BIRZEIT UNIVERSITY

Faculty of Engineering and Technology
Department of Electrical and Computer Engineering

Modern Communication Systems, ENEE3306

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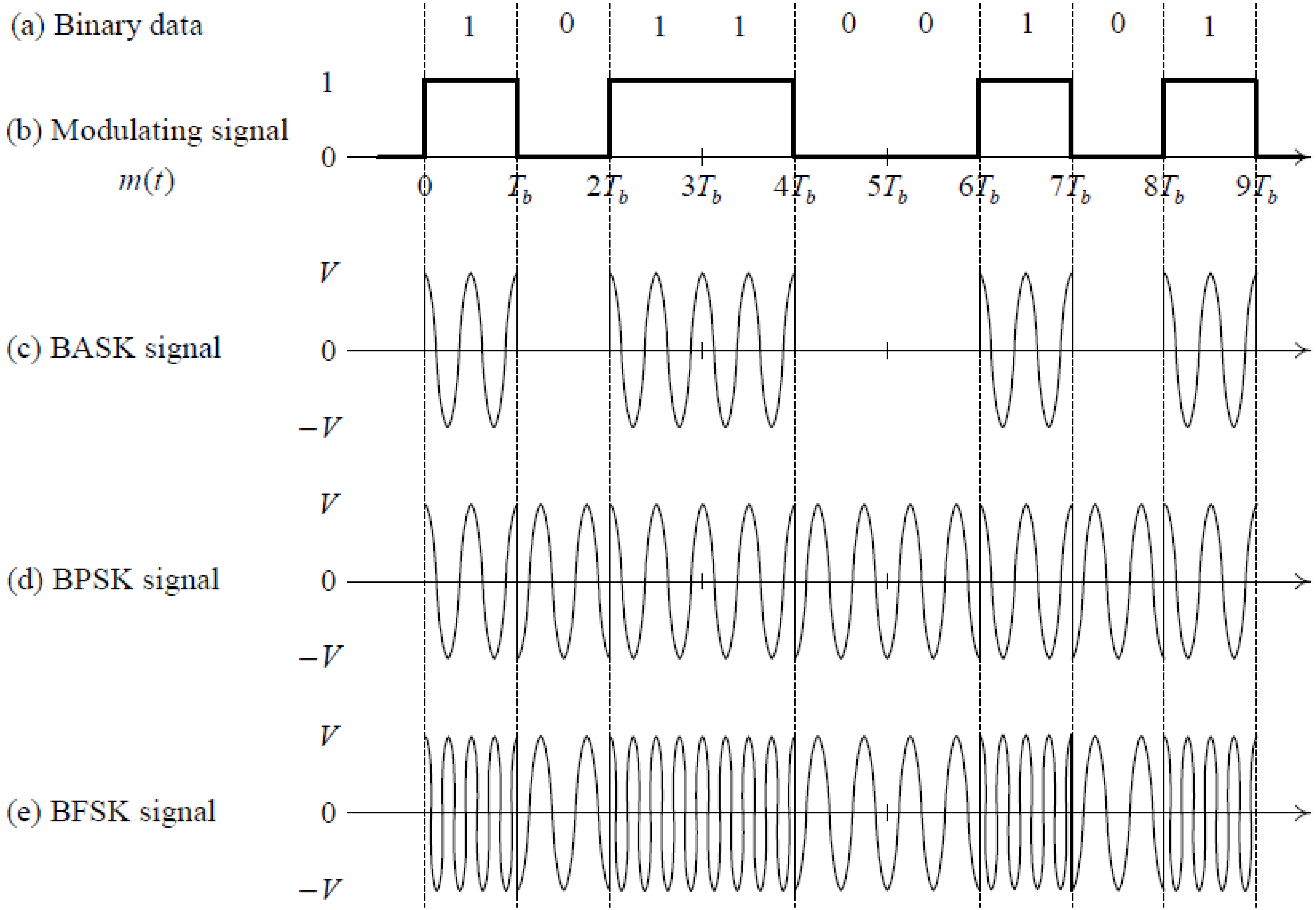
Lecture 3

- Baseband transmission is conducted at low frequencies.
- Passband transmission happens in a frequency band toward the high end of the spectrum.
- Satellite communication is in the 6–8 GHz band, while mobile phones systems are in the 800 MHz–2.0 GHz band.
- Bits are encoded as a variation of the amplitude, phase or frequency, or some combination of these parameters of a sinusoidal carrier.
- The carrier frequency is much higher than the highest frequency of the modulating signals (or messages).
- Shall consider binary amplitude-shift keying (BASK), binary phase-shift keying (BPSK) and binary frequency-shift keying (BFSK): Error performance, optimum receivers, spectra.
- Extensions to quadrature phase-shift keying (QPSK), offset QPSK (OQPSK) and minimum shift keying (MSK).

Digital Passband Modulation – Binary Passband Modulated Signals



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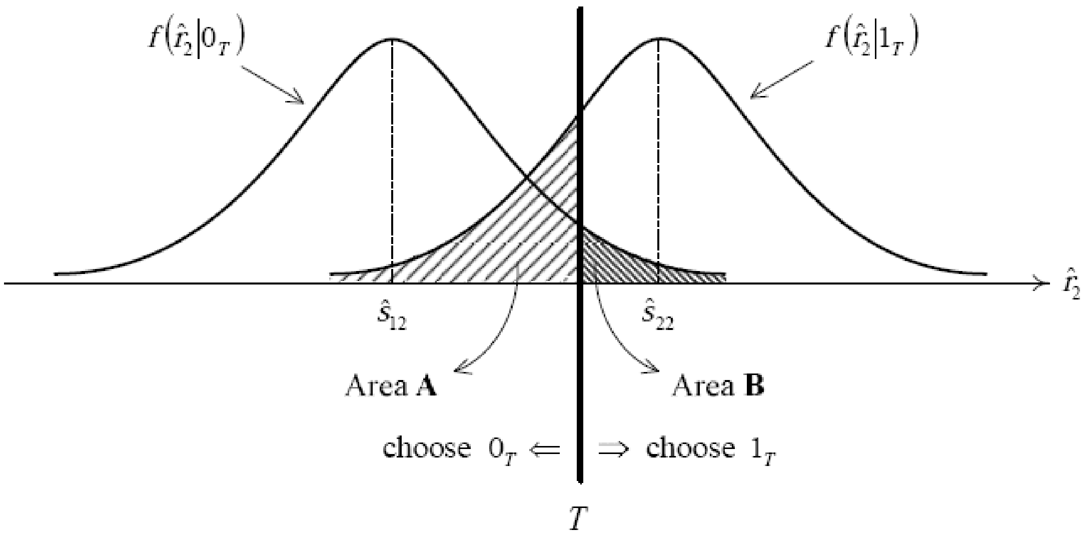
- To detect b_k, compare $\hat{r}_2$ to the threshold value T

$$\hat{r}_2 = \int_0^{T_b} r(t)\hat{\Phi}_2(t)dt$$

$$T = \frac{\hat{s}_{22} + \hat{s}_{12}}{2} + \left(\frac{N_o/2}{\hat{s}_{22} - \hat{s}_{12}}\right)\ln\left(\frac{P_1}{P_2}\right)$$

$$P[error] = P\left[\begin{array}{l}(\text{"0" decided and "1" transmitted}) \text{ or} \\ (\text{"1" decided and "0" transmitted})\end{array}\right]$$



$$P[\text{error}] = Q\left(\frac{\hat{s}_{22} - \hat{s}_{12}}{2\sqrt{N_o/2}}\right) = Q\left(\frac{\text{distance between the signals}}{2 \times \text{noise RMS value}}\right)$$

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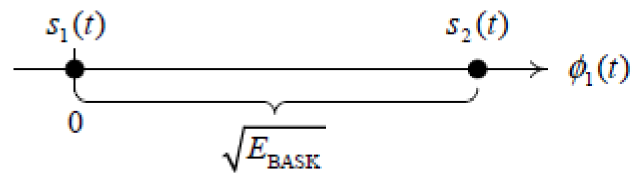
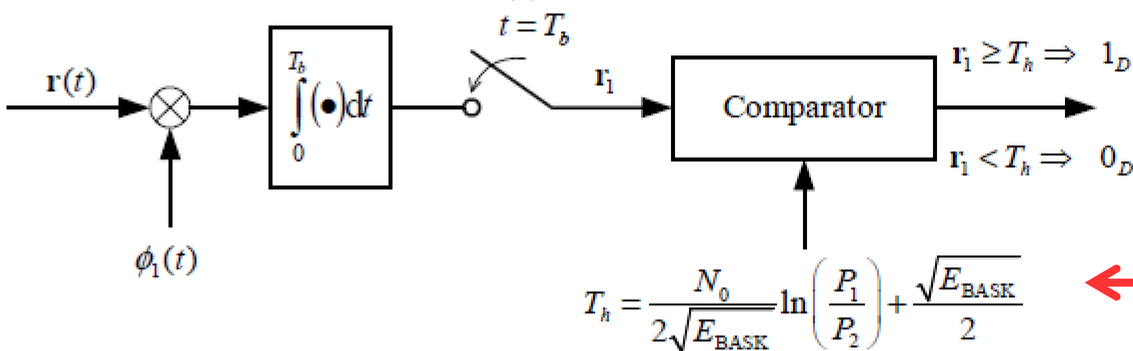
$$\begin{cases} s_1(t) = 0, & \text{“}0_T\text{”} \\ s_2(t) = V\cos(2\pi f_c t), & \text{“}1_T\text{”} \end{cases}, \quad 0 < t \le T_b,\ f_c = n/T_b$$

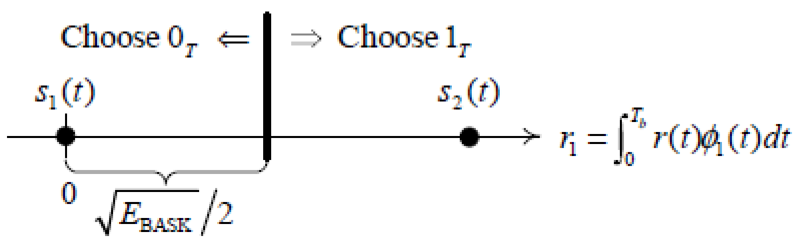


(a)



(b)

Try to derive this

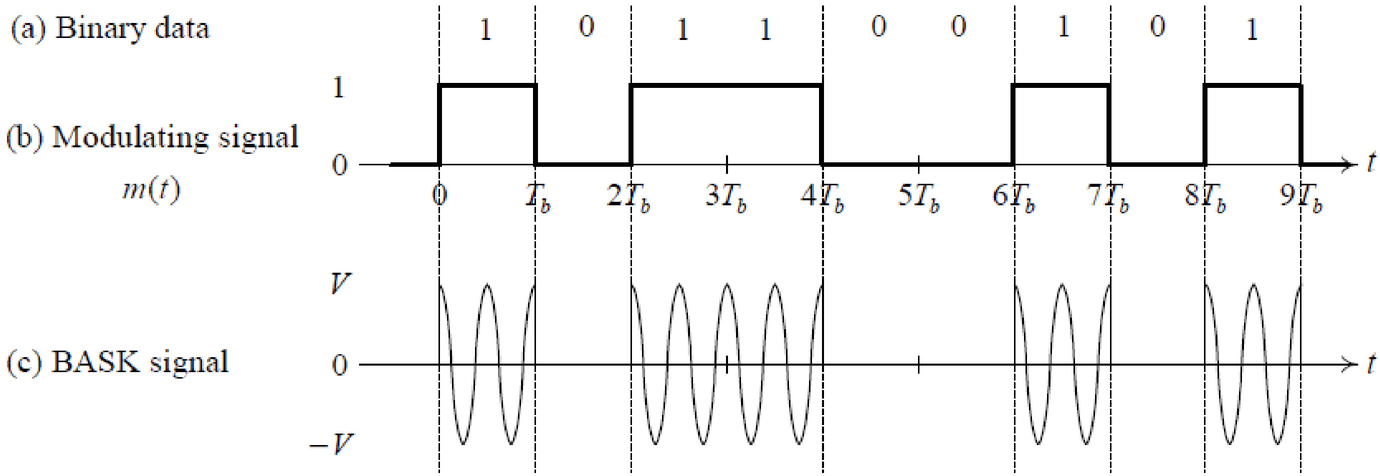


(c)

$$P[\text{error}]_{\text{BASK}} = Q\left(\sqrt{\frac{E_{\text{BASK}}}{2N_0}}\right).$$

- Notice that

$$s_{BASK}(t) = s_{NRZ}(t) \times cos(2\pi f_c t)$$



$$S_{BASK}(f) = S_{NRZ}(f) * F\left(cos^2(2\pi f_c t)\right)$$

$$= \left(\frac{V^2 T_b sinc^2(\pi f T_b)}{4} + \frac{V^2 \delta(f)}{4}\right) * \left(\frac{\delta(f - f_c) + \delta(f + f_c)}{4}\right)$$

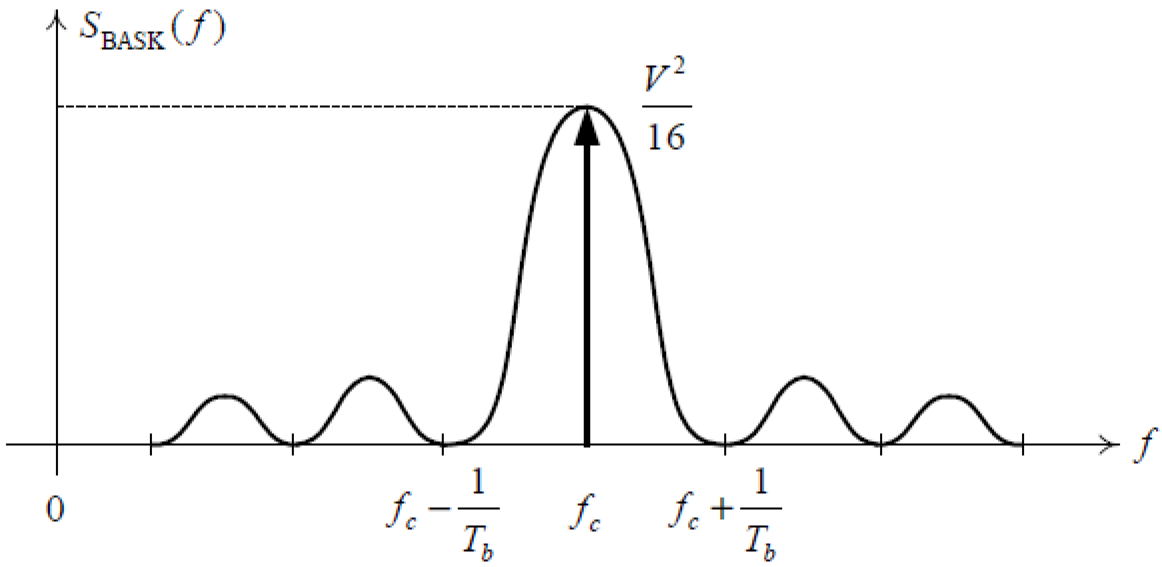
$$S_{BASK}(f) = S_{NRZ}(f) * F\left(cos(2\pi f_c t)\right)$$

$$= \left(\frac{V^2 T_b sinc^2(\pi f T_b)}{4} + \frac{V^2 \delta(f)}{4}\right) * \left(\frac{\delta(f - f_c) + \delta(f + f_c)}{4}\right)$$

$$= \left(\frac{V^2 T_b sinc^2(\pi T_b(f - f_c))}{16} + \frac{V^2 \delta(f - f_c)}{16}\right)$$
$$+ \left(\frac{V^2 T_b sinc^2(\pi T_b(f + f_c))}{16} + \frac{V^2 \delta(f + f_c)}{16}\right)$$
$$= \frac{V^2}{16}\left(\delta(f - f_c) + \delta(f + f_c) + T_b sinc^2(\pi T_b(f - f_c))\right)$$

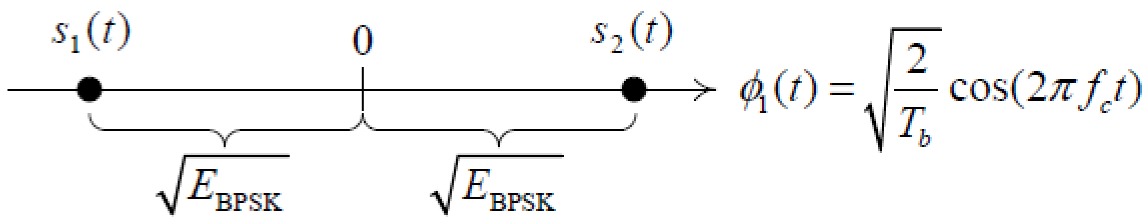
Digital Passband Modulation – PSD of BASK

$$S_{\mathsf{BASK}}(f) = \frac{V^2}{16}\left[\delta(f-f_c)+\delta(f+f_c)+\frac{\sin^2[\pi T_b(f+f_c)]}{\pi^2 T_b(f+f_c)^2}+\frac{\sin^2[\pi T_b(f-f_c)]}{\pi^2 T_b(f-f_c)^2}\right].$$



Approximately 95% of the total transmitted power lies in a band of $3/T_b$ (Hz), centered at f_c.

$$\begin{cases} s_1(t) = -V\cos(2\pi f_c t), & \text{if "}0_T\text{"} \\ s_2(t) = +V\cos(2\pi f_c t), & \text{if "}1_T\text{"} \end{cases}, \quad 0 < t \le T_b,$$

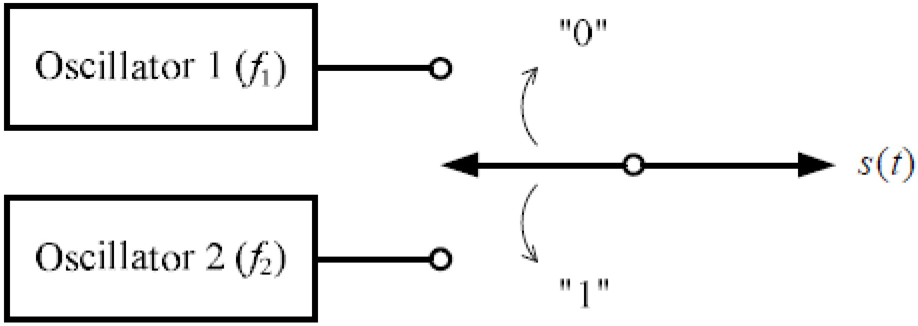


Try to derive these

$$P[\text{error}]_{\text{BPSK}} = Q\left(\sqrt{\frac{2E_{\text{BPSK}}}{N_0}}\right).$$

$$S_{\text{BPSK}}(f) = \frac{V^2}{4}\left[\frac{\sin^2[\pi(f-f_c)T_b]}{\pi^2(f-f_c)^2)T_b} + \frac{\sin^2[\pi(f+f_c)T_b]}{\pi^2(f+f_c)^2T_b}\right].$$

Similar to that of BASK, but no impulse functions at $\pm f_c$.



(a)

$$\begin{cases} s_1(t) = V\cos(2\pi f_1 t + \theta_1), & \text{if "}0_T\text{"} \\ s_2(t) = V\cos(2\pi f_2 t + \theta_2), & \text{if "}1_T\text{"} \end{cases}, \quad 0 < t \leq T_b.$$

- Minimum frequency separation for *coherent* orthogonality ($\theta_1 = \theta_2$):

Try to derive this

$$(\Delta f)_{\min}^{\text{[coherent]}} = \frac{1}{2T_b}.$$

- Minimum frequency separation for *noncoherent* orthogonality ($\theta_1 \neq \theta_2$):

$$(\Delta f)_{\min}^{\text{[noncoherent]}} = \frac{1}{T_b}.$$

(b)

$$\begin{cases} s_1(t) = V\cos 2\pi(f_c - f_d)t \\ s_2(t) = V\cos 2\pi(f_c + f_d)t \end{cases}, \quad 0 < t \le T_b,$$

- where f_c is the carrier frequency and f_d is the *frequency deviation*

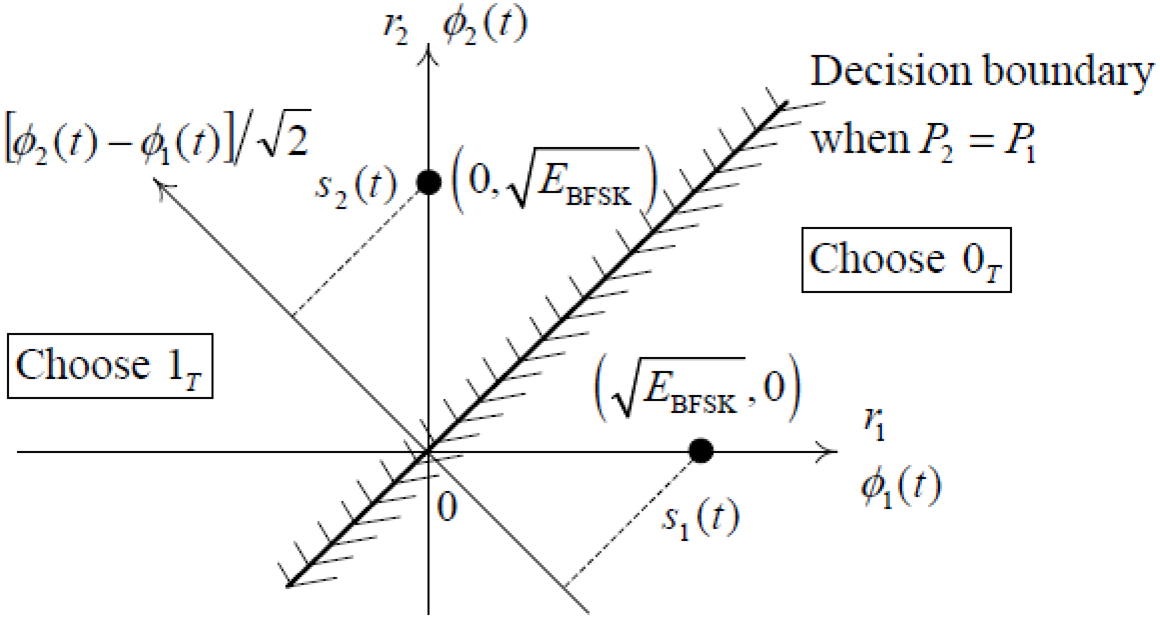
$$f_c = n/4T_b,$$

$$f_d = \begin{cases} m/4T_b & \text{(coherent orthogonality)} \\ m/2T_b & \text{(noncoherent orthogonality)} \end{cases},$$

Try to derive this

- Where $n \gg m$

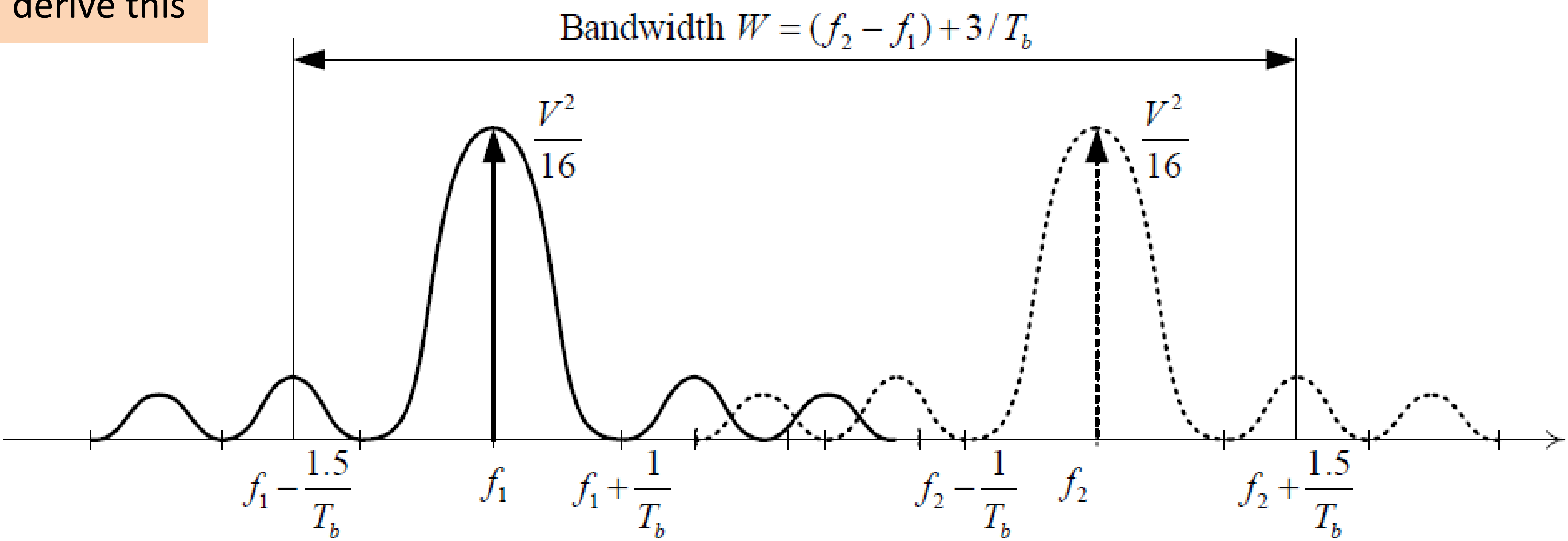
$$\phi_1(t) = \frac{s_1(t)}{\sqrt{E_{\mathsf{BFSK}}}}, \quad \phi_2(t) = \frac{s_2(t)}{\sqrt{E_{\mathsf{BFSK}}}}.$$



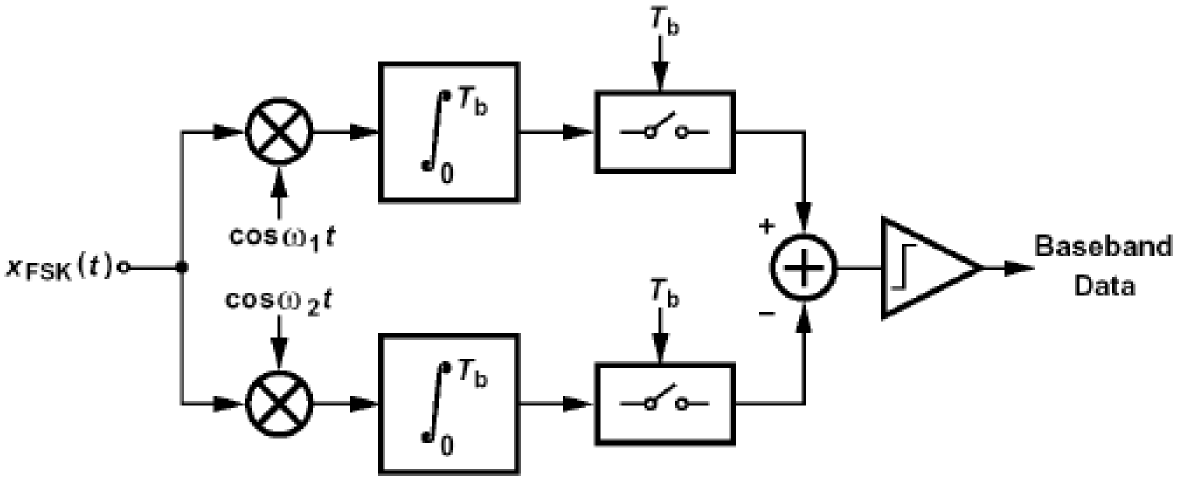
$$P[\mathsf{error}]_{\mathsf{BFSK}} = Q\left(\sqrt{\frac{E_{\mathsf{BFSK}}}{N_0}}\right).$$

$$S_{\mathrm{BFSK}}(f) = \frac{V^2}{16}\left[\delta(f-f_2)+\delta(f+f_2)+\frac{\sin^2[\pi T_b(f+f_2)]}{\pi^2 T_b(f+f_2)^2}+\frac{\sin^2[\pi T_b(f-f_2)]}{\pi^2 T_b(f-f_2)^2}\right]$$
$$+ \frac{V^2}{16}\left[\delta(f-f_1)+\delta(f+f_1)+\frac{\sin^2[\pi T_b(f+f_1)]}{\pi^2 T_b(f+f_1)^2}+\frac{\sin^2[\pi T_b(f-f_1)]}{\pi^2 T_b(f-f_1)^2}\right].$$

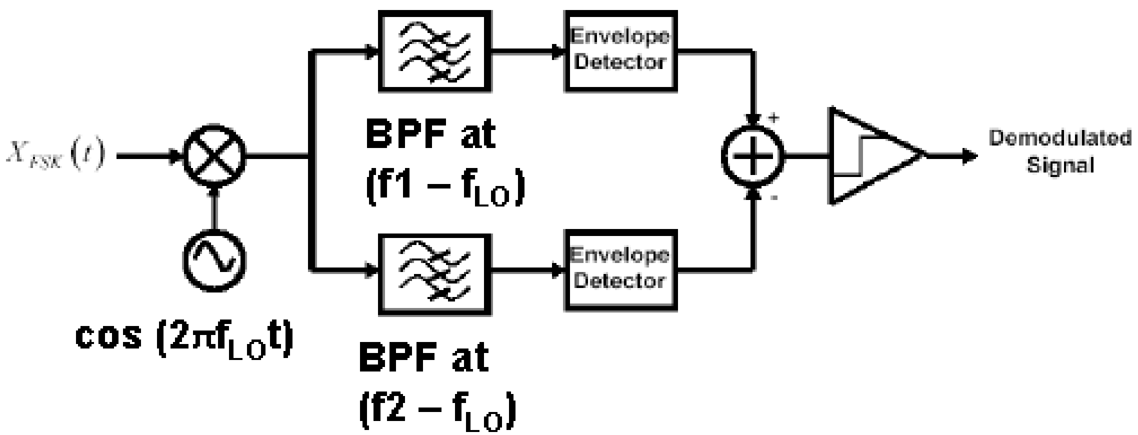
Try to derive this

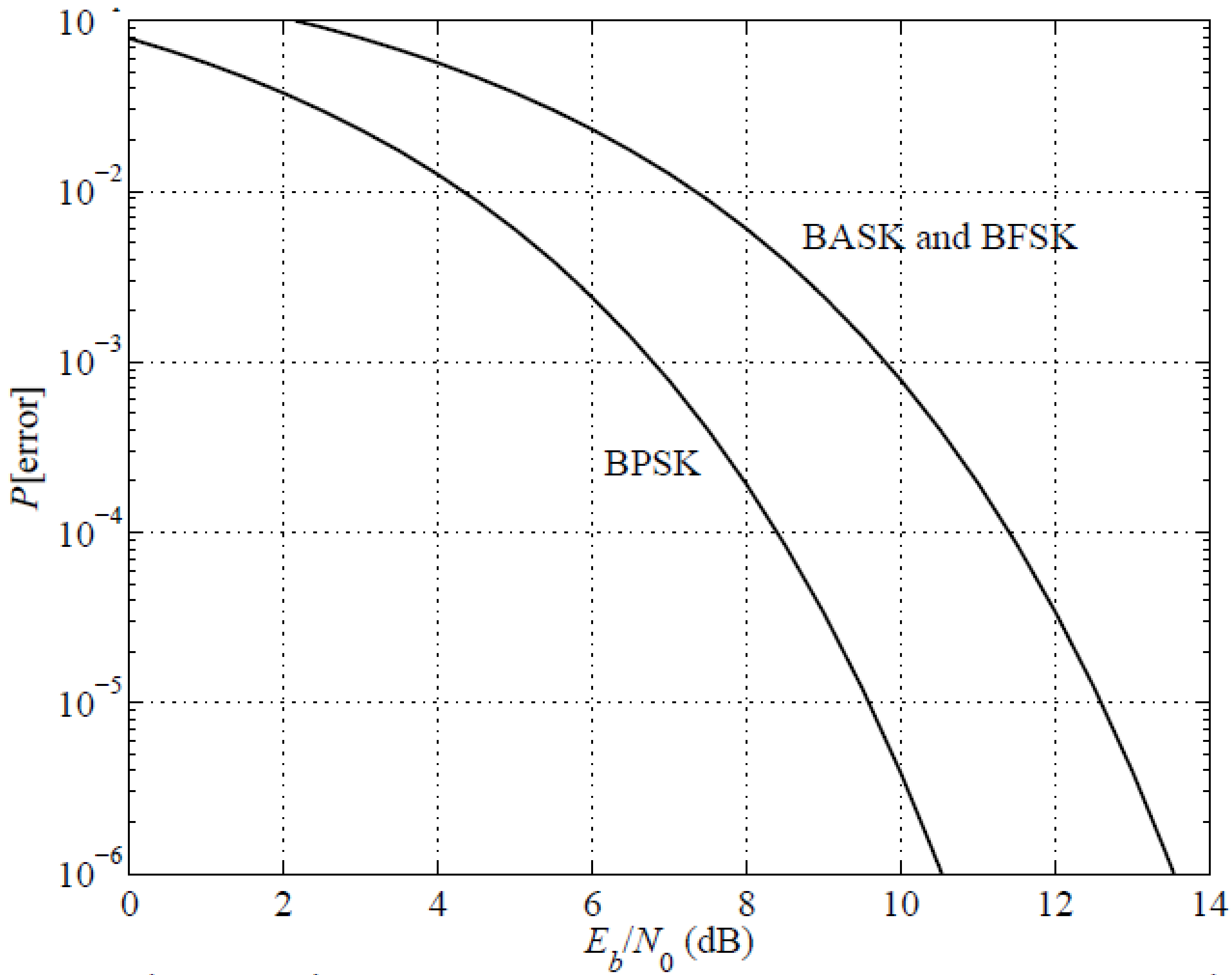


FSK Demodulation (Coherent)



FSK Demodulation (Non-Coherent)

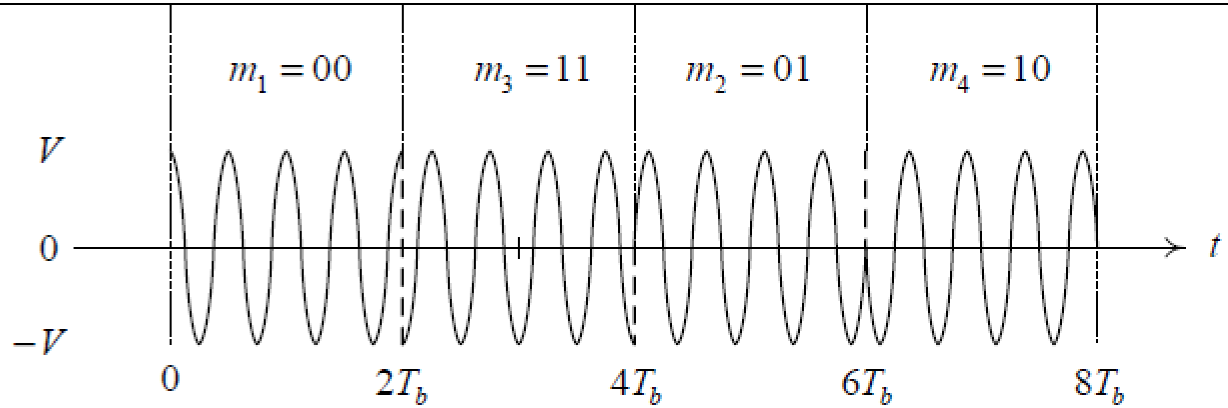




$$P[\text{error}]_{\text{BPSK}} = Q\left(\sqrt{\frac{2E_b}{N_0}}\right),\ P[\text{error}]_{\text{BASK}} = P[\text{error}]_{\text{BFSK}} = Q\left(\sqrt{\frac{E_b}{N_0}}\right).$$

- Basic idea behind QPSK: $\cos(2\pi f_c t)$ and $\sin(2\pi f_c t)$ are orthogonal over $[0, T_b]$ when $f_c = k/T_b$, k integer $\Rightarrow$ Can transmit two different bits over the same frequency band at the same time.
- The symbol signaling rate (i.e., the *baud rate*) is $r_s = 1/T_s = 1/(2T_b) = r_b/2$ (symbols/sec), i.e., *halved*.

Bit Pattern	Message	Signal Transmitted	
00	m_1	$s_1(t) = V\cos(2\pi f_c t)$,	$0 \le t \le T_s = 2T_b$
01	m_2	$s_2(t) = V\sin(2\pi f_c t)$,	$0 \le t \le T_s = 2T_b$
11	m_3	$s_3(t) = -V\cos(2\pi f_c t)$,	$0 \le t \le T_s = 2T_b$
10	m_4	$s_4(t) = -V\sin(2\pi f_c t)$,	$0 \le t \le T_s = 2T_b$



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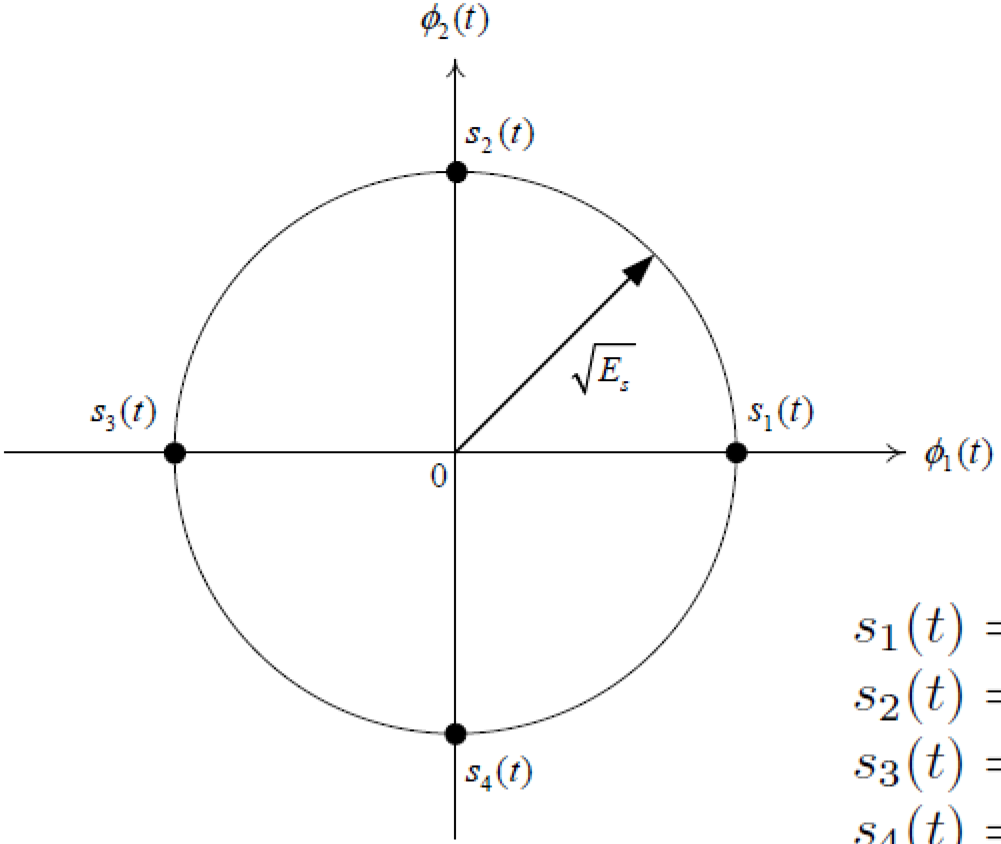
$$\int_0^{T_s} s_i^2(t)\mathrm{d}t = \frac{V^2}{2}T_s = V^2 T_b = E_s,$$

$$\phi_1(t) = \frac{s_1(t)}{\sqrt{E_s}}, \quad \phi_2(t) = \frac{s_2(t)}{\sqrt{E_s}}.$$



$$\begin{aligned} s_1(t) &= V\cos(2\pi f_c t), \\ s_2(t) &= V\sin(2\pi f_c t), \\ s_3(t) &= -V\cos(2\pi f_c t), \\ s_4(t) &= -V\sin(2\pi f_c t), \end{aligned}$$

$$P[\text{error}] = 1 - P[\text{correct}]$$

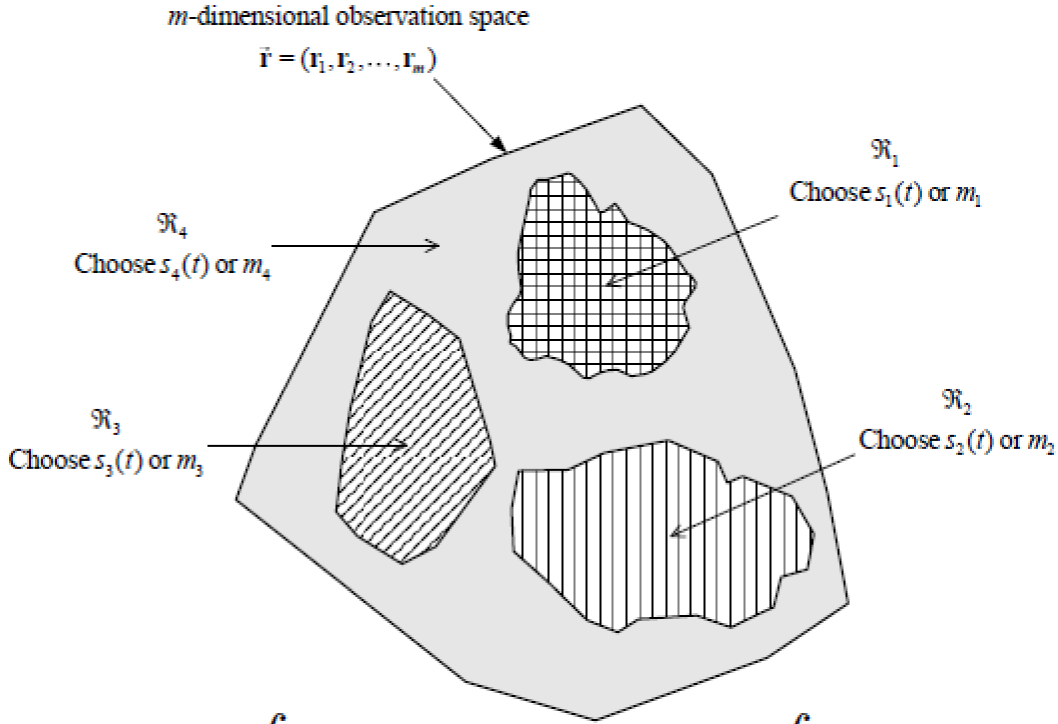
Minimize Error or Maximize correct decision

$$P[\text{correct}] = P\begin{bmatrix} (S_1 \text{ decided and } S_1 \text{ transmitted}) \text{ or} \\ (S_2 \text{ decided and } S_2 \text{ transmitted}) \text{ or} \\ (S_3 \text{ decided and } S_3 \text{ transmitted}) \text{ or} \\ (S_4 \text{ decided and } S_4 \text{ transmitted}) \text{ or} \end{bmatrix}$$

$$= P[S_{1,D}, S_{1,T}] + P[S_{2,D}, S_{2,T}] + P[S_{3,D}, S_{3,T}] + P[S_{4,D}, S_{4,T}]$$

$$= P[S_{1,D} / S_{1,T}]P[S_{1,T}] + P[S_{2,D} / S_{2,T}]P[S_{2,T}] + P[S_{3,D} / S_{3,T}]P[S_{3,T}] + P[S_{4,D} / S_{4,T}]P[S_{4,T}]$$

$$= P_1 \int_{R_1} f(\vec{r} / S_{1,T}) d\vec{r} + P_2 \int_{R_2} f(\vec{r} / S_{2,T}) d\vec{r} + P_3 \int_{R_3} f(\vec{r} / S_{3,T}) d\vec{r} + P_4 \int_{R_4} f(\vec{r} / S_{4,T}) d\vec{r}$$



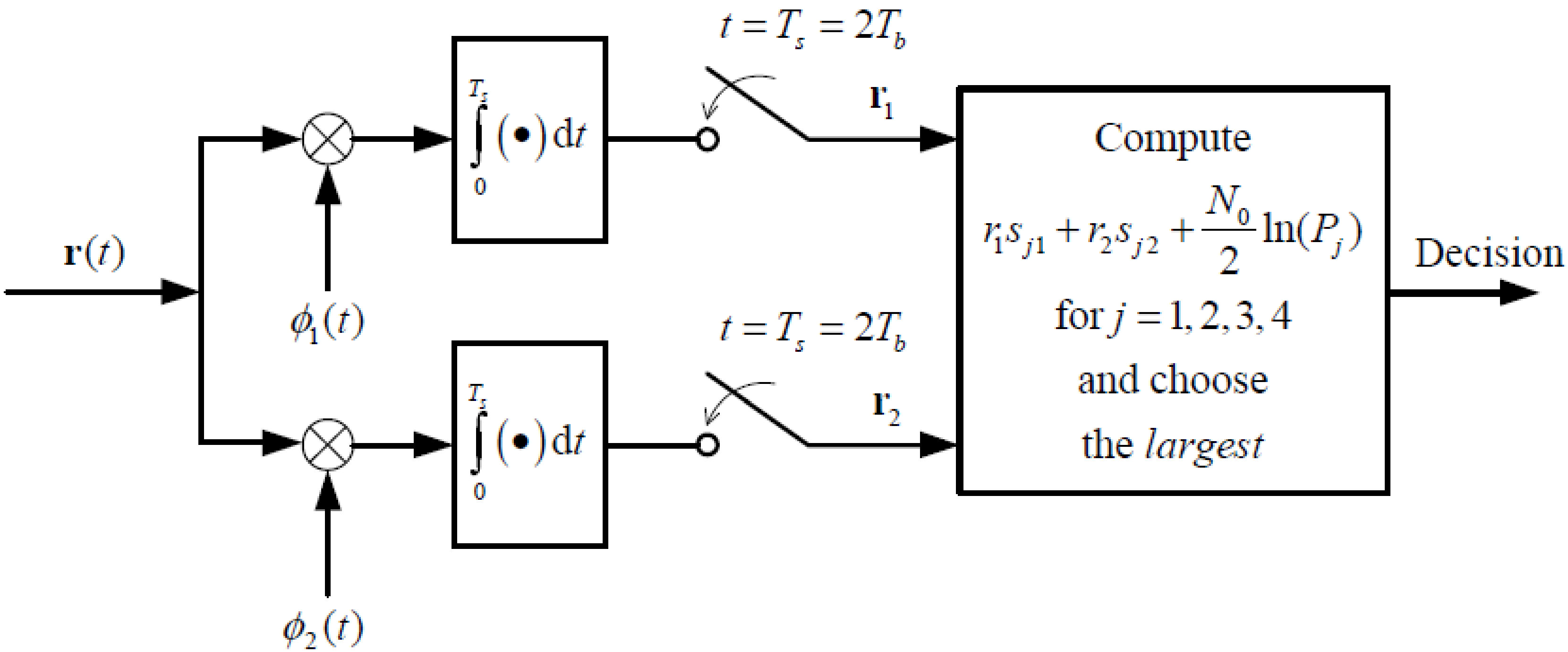
$$
\begin{aligned}
P[\text{correct}] &= \int_{\Re_1} P_1 f(\vec{r}|s_1(t))\mathrm{d}\vec{r} + \int_{\Re_2} P_2 f(\vec{r}|s_2(t))\mathrm{d}\vec{r} \\
&+ \int_{\Re_3} P_3 f(\vec{r}|s_3(t))\mathrm{d}\vec{r} + \int_{\Re_4} P_4 f(\vec{r}|s_4(t))\mathrm{d}\vec{r}.
\end{aligned}
$$

Choose $s_i(t)$ if $P_i f(\vec{r}|s_i(t)) > P_j f(\vec{r}|s_j(t)),\ j = 1, 2, 3, 4;\ j \neq i.$

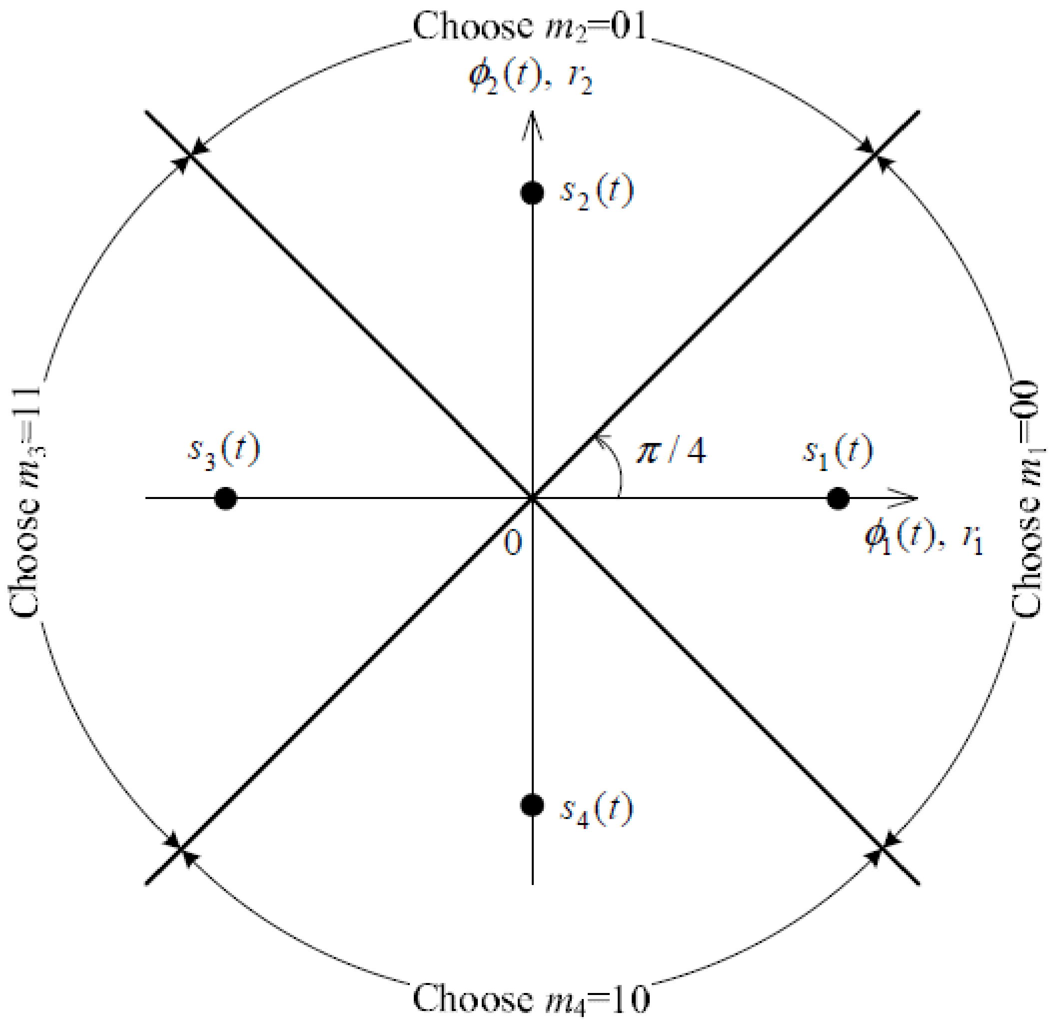
Choose $s_i(t)$ if

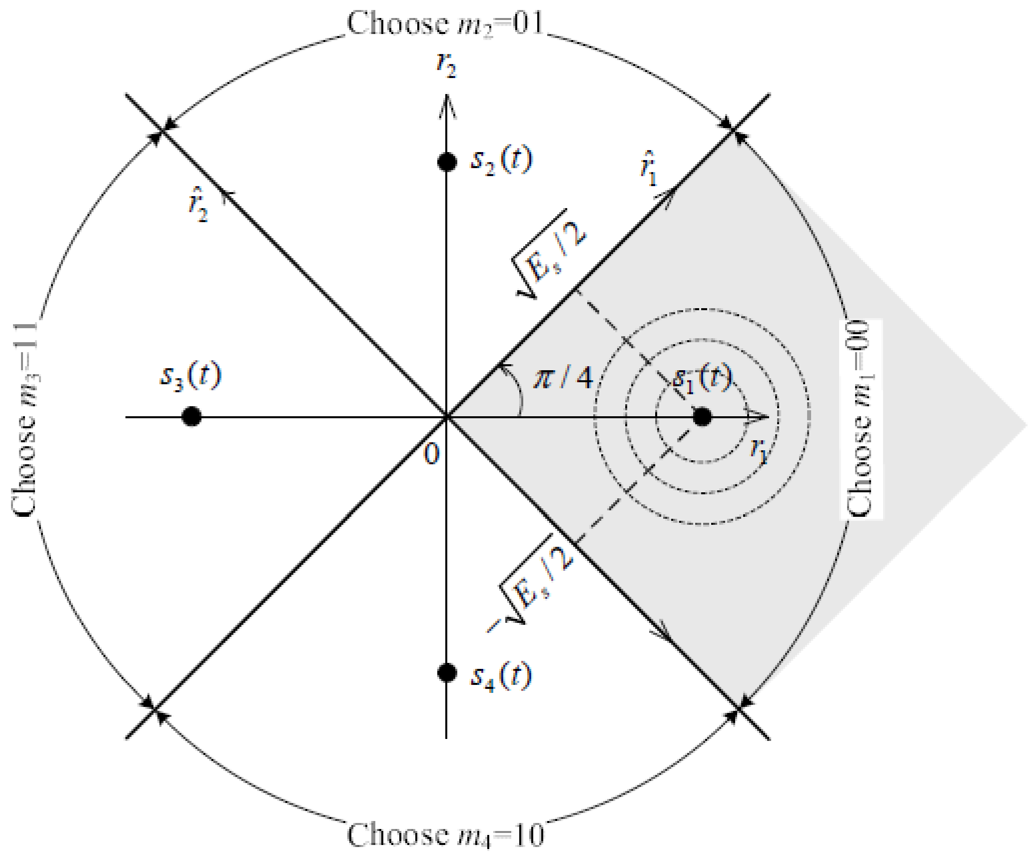
$$\frac{N_0}{2}\ln P_i + r_1 s_{i1} + r_2 s_{i2} > \frac{N_0}{2}\ln P_j + r_1 s_{j1} + r_2 s_{j2}$$

$$j = 1, 2, 3, 4;\; j \neq i.$$

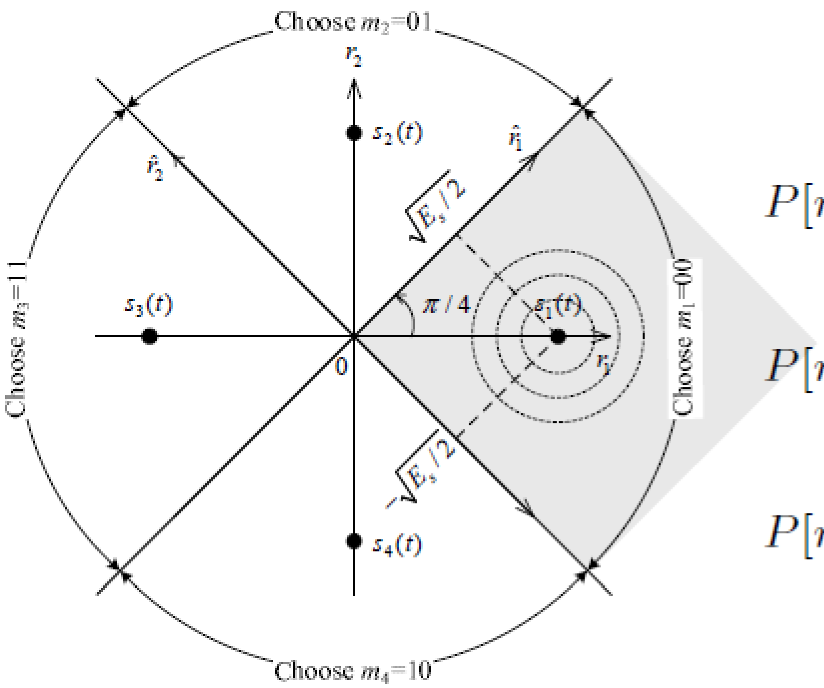


Choose $s_i(t)$ if $(r_1 - s_{i1})^2 + (r_2 - s_{i2})^2$ is *the smallest*





$$P[\text{error}] = P[\text{error}|s_i(t)] = 1 - P[\text{correct}|s_i(t)] = 1 - \left[1 - Q\left(\sqrt{\frac{E_s}{N_0}}\right)\right]^2$$

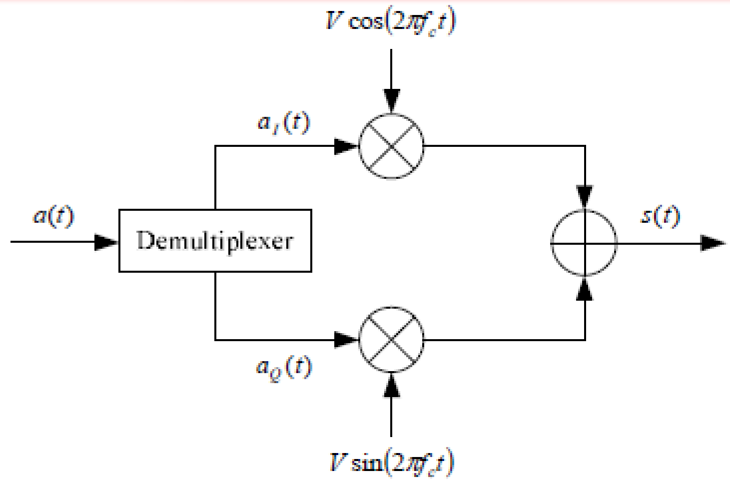


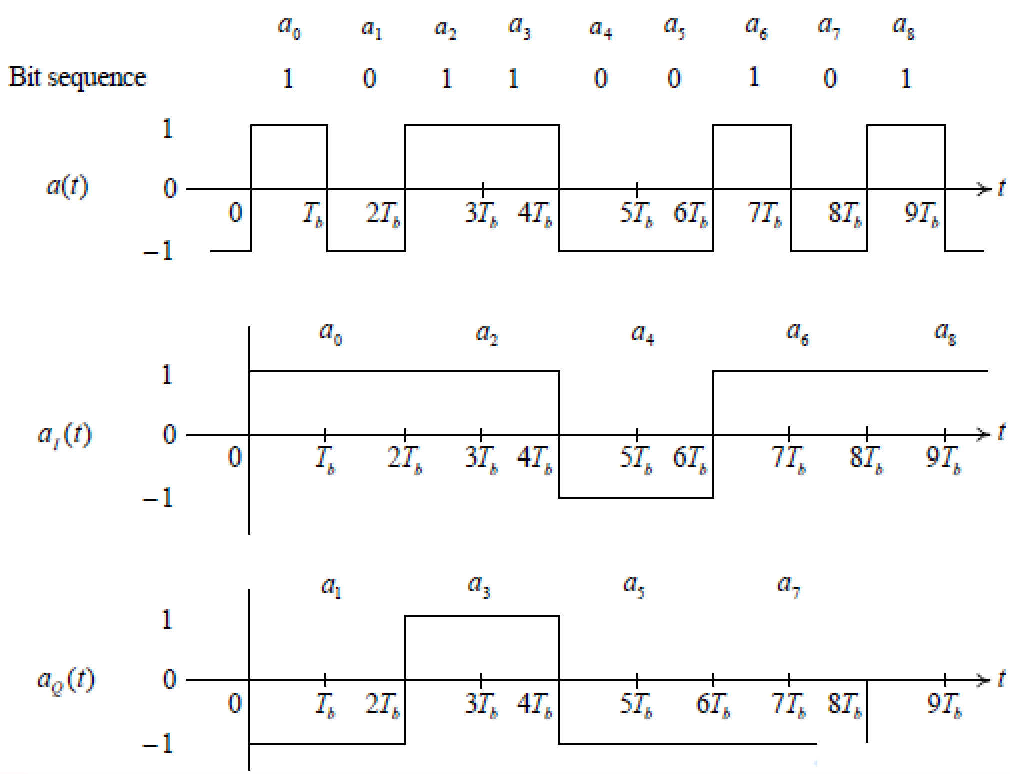
$$
\begin{aligned}
P[m_2|m_1] &= Q\left(\sqrt{\frac{E_s}{N_0}}\right)\left[1 - Q\left(\sqrt{\frac{E_s}{N_0}}\right)\right], \\
P[m_3|m_1] &= Q^2\left(\sqrt{\frac{E_s}{N_0}}\right), \\
P[m_4|m_1] &= Q\left(\sqrt{\frac{E_s}{N_0}}\right)\left[1 - Q\left(\sqrt{\frac{E_s}{N_0}}\right)\right].
\end{aligned}
$$

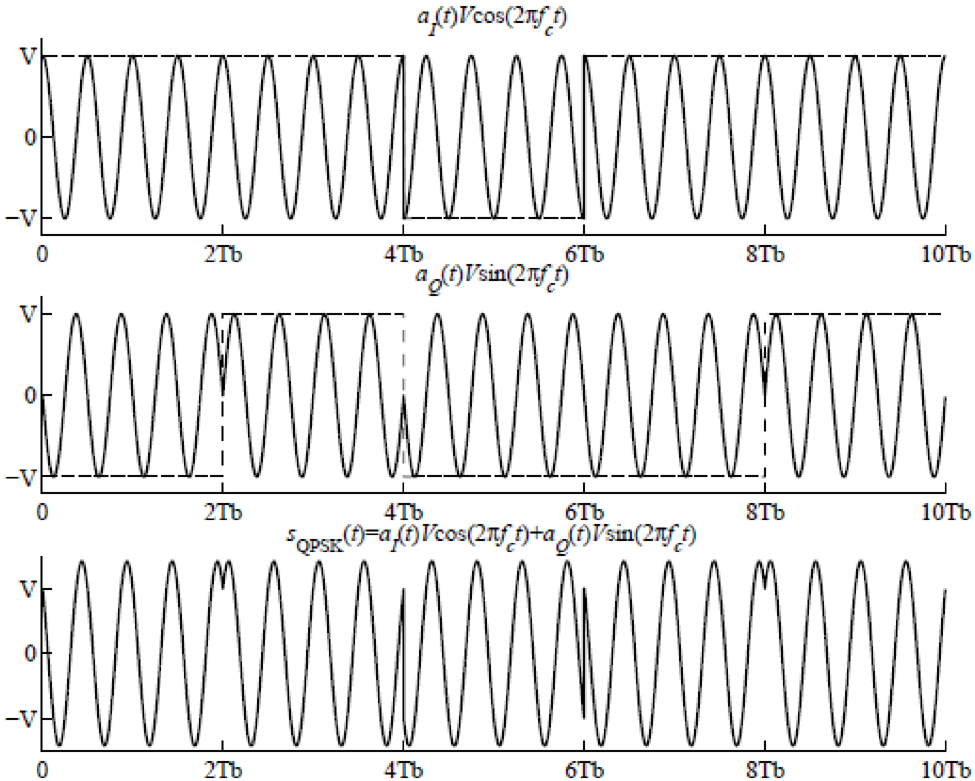
$$
\begin{aligned}
P[\text{bit error}] &= 0.5P[m_2|m_1] + 0.5P[m_4|m_1] + 1.0P[m_3|m_1] \\
&= Q\left(\sqrt{\frac{E_s}{N_0}}\right).
\end{aligned}
$$

Gray mapping: *Nearest neighbors* are mapped to the bit pairs that differ in only one bit.

Digital Passband Modulation – An Alternative Representation of QPSK





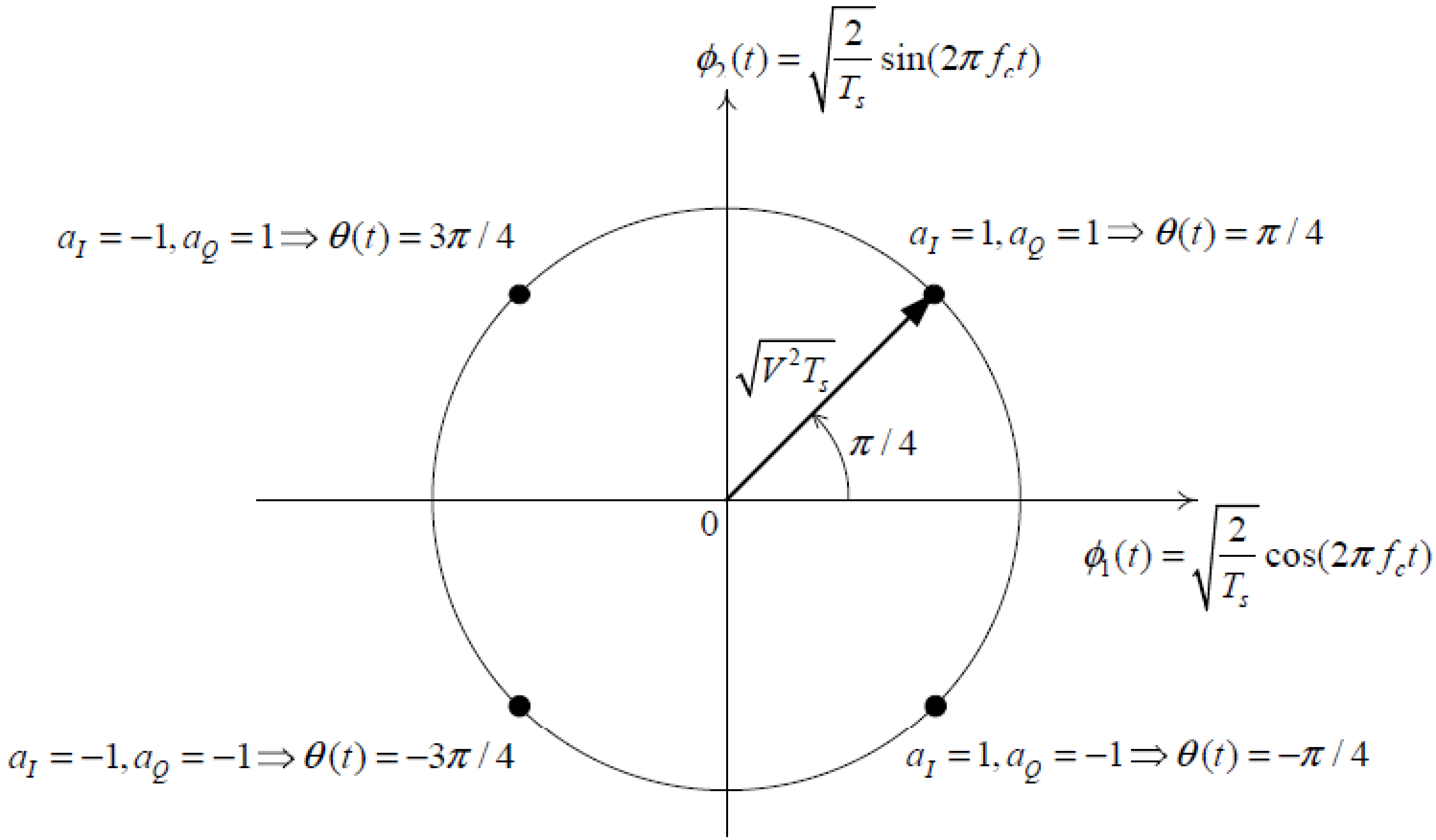


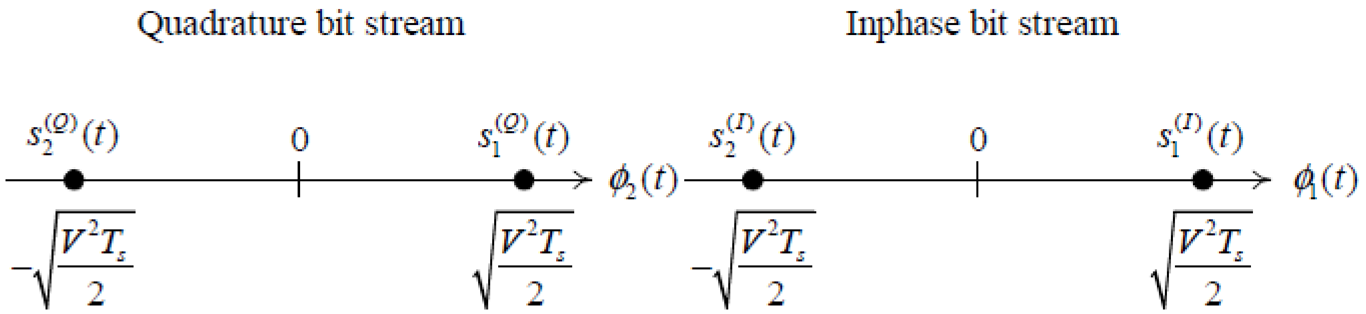
$$
\begin{aligned}
s(t) &= a_I(t)V\cos(2\pi f_c t) + a_Q(t)V\sin(2\pi f_c t) \\
&= \sqrt{a_I^2(t)+a_Q^2(t)}V\cos\left(2\pi f_c t - \tan^{-1}\left(\frac{a_Q(t)}{a_I(t)}\right)\right) = \sqrt{2}V\cos[2\pi f_c t - \theta(t)].
\end{aligned}
$$

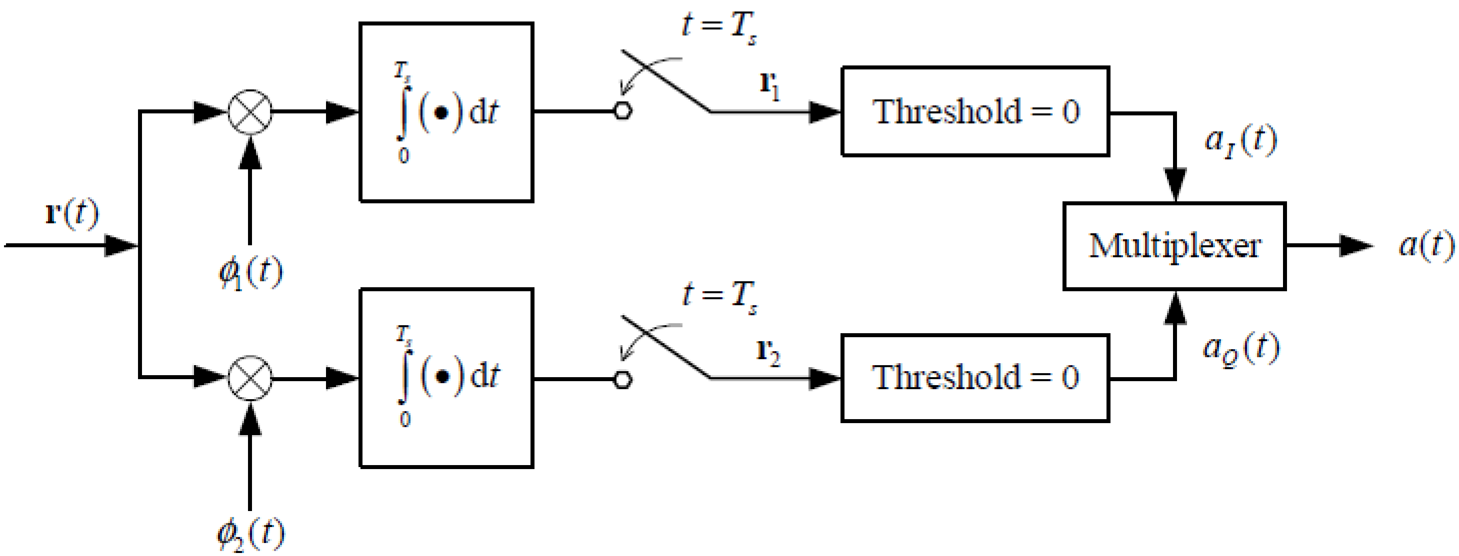
$$
\theta(t) = \begin{cases} \pi/4, & \text{if } a_I = +1, a_Q = +1 \text{ (bits are 11)} \\ -\pi/4, & \text{if } a_I = +1, a_Q = -1 \text{ (bits are 10)} \\ 3\pi/4, & \text{if } a_I = -1, a_Q = +1 \text{ (bits are 01)} \\ -3\pi/4, & \text{if } a_I = -1, a_Q = -1 \text{ (bits are 00)} \end{cases}.
$$

Digital Passband Modulation – Signal Space Representation of QPSK

$$\begin{cases} \phi_1(t) = \dfrac{V\cos(2\pi f_c t)}{\sqrt{V^2 T_b}} \\ \phi_2(t) = \dfrac{V\sin(2\pi f_c t)}{\sqrt{V^2 T_b}} \end{cases}, \quad 0 < t < T_s = 2T_b,$$

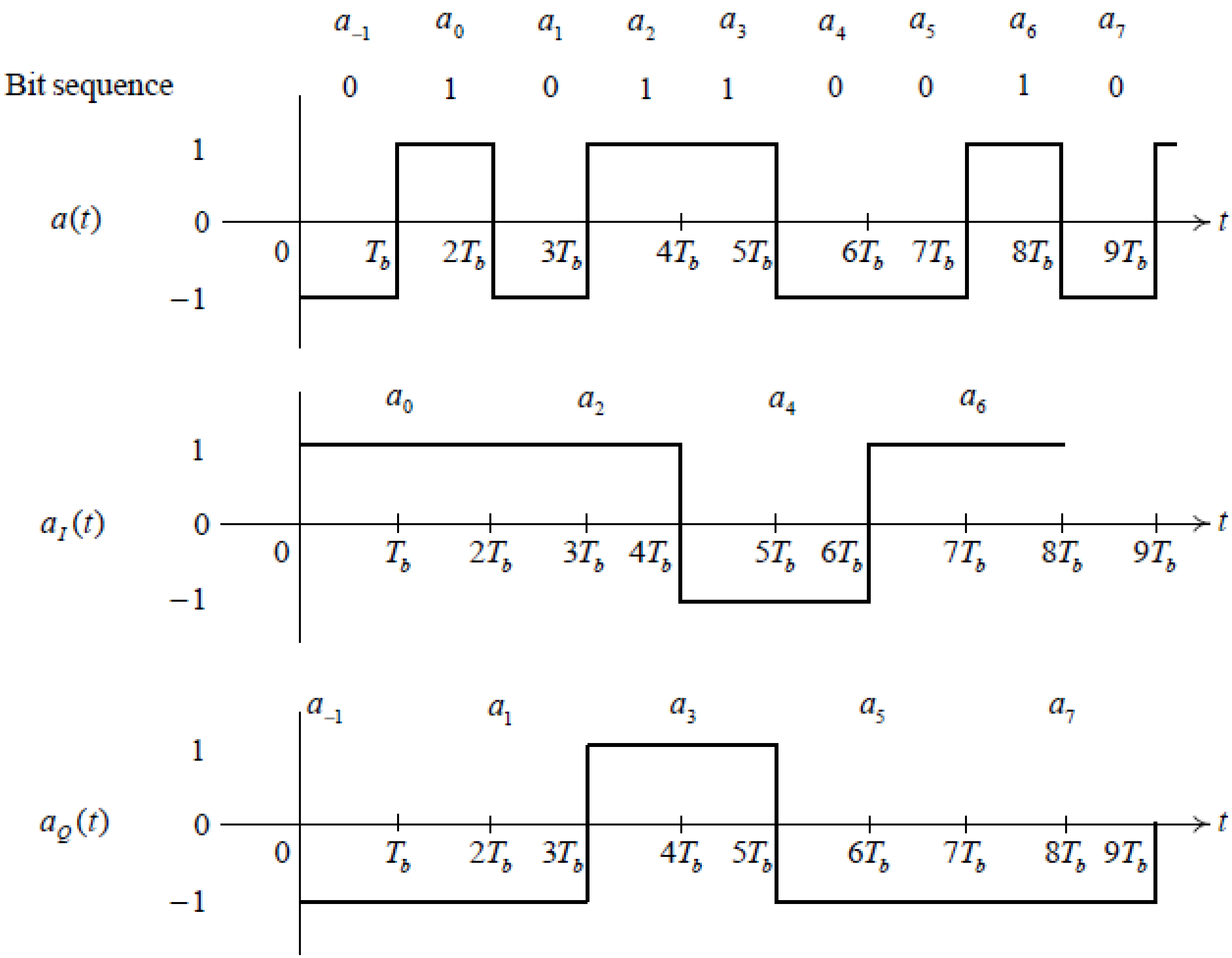




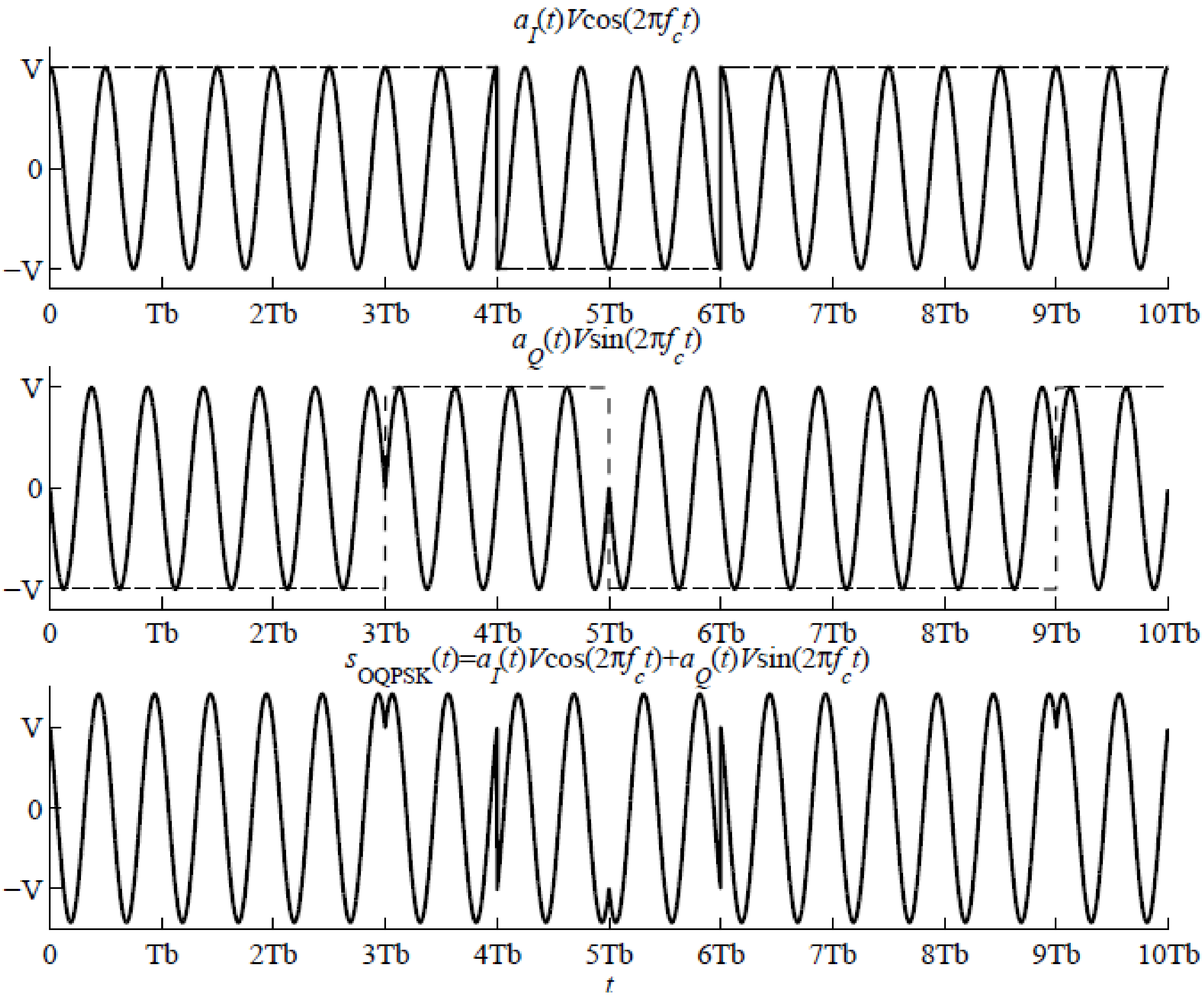


$$P[\text{bit error}] = Q\left(\sqrt{\frac{V^2T_s}{N_0}}\right) = \cdots = Q\left(\sqrt{\frac{2E_b}{N_0}}\right).$$

In OQPSK the $a_I(t)$ and $a_Q(t)$ bit streams are offset by one bit interval T_b.

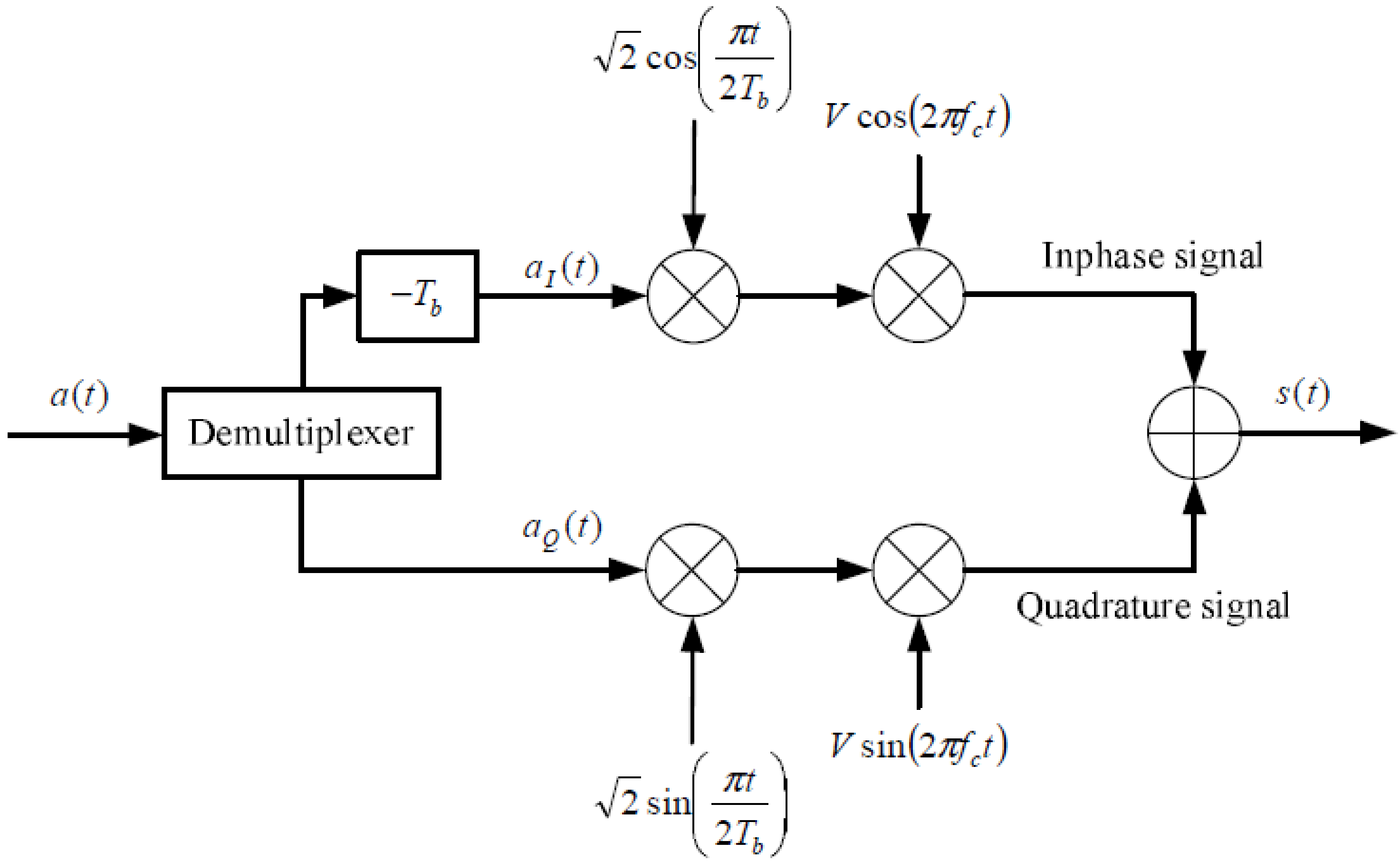


Digital Passband Modulation – Offset Quadrature Phase Shift Keying (OQPSK)

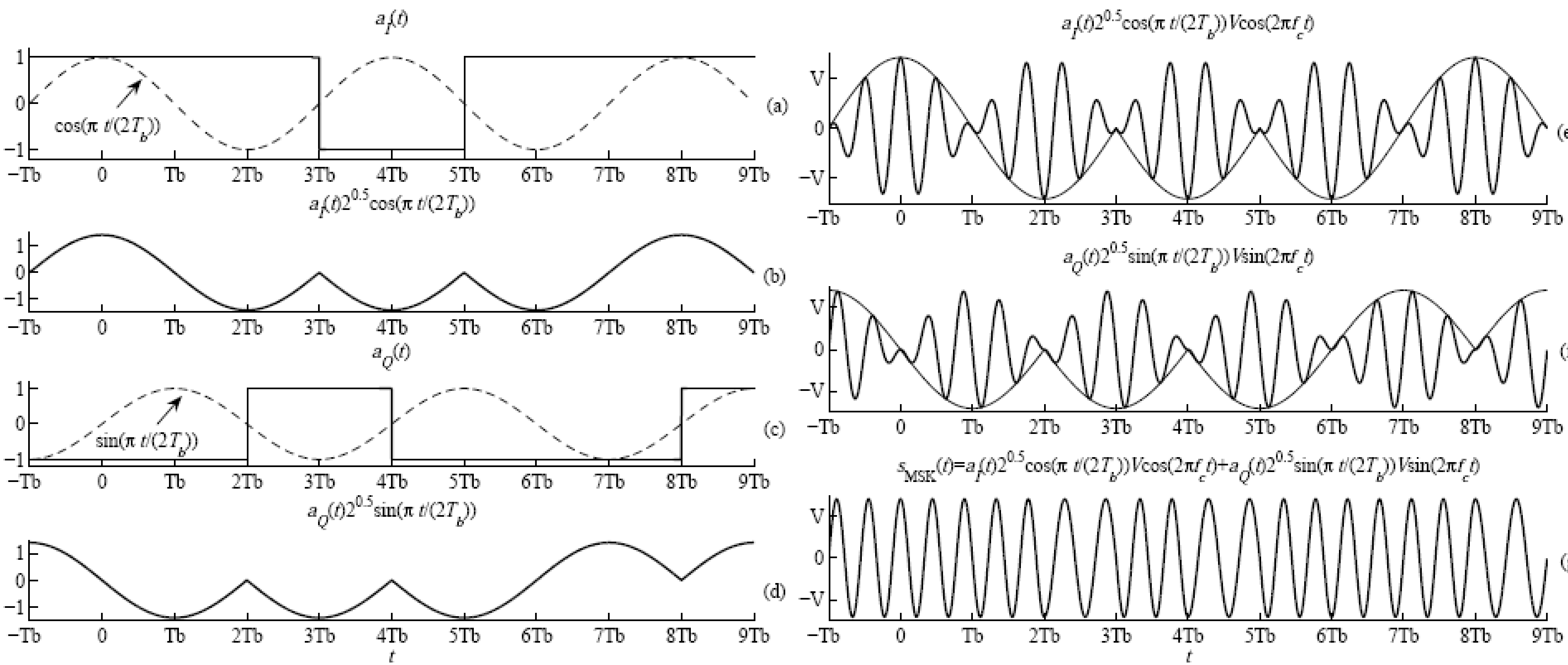


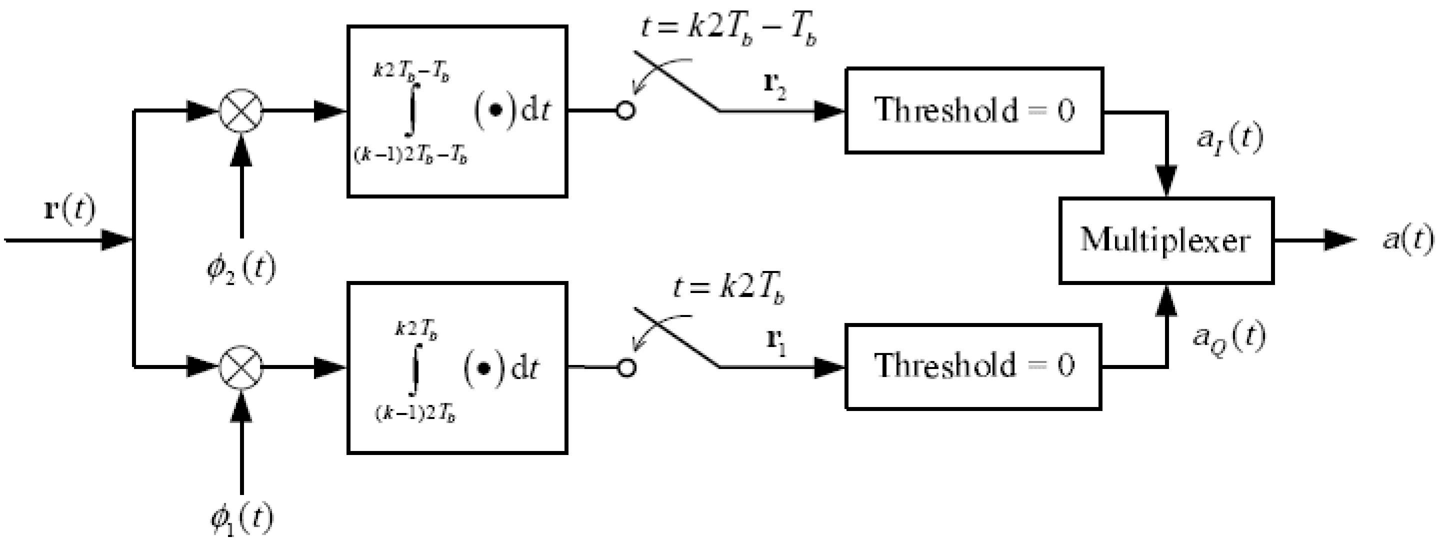
Both QPSK and OQPSK signals have sudden jumps.

MSK eliminates the jumps altogether by applying weighting functions to the carriers $V\cos(2\pi f_c t)$ and $V\sin(2\pi f_c t)$.



Digital Passband Modulation – Generation of an MSK Signal





$$\phi_1(t) = \left[\sqrt{2}\sin\left(\frac{\pi t}{2T_b}\right) V\sin(2\pi f_c t)\right] \Big/ \sqrt{V^2 T_b},$$

$$\phi_2(t) = \left[\sqrt{2}\cos\left(\frac{\pi t}{2T_b}\right) V\cos(2\pi f_c t)\right] \Big/ \sqrt{V^2 T_b}.$$

$$P[\text{bit error}] = Q\left(\sqrt{\frac{2E_b}{N_0}}\right), \quad E_b = V^2 T_b \text{ is the energy per bit.}$$

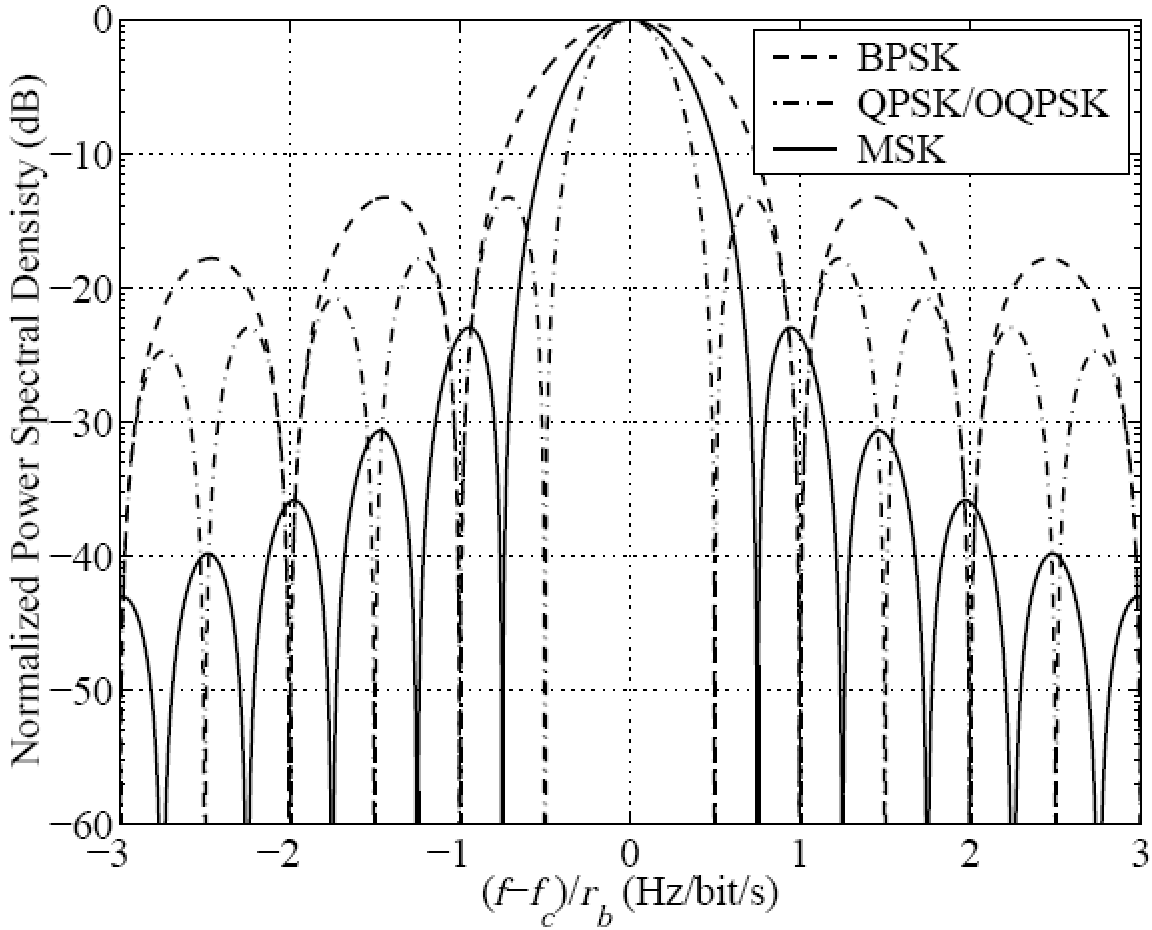
$$\begin{aligned} s(t) &= a_I(t)\sqrt{2}\cos\left(\frac{\pi t}{2T_b}\right)V\cos(2\pi f_c t) + a_Q(t)\sqrt{2}\sin\left(\frac{\pi t}{2T_b}\right)V\sin(2\pi f_c t) \\ &= A\cos(2\pi f_c t - \theta). \end{aligned}$$

$$A = \left[a_I^2(t)2V^2\cos^2\left(\frac{\pi t}{2T_b}\right) + a_Q^2(t)2V^2\sin^2\left(\frac{\pi t}{2T_b}\right)\right]^{\frac{1}{2}} = \sqrt{2}V$$

$$\theta = \tan^{-1}\left\{\frac{a_Q(t)\sin\left(\frac{\pi t}{2T_b}\right)}{a_I(t)\cos\left(\frac{\pi t}{2T_b}\right)}\right\} = \tan^{-1}\left\{\pm\tan\left(\frac{\pi t}{2T_b}\right)\right\} = \pm\frac{\pi t}{2T_b}.$$

$$\Rightarrow s(t) = \sqrt{2}V\cos\left[2\pi\left(f_c \pm \frac{1}{4T_b}\right)t\right].$$

An MSK signal is of either frequency $f_2 = f_c + \frac{1}{4T_b}$ or $f_1 = f_c - \frac{1}{4T_b} \Rightarrow$ can be viewed as *frequency-shift keying* signal with continuous phase.



$$S_{\mathsf{MSK}}(f) = K\left\{\left[\frac{\cos[2\pi(f-f_c)T_b]}{4\pi^2(f-f_c)^2-\pi^2/(4T_b^2)}\right]^2+\left[\frac{\cos[2\pi(f+f_c)T_b]}{4\pi^2(f+f_c)^2-\pi^2/(4T_b^2)}\right]^2\right\}.$$

	GSM/DCS-1800	IS-54/136	PDC	1S-95
Region	Europe	North America	Japan	North America
Frequency band (MHz)	900/1800/1900	800/1900	700/1500	800/1900
Multiple access	F/TDMA	F/TDMA	F/TDMA	F/CDMA
Carrier spacing (kHz)	200	30	25	1250
Modulation	GMSK	OQPSK	OQPSK	BPSK/QPSK
Speech coding (kb/s)	VSELP (HR-5.6) RPE-LTP (FR-13) ACELP (EFR-12.2)	VSELP (FR-7.95) ACELP (EFR-7.4)	PSI-CELP (HR-3.45) VSELP (FR-6.7)	QCELP (8, 4, 2, 1) RCELP (EVRC)
Frame size (ms)	4.6	40	20	20
Channel coding (convolution code)	Rate 1/2	Rate 1/2	Rate 1/2	Rate 1/2 or 1/3

HR: half-rate codec; FR: full-rate codec; EFR: enhanced full-rate codec; EVRC: enhanced variable rate codec
An adaptive multitate (AMR) codec for GSM is currently being standardized by ETSI

Table 1. *Air interface characteristics of 2G systems.*

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Proposal	UTRA	cdma2000	WCDMA/NA	WIMS W-CDMA	W-CDMA	TD-SCDMA	CDMA II	CDMA I
Multiple-access	FDD: DS-CDMA TDD: T/CDMA	FDD: DS-CDMA TDD: T/CDMA	FDD: DS-CDMA TDD: T/CDMA	FDD: DS-CDMA TDD: DS-W-CDMA(FL) DS-S-TDMA (RL)	FDD: DS-CDMA TDD: T/CDMA	TDMA/CDMA	DS-CDMA	DS-CDMA
Duplex scheme	FDD/TDD	FDD/TDD	FDD/TDD	W-CDMA FDD mode: FDD S-TDMA TDD mode: TDD	FDD/TDD	TDD	FDD	FDD
Chip rate (Mc/s)	FDD: 4.096/ 8.192/16.384 TDD: 4.096	1.2288xN Mcps (NX)	FDD: 4.096/8.192/ 16.384 TDD: 4.096	4.096/8.192/ 16.384	1.024/4 096/ 8.192/16 384	1.1136	1.024/4.096/ 8.192/16.384	0.9216/3.6864/ 14.7456
Frame length	10 ms	20/5 ms	10 ms	10 ms	10 ms	5 ms	10 ms	10 ms
Channel coding	Convolutional coding (rate 1/2, 1/3, $K = 9$); optional outer RS coding (rate TBD)	Convolutional coding ($R = 1/2$, 1/3, 1/4, $K = 9$); Turbo code of $R = 1/2$, 1/3, 1/4 and $K = 4$ (preferred for date transmission over 14.4 kb/s on supplemental channel)	Convolutional coding (rate 1/2, 1/3, $K = 9$); optional outer RS coding ($R = 4/5$)	Convolutional coding (FL: $R = 1/2$, $K = 7$, RL: $R = 1/3$, $K = 9$)	Convolutional coding ($R = 1/2$, 1/3, $K = 9$); Turbo code of $R = 1/3$ $K = 3$ (data transmission over 32 kb/s)	Convolutional coding ($R = 3/4$, $K = 9$); optional outer RS code; Turbo code of $K = 4$, $R = 1/2$ (preferred for data rate greater than 19.2 kb/s NRT service)	Convolutional coding ($R = 1/2$, 1/3, 1/4, 1/6, $K = 9$), selectable FEC for low rate data; Turbo code of $R = 1/3$ and $K = 3$ for high rate data and packet data	Convolutional coding $R = 1/2$, 1/3, 1/4, 1/6); optional outer (47, 41) RS code
Interleaving	Inter/intraframe	Intraframe	Inter/intraframe	Block interleaving (no details given)	Multistage intra or inter-frame	Interframe,	Intraframe	Intraframe
Data modulation	FDD: FL: QPSK, RL: Dual-channel-QPSK; TDD: QPSK (RL&FL)	QPSK (FL) BPSK (RL)	FDD: FL: QPSK, RL: Dual-channel QPSK; TDD: QPSK (RL&FL)	QPSK	FDD: FL: QPSK, RL: Dual-channel QPSK TDD: QPSK	DQPSK, and 16QAM for high data rate	QPSK (FL) BPSK (RL)	FL: QPSK RL: BPSK