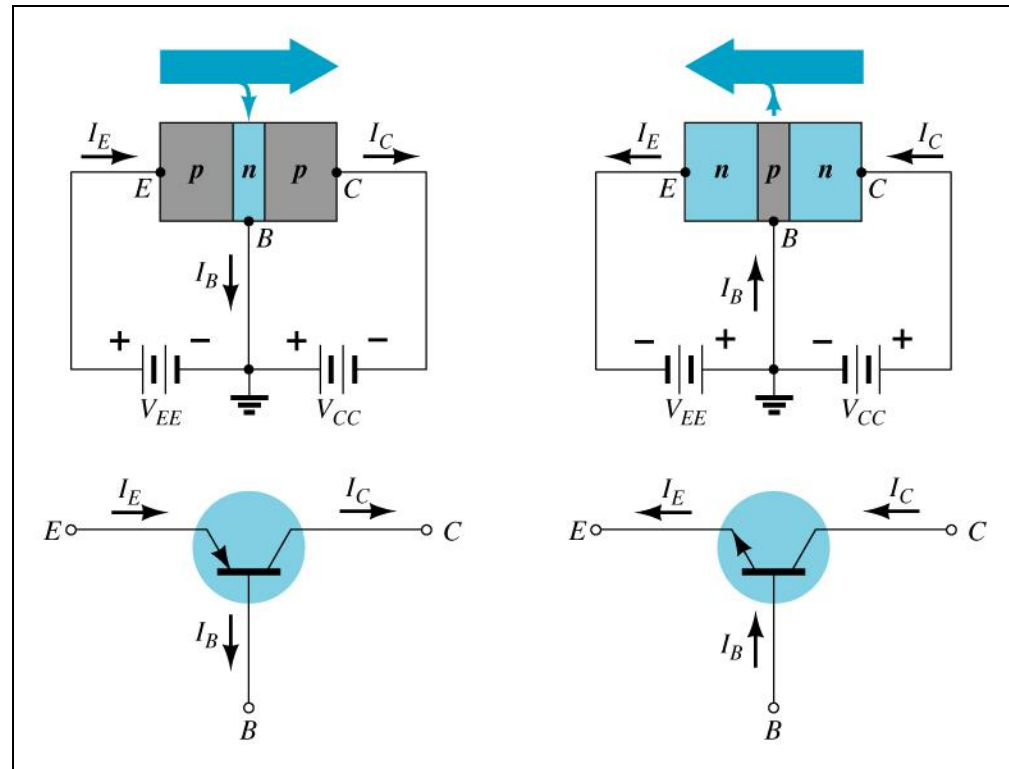


$$I_E = I_C + I_B$$



# Common Base Configuration

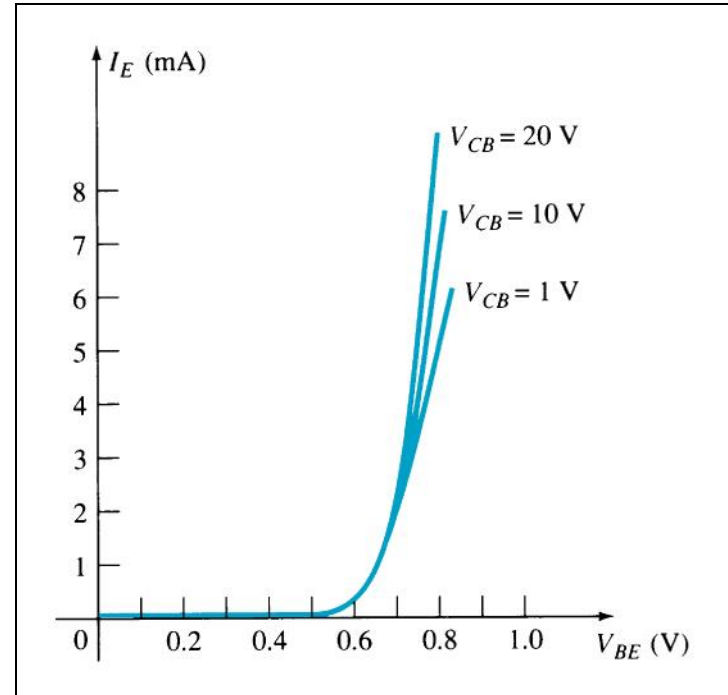
- The common-base terminology is derived from the fact that the base is common to both the input and output sides of the configuration
- In addition, the base is usually the terminal closest to , or at, ground potential) through the device
- Throughout this course all current directions will refer to conventional (hole) flow rather than electron flow
- The arrow in the graphic symbol defines the direction of emitter current(conventional flow) through the device



# Common-Base Amplifier

## Input Characteristics

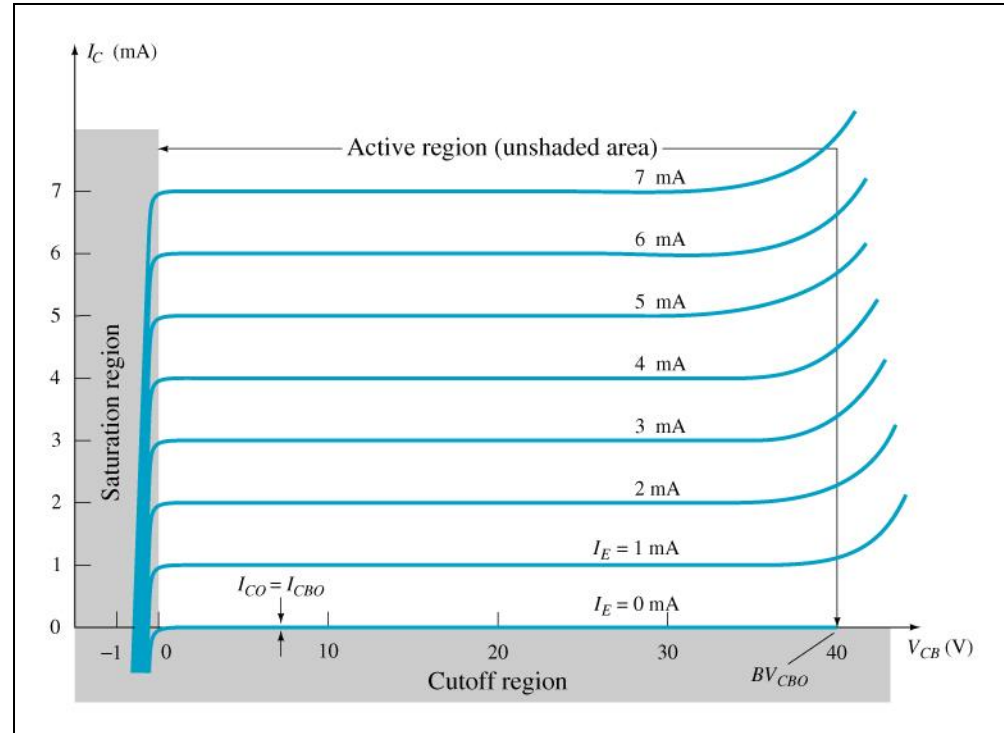
This curve shows the relationship between of input current ( $I_E$ ) to input voltage ( $V_{BE}$ ) for three output voltage ( $V_{CB}$ ) levels.



# Common-Base Amplifier

## Output Characteristics

This graph demonstrates the output current ( $I_C$ ) to an output voltage ( $V_{CB}$ ) for various levels of input current ( $I_E$ ).







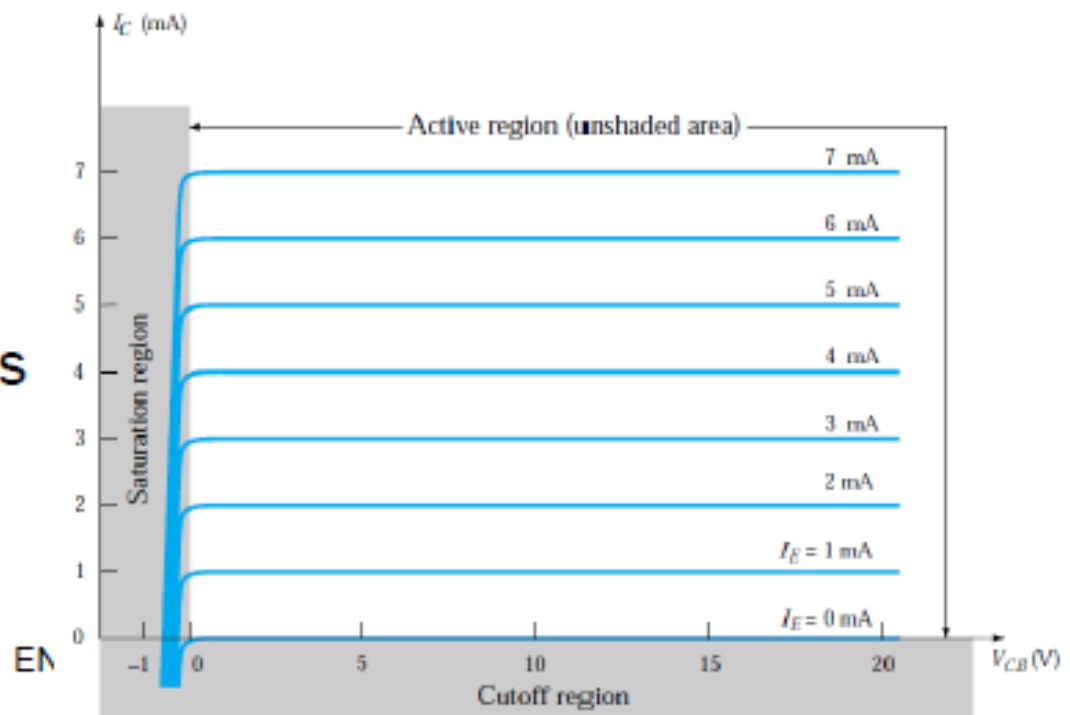
# Operation Modes

- Cutoff region: defined as that region where collector current is 0A
- In this region both junctions are reverse biased
- Saturation region is defined as the region where both junctions are forward biased

# Operation Modes of BJT

- **The saturation region** is defined as that region of the characteristics to the left of  $V_{CB} = 0$  V.
- The horizontal scale in this region was expanded to clearly show the dramatic change in characteristics in this region.

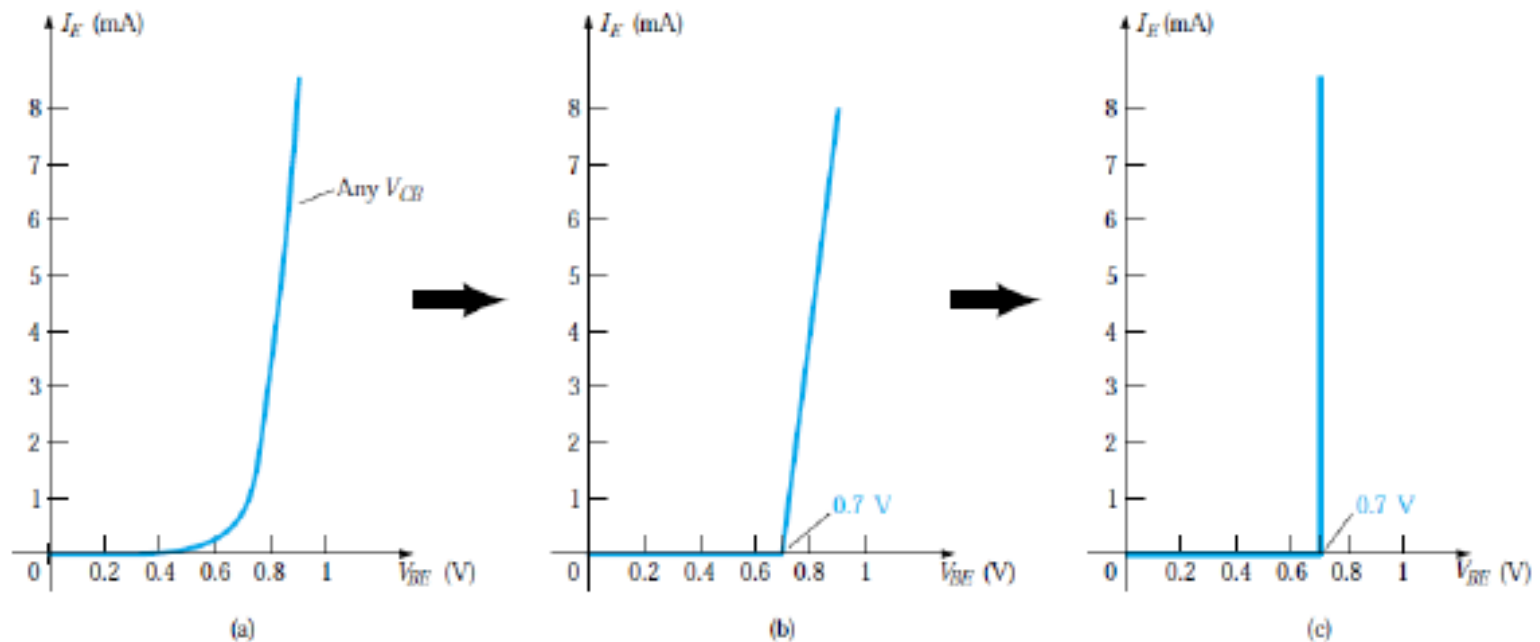
Note the exponential increase in collector current as the voltage  $V_{CB}$  increases toward 0 V.



# Operation Modes of BJT

- Once a transistor is in the “on” state, the base-to-emitter voltage will be assumed to be the following:

$$V_{BE} = 0.7 \text{ V}$$



## Ch.3 Summary

# Approximations

**Emitter and collector currents:**

$$I_C \cong I_E$$

**Base-emitter voltage:**

$$V_{BE} = 0.7 \text{ V (for Silicon)}$$

# Alpha ( $\alpha$ )

Alpha ( $\alpha$ ) is the ratio of  $I_C$  to  $I_E$ :

$$\alpha_{dc} = \frac{I_C}{I_E}$$

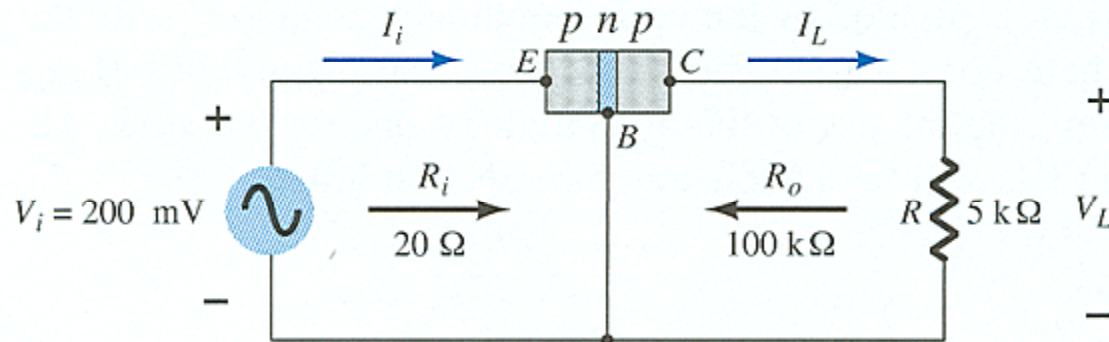
Ideally:  $\alpha = 1$

In reality:  $\alpha$  falls somewhere between 0.9 and 0.998

Alpha ( $\alpha$ ) in the AC mode:

$$\alpha_{ac} = \frac{\Delta I_C}{\Delta I_E}$$

# Transistor Amplifier



**Currents and Voltages:**

$$I_E = I_i = \frac{V_i}{R_i} = \frac{200 \text{ mV}}{20 \Omega} = 10 \text{ mA}$$

$$I_C \cong I_E$$

$$I_L \cong I_i = 10 \text{ mA}$$

$$V_L = I_L R = (10 \text{ mA})(5 \text{ k}\Omega) = 50 \text{ V}$$

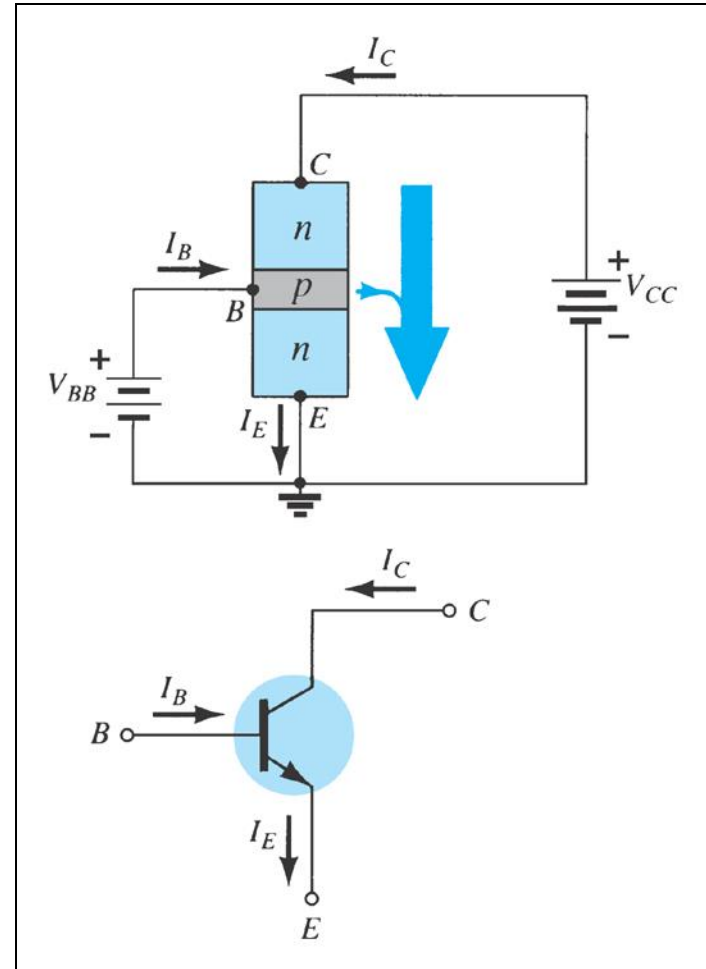
**Voltage Gain:**

$$A_V = \frac{V_L}{V_i} = \frac{50 \text{ V}}{200 \text{ mV}} = 250$$

# Common-Emitter Configuration

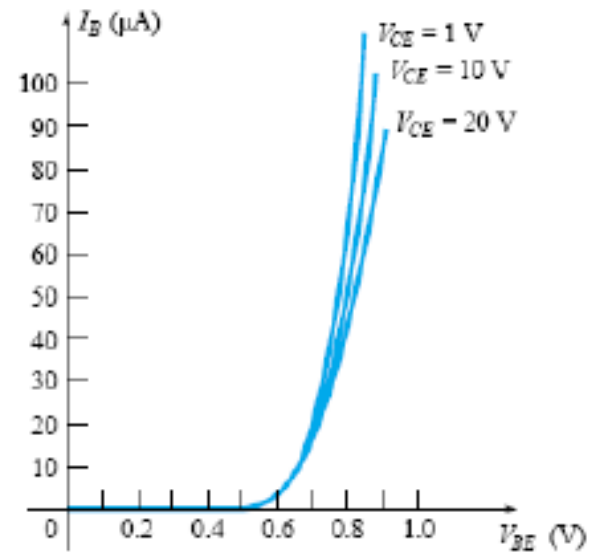
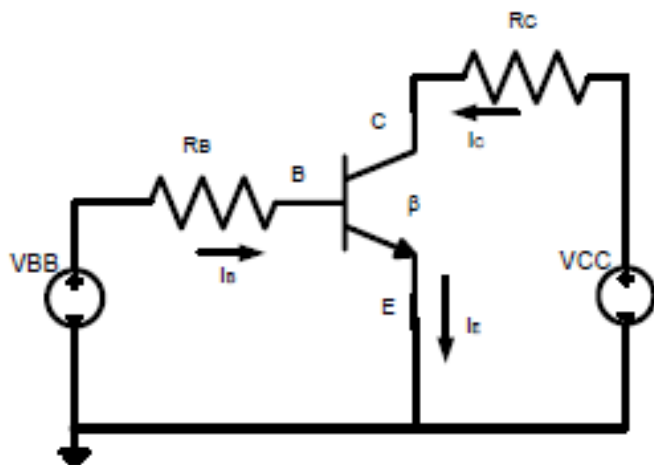
The emitter is common to both input (base-emitter) and output (collector-emitter) circuits.

The input is applied to the base and the output is taken from the collector.



# Common Emitter Configuration

Input characteristics : Input current ( $I_B$ ) versus the input voltage ( $V_{BE}$ ) for a range of values of output voltage ( $V_{CE}$ ).

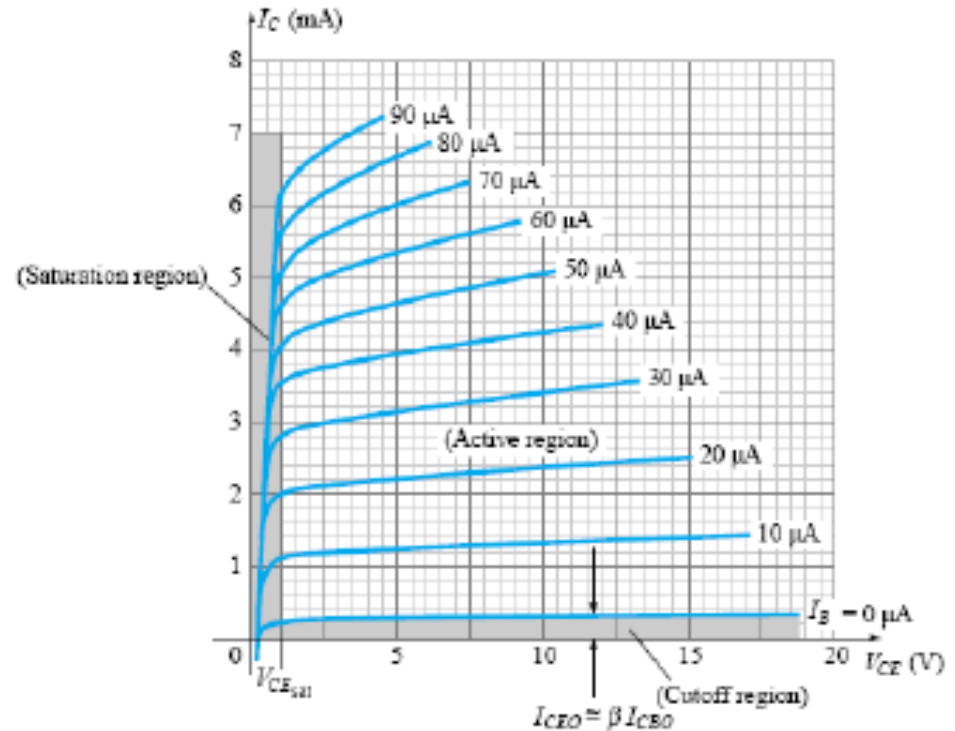




# CE output Characteristic (npn)

## Active Region:

- is that portion of the upper-right quadrant that has the greatest linearity, that is, that region in which the curves for  $I_B$  are nearly straight and equally spaced.
- This region exists to the right of the vertical dashed line at  $V_{CEsat}$  and above the curve for  $I_B$  equal to zero.
- BE forward & BC is reverse
- Note  $I_B$  is small compared to  $I_C$ . ( $I_B \ll I_C$ )
- Transistor amplifies current  
 $I_C = \beta I_B$

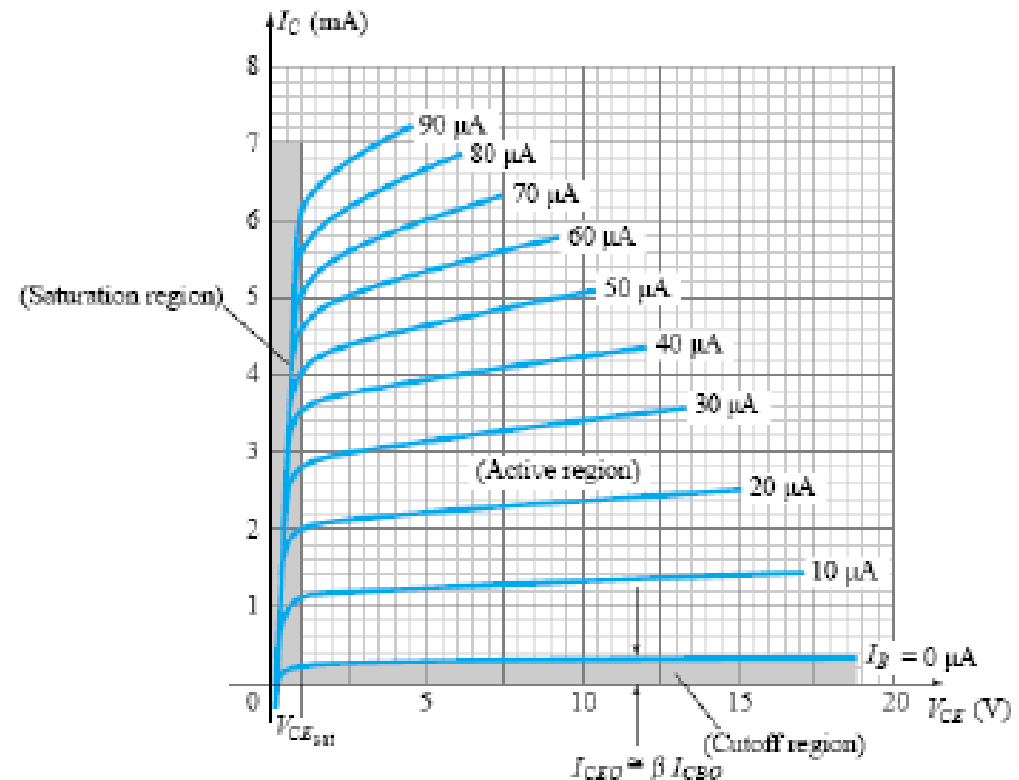


# CE output Characteristic (npn)

## Cutoff Region:

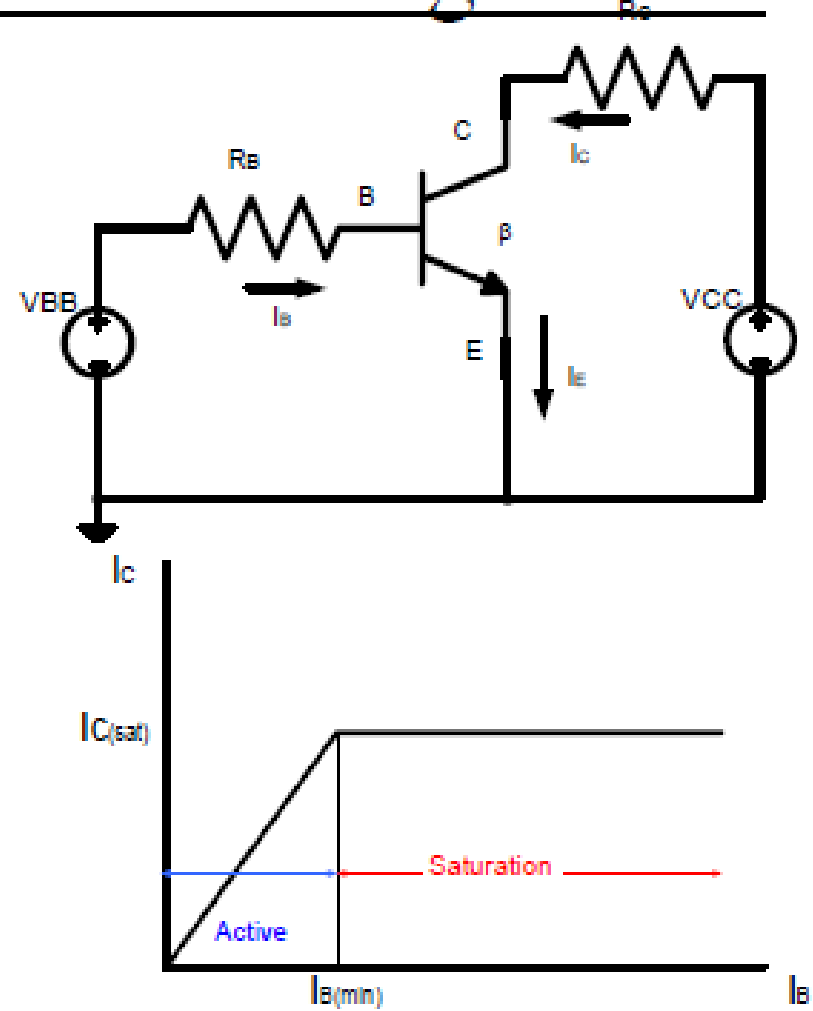
- This region exists above the curve for  $I_B$  equal to zero.
- BE reverse & BC is reverse
- Note  $I_B = 0$  while  $I_C$  is only the reverse leakage current compared
- Transistor acts as an open switch

Saturation Region: BE & BC both forward and  $I_B$  increase

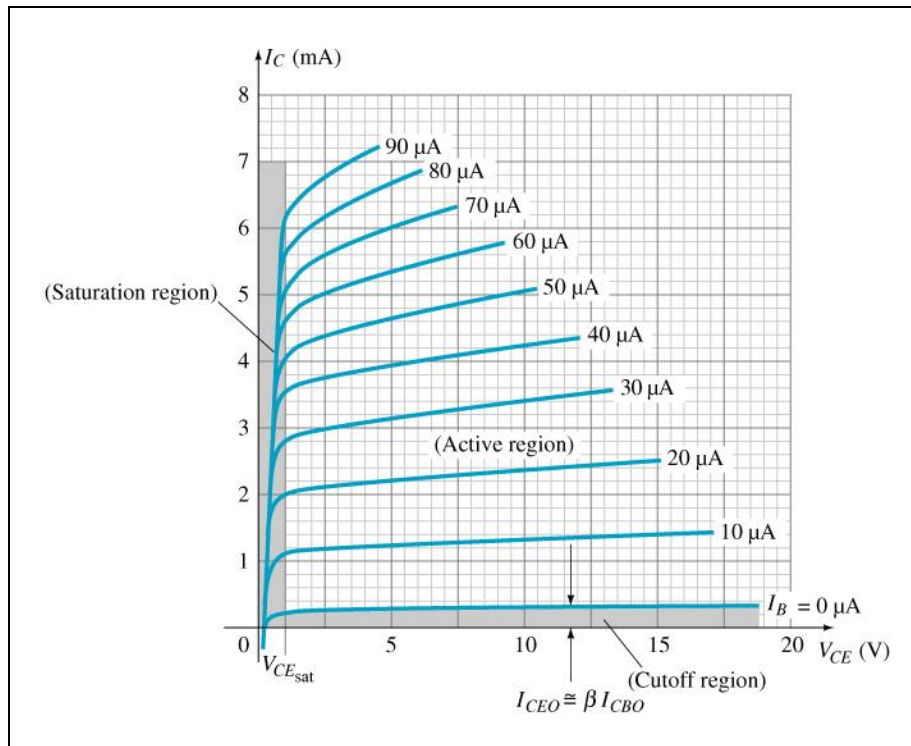


# Saturation Mode for CE Configuration

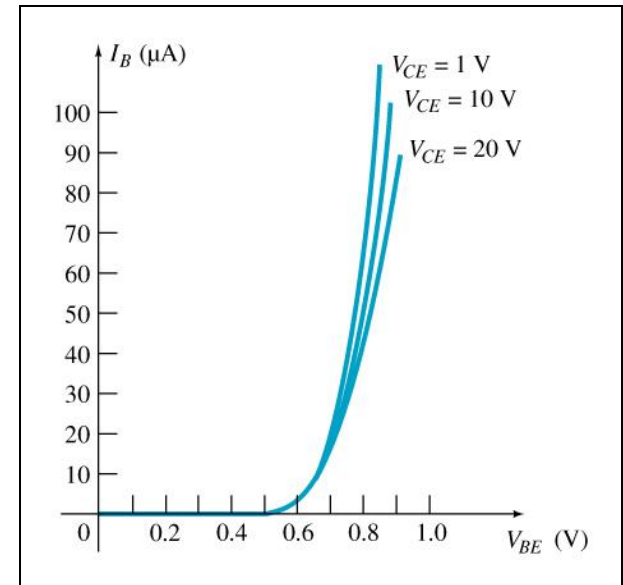
- BE forward & BC become forward
- As  $I_B$  increase and  $I_C$  increase
- Since  $V_{CE} = V_{CC} - I_C R_C$   $V_{CE}$  becomes smaller since  $I_C R_C$  increase
- $V_{CE}$  reaches its saturation value  $V_{CE(sat)} = 0.2V$
- Now BC become forward and  $I_C$  cannot increase any more even with increase of  $I_B$  and the relationship between them  $I_C = \beta I_B$  is not valid anymore
- Transistor acts as a closed switch
- $I_B > I_B(\min) \rightarrow I_B(\min) = I_{C(sat)} / \beta$   
 $V_{CE} = V_{CE(sat)}$   
 $I_C = I_{C(sat)} = [V_{CC} - V_{CE(sat)}] / R_C$



# Common-Emitter Characteristics



**Collector Characteristics**



**Base Characteristics**

# Common-Emitter Amplifier Currents

## Ideal Currents

$$I_E = I_C + I_B$$

$$I_C = \alpha I_E$$

## Actual Currents

$$I_C = \alpha I_E + I_{CBO}$$

where  $I_{CBO}$  = minority collector current

$I_{CBO}$  is usually so small that it can be ignored, except in high power transistors and in high temperature environments.

When  $I_B = 0 \mu\text{A}$  the transistor is in cutoff, but there is some minority current flowing called  $I_{CEO}$ .

$$I_{CEO} = \frac{I_{CBO}}{1 - \alpha} \Big|_{I_B = 0 \mu\text{A}}$$

# Beta ( $\beta$ )

$\beta$  represents the amplification factor of a transistor.

*In DC mode:*

$$\beta_{dc} = \frac{I_C}{I_B}$$

*In AC mode:*

$$\beta_{ac} = \left. \frac{\Delta I_C}{\Delta I_B} \right|_{V_{CE}=\text{constant}}$$

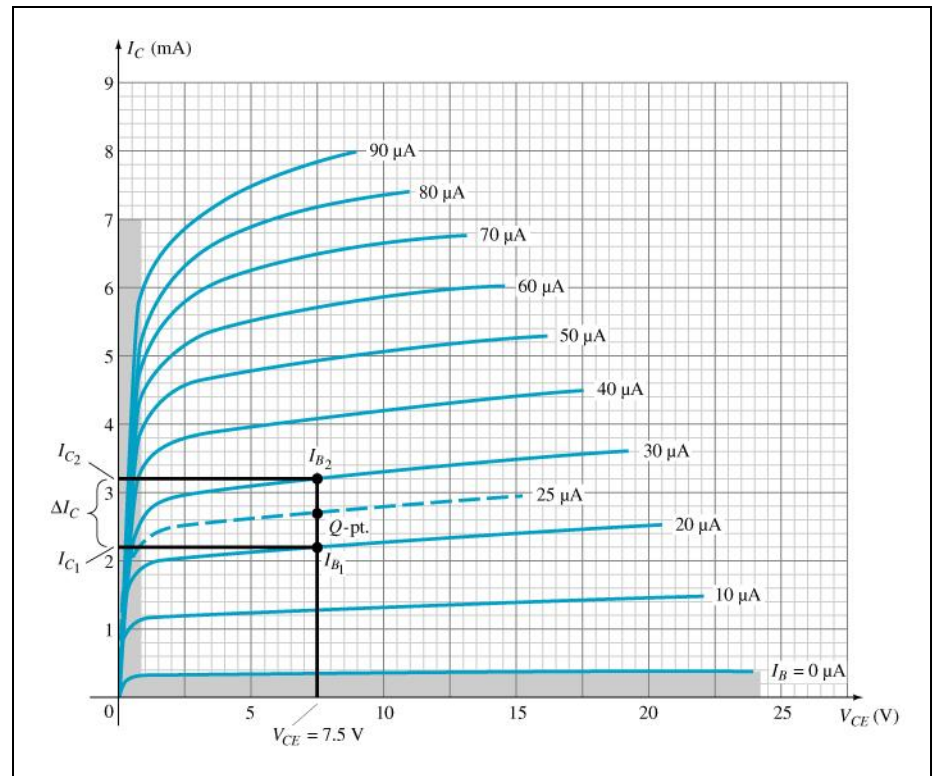
$\beta_{ac}$  is sometimes referred to as  $h_{fe}$ , a term used in transistor modeling calculations

# Beta ( $\beta$ )

## Determining $\beta$ from a Graph

$$\begin{aligned}\beta_{AC} &= \frac{(3.2 \text{ mA} - 2.2 \text{ mA})}{(30 \mu\text{A} - 20 \mu\text{A})} \\ &= \frac{1 \text{ mA}}{10 \mu\text{A}} \Big|_{V_{CE}=7.5 \text{ V}} \\ &= 100\end{aligned}$$

$$\begin{aligned}\beta_{DC} &= \frac{2.7 \text{ mA}}{25 \mu\text{A}} \Big|_{V_{CE}=7.5 \text{ V}} \\ &= 108\end{aligned}$$



# Beta ( $\beta$ )

Relationship between amplification factors  $\beta$  and  $\alpha$  :

$$\alpha = \frac{\beta}{\beta + 1}$$

$$\beta = \frac{\alpha}{\alpha - 1}$$

Relationship Between Currents:

$$I_C = \beta I_B$$

$$I_E = (\beta + 1) I_B$$

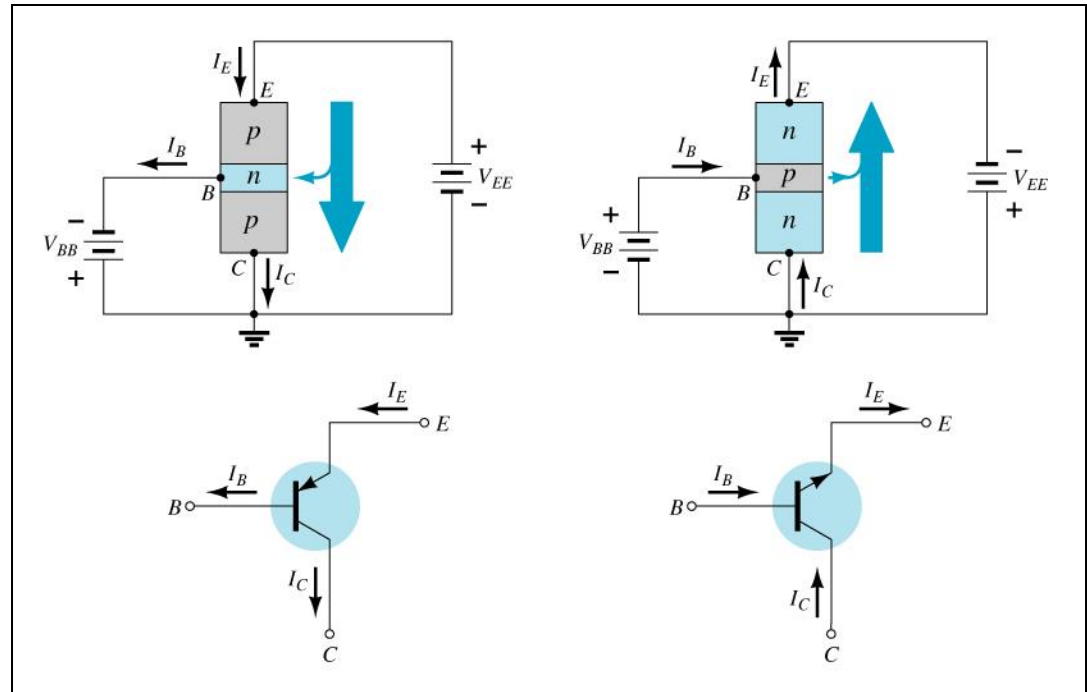


# Common Collector Configuration

- The third and final transistor configuration is the *common-collector configuration*,
- The common-collector configuration is used primarily for impedance-matching purposes since it has a high input impedance and low output impedance, opposite to that of the common-base and common-emitter configurations.

# Common-Collector Configuration

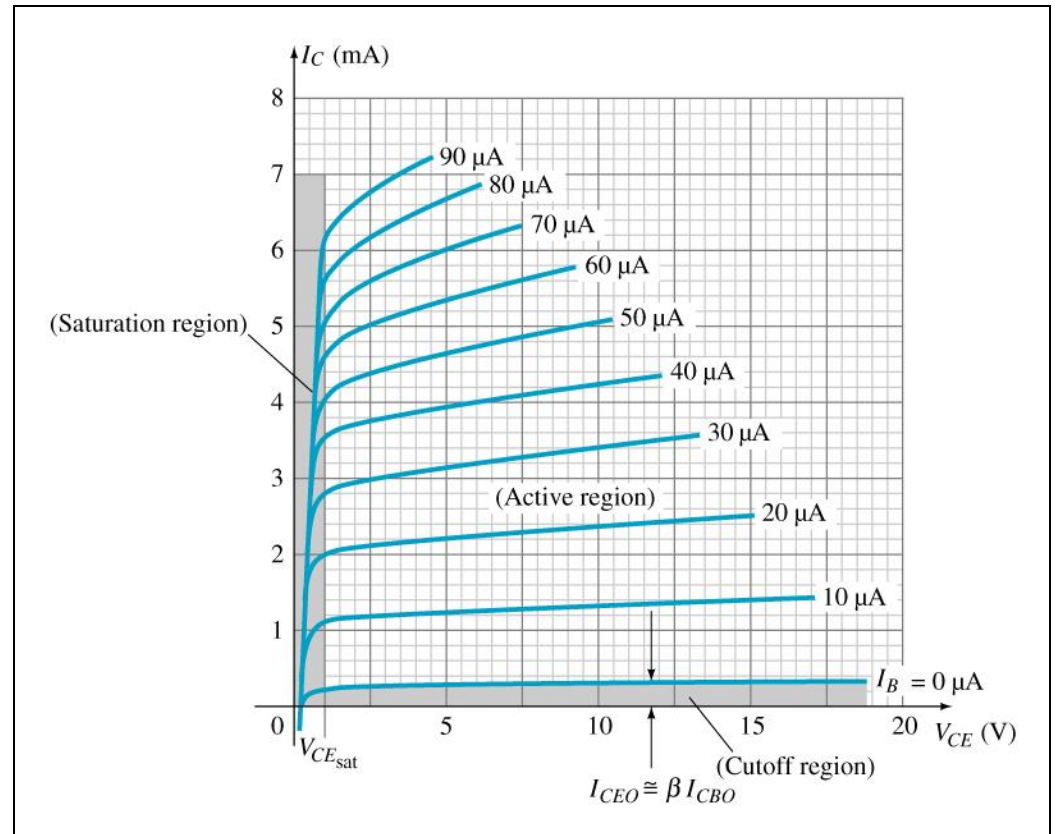
The input is on the base and the output is on the emitter.



- From a design viewpoint, there is no need for a set of common collector characteristics to choose the parameters of the circuit.
- It can be designed using the common-emitter characteristics.
- For all practical purposes, the output characteristics of the common-collector configuration are the same as for the common-emitter configuration.
- For the common-collector configuration the output characteristics are a plot of  $I_E$  versus  $V_{EC}$  for a range of values of  $I_B$ .
- The input current, therefore, is the same for both the common-emitter and common collector
- characteristics

# Common-Collector Configuration

The characteristics are similar to those of the common-emitter amplifier, except the vertical axis is  $I_E$ .



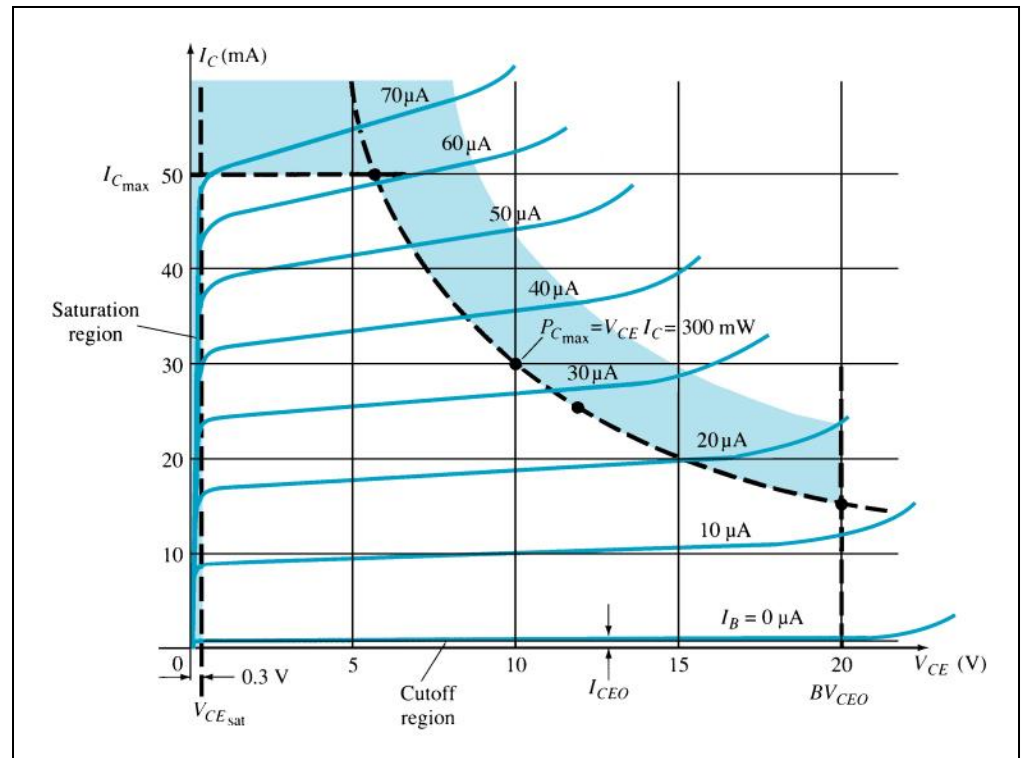
# Operating Limits

$V_{CE}$  is maximum and  $I_C$  is minimum in the cutoff region.

$$I_{C(\max)} = I_{CEO}$$

$I_C$  is maximum and  $V_{CE}$  is minimum in the saturation region.

$$V_{CE(\max)} = V_{CE(\text{sat})} = V_{CEO}$$



The transistor operates in the active region between saturation and cutoff.

# Power Dissipation

**Common-base:**

$$P_{Cmax} = V_{CB} I_C$$

**Common-emitter:**

$$P_{Cmax} = V_{CE} I_C$$

**Common-collector:**

$$P_{Cmax} = V_{CE} I_E$$

# Transistor Specification Sheet

## ELECTRICAL CHARACTERISTICS ( $T_A = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Max	Unit
----------------	--------	-----	-----	------

### OFF CHARACTERISTICS

Collector-Emitter Breakdown Voltage (1) ( $I_C = 1.0 \text{ mAdc}$ , $I_E = 0$ )	$V_{(BR)CEO}$	30		Vdc
Collector-Base Breakdown Voltage ( $I_C = 10 \mu\text{Adc}$ , $I_E = 0$ )	$V_{(BR)CBO}$	40		Vdc
Emitter-Base Breakdown Voltage ( $I_E = 10 \mu\text{Adc}$ , $I_C = 0$ )	$V_{(BR)EBO}$	5.0	–	Vdc
Collector Cutoff Current ( $V_{CB} = 20 \text{ Vdc}$ , $I_E = 0$ )	$I_{CBO}$	–	50	nAdc
Emitter Cutoff Current ( $V_{BE} = 3.0 \text{ Vdc}$ , $I_C = 0$ )	$I_{EBO}$	–	50	nAdc

### ON CHARACTERISTICS

DC Current Gain(1) ( $I_C = 2.0 \text{ mAdc}$ , $V_{CE} = 1.0 \text{ Vdc}$ ) ( $I_C = 50 \text{ mAdc}$ , $V_{CE} = 1.0 \text{ Vdc}$ )	$h_{FE}$	50 25	150 –	–
Collector-Emitter Saturation Voltage(1) ( $I_C = 50 \text{ mAdc}$ , $I_B = 5.0 \text{ mAdc}$ )	$V_{CE(sat)}$	–	0.3	Vdc
Base-Emitter Saturation Voltage(1) ( $I_C = 50 \text{ mAdc}$ , $I_B = 5.0 \text{ mAdc}$ )	$V_{BE(sat)}$	–	0.95	Vdc

# Transistor Specification Sheet

## SMALL-SIGNAL CHARACTERISTICS

Current-Gain – Bandwidth Product ( $I_C = 10 \text{ mAdc}$ , $V_{CE} = 20 \text{ Vdc}$ , $f = 100 \text{ MHz}$ )	$f_T$	250		MHz
Output Capacitance ( $V_{CB} = 5.0 \text{ Vdc}$ , $I_E = 0$ , $f = 100 \text{ MHz}$ )	$C_{obo}$	–	4.0	pF
Input Capacitance ( $V_{BE} = 0.5 \text{ Vdc}$ , $I_C = 0$ , $f = 100 \text{ kHz}$ )	$C_{ibo}$	–	8.0	pF
Collector-Base Capacitance ( $I_E = 0$ , $V_{CB} = 5.0 \text{ V}$ , $f = 100 \text{ kHz}$ )	$C_{cb}$	–	4.0	pF
Small-Signal Current Gain ( $I_C = 2.0 \text{ mAdc}$ , $V_{CE} = 10 \text{ Vdc}$ , $f = 1.0 \text{ kHz}$ )	$h_{fe}$	50	200	–
Current Gain – High Frequency ( $I_C = 10 \text{ mAdc}$ , $V_{CE} = 20 \text{ Vdc}$ , $f = 100 \text{ MHz}$ ) ( $I_C = 2.0 \text{ mAdc}$ , $V_{CE} = 10 \text{ V}$ , $f = 1.0 \text{ kHz}$ )	$h_{fe}$	2.5 50	– 200	–
Noise Figure ( $I_C = 100 \text{ } \mu\text{A}$ , $V_{CE} = 5.0 \text{ Vdc}$ , $R_S = 1.0 \text{ k ohm}$ , $f = 1.0 \text{ kHz}$ )	NF	–	6.0	dB

(1) Pulse Test: Pulse Width = 300  $\mu\text{s}$ . Duty Cycle = 2.0%



# Transistor Testing

**Curve Tracer** Provides a graph of the characteristic curves.

**DMM** Some DMMs measure  $\beta_{DC}$  or  $h_{FE}$ .

**Ohmmeter:**

