



Faculty of Engineering and Technology
Department of Electrical and Computer Engineering
 Modern Communication Systems ENEE 3306
 Instructor: Dr. Wael Hashlamoun
 Midterm Exam
 Second Semester 2018-2019

Date: Sunday April 7, 2019
 Name: _____

Time: 75 minutes
 Student #: _____

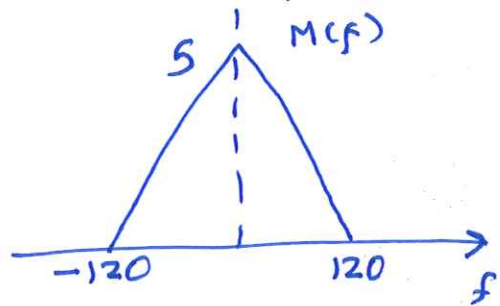
Opening Remarks:

- Calculators are allowed, however, mobile phones, books, notes, formula sheets, and other aids are not allowed.
- You are required to show all your work and provide the necessary explanations everywhere to get full credit.

Problem 1: 25 Points

Consider the signal $m(t)$, with spectrum $M(f)$, given by,

$$M(f) = \begin{cases} 5 - f/120, & 0 < f \leq 120 \\ 5 + f/120, & -120 \leq f \leq 0 \\ 0, & \text{otherwise} \end{cases}$$



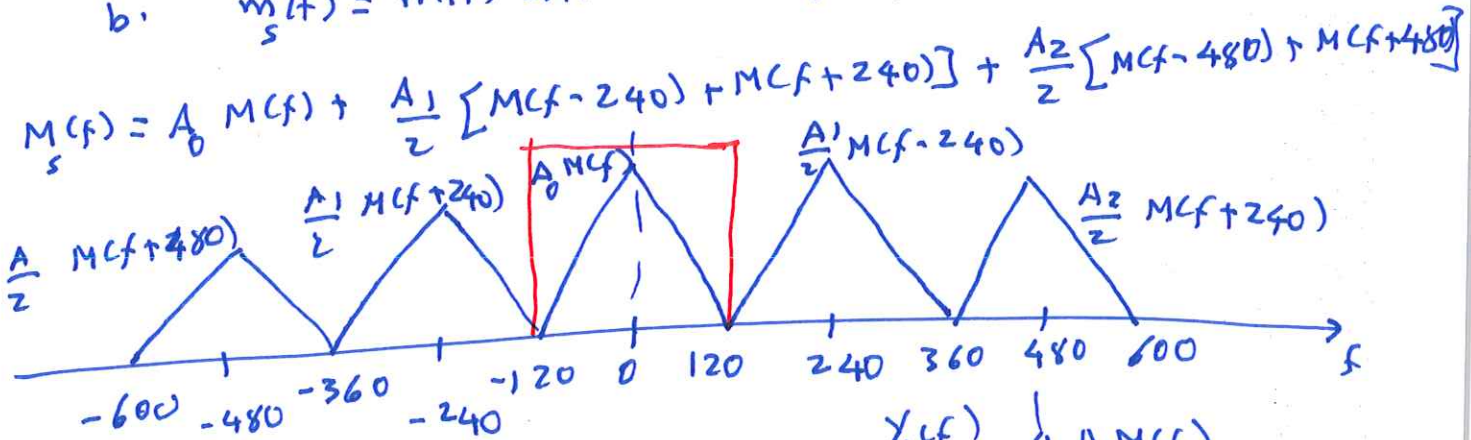
This signal is multiplied by $c(t)$ to get signal $m_s(t)$, where

$$c(t) = A_0 + A_1 \cos(2\pi(240)t) + A_2 \cos(2\pi(480)t)$$

- What is the absolute bandwidth of $m(t)$?
- Sketch $M_s(f)$, the spectrum of $m_s(t)$.
- If $m_s(t)$ is passed through an ideal low pass filter with a bandwidth of 120 Hz to produce an output $y_1(t)$. Sketch $Y_1(f)$.
- Is $y_1(t)$ proportional to $m(t)$? Explain why.

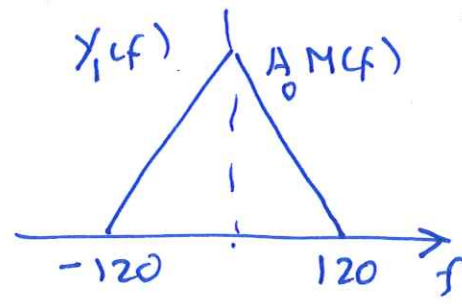
a. B.W = $(120 - 0) = 120$ Hz

b. $m_s(t) = m(t) c(t) = m(t) [A_0 + A_1 \cos 2\pi(240)t + A_2 \cos 2\pi(480)t]$



$Y_1(f) = M_s(f) H(f) = A_0 M(f)$

$Y_1(f) \propto M(f)$
 $\Rightarrow y_1(t) \propto m(t)$
 \Rightarrow No distortion

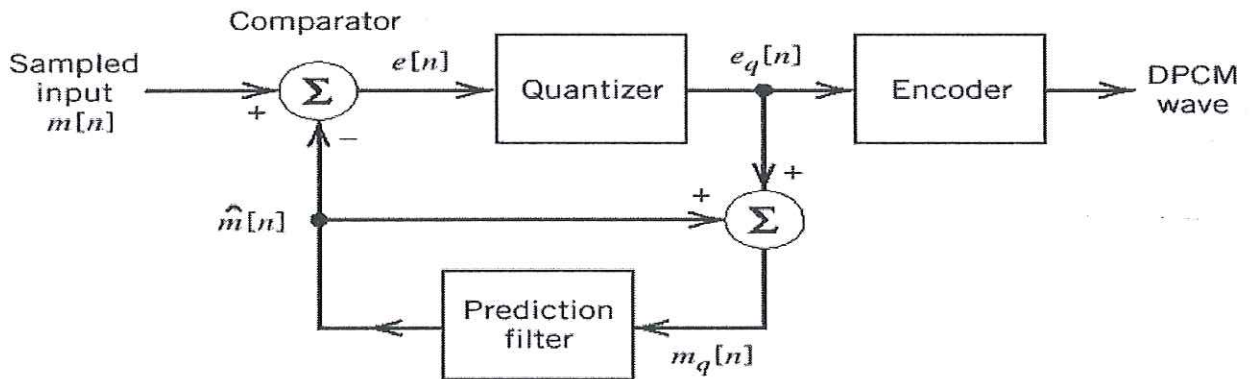


Problem 2: 25 Points

Consider a differential pulse code modulator similar to the one shown in the figure below. A signal $m(t)$ with bandwidth 400 Hz is sampled at its Nyquist rate. The error $e(t)$ is applied to a 16-level uniform quantizer. The prediction is made based only on the previous sample, i.e., $\hat{m}(nT_s) = w_1 m((n-1)T_s)$. The autocorrelation function of $m(t)$ is given by

$$R_m(\tau) = \begin{cases} 3 + \left(\frac{\tau}{3T_s}\right), & -3T_s \leq \tau \leq 0 \\ 3 - \left(\frac{\tau}{3T_s}\right), & 0 \leq \tau \leq 3T_s \\ 0, & \text{otherwise} \end{cases}$$

- Find w_1 that minimizes the mean square error between the sample and its predicted value.
- Find the sampling frequency.
- Find the data rate in bits/sec.



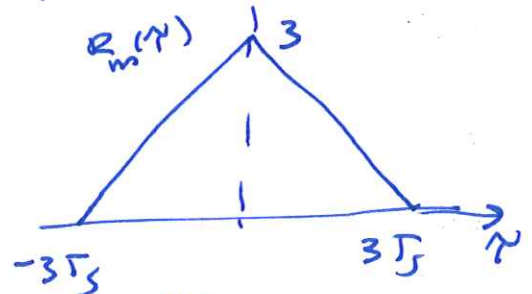
$$\hat{x}(n) = w_1 x(n-1) \Rightarrow E = E \{ [x(n) - \hat{x}(n)]^2 \}$$

11 a. $R_w = R_x$ or $R_x(0) w_1 = R_x(T_s)$

$$\Rightarrow w_1 = \frac{R_x(T_s)}{R_x(0)}$$

$$w_1 = \frac{3 - \left(\frac{T_s}{3T_s}\right)}{3}$$

$$= \frac{3 - 1/3}{3} = \frac{8/3}{3} = \frac{8}{9} = 8/9$$



7 b. $f_s = 2W = 2(400) = 800$ samples/sec

7 c. $L = 16 \Rightarrow L = 16 = 2^r = 4$ digits = r

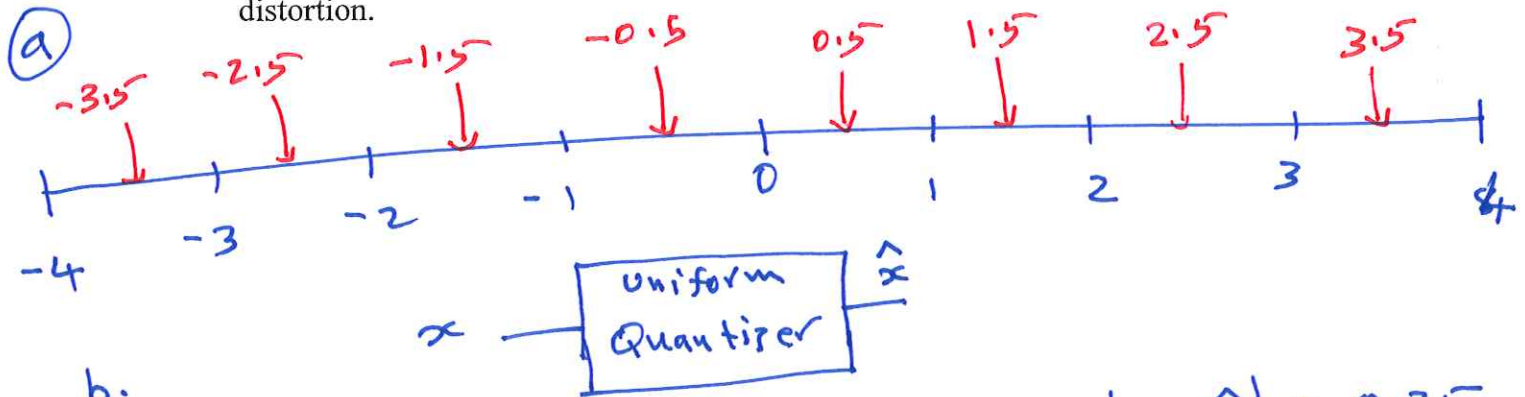
$$R_b = (2W) \text{ samples/sec} \times 4 \text{ digits/sample}$$

$$= 8W \text{ bits/sec}$$

$$= 3200 \text{ bits/sec}$$

Problem 3: 25 Points

- 8 a. Design an 8-level uniform quantizer with a dynamic range $(-4, 4)$ V, i.e., find the thresholds and representation values.
- 6 b. If a sample with a 0.15 V value is applied to the uniform quantizer of Part a, find the received signal value corresponding to this sample, and the amount of distortion affecting this sample.
- 11 c. If a sample with a 0.15 V value is applied to a μ -law companding system with $\mu = 255$ (a compressor followed by the uniform quantizer of Part a and then an expander), find the received signal value corresponding to this sample, and the amount of distortion.



b.

$$x = 0.15 \Rightarrow \hat{x} = 0.5 \Rightarrow \epsilon = |x - \hat{x}| = 0.35$$

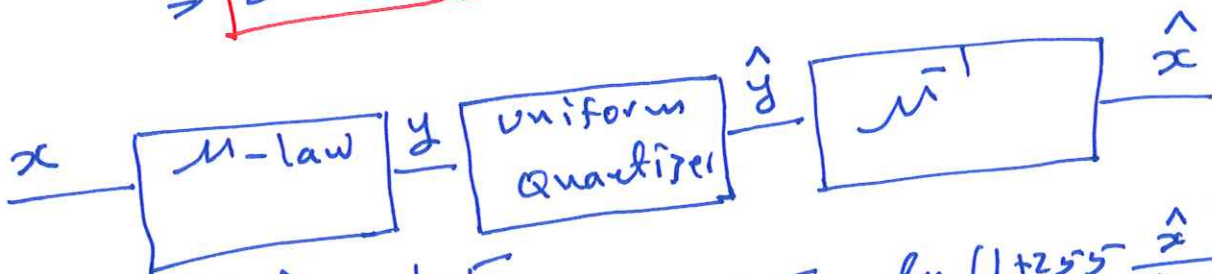
c.

$$\frac{y}{y_{\max}} = \frac{\ln(1 + 255 \frac{x}{x_{\max}})}{\ln(1 + 255)}$$

; Let $x = 0.15$

$$\frac{y}{4} = \frac{\ln(1 + 255 \frac{0.15}{4})}{\ln(256)} = \frac{2.357}{5.545} = 0.425$$

$$\Rightarrow y = 1.7$$



$\Rightarrow \hat{y} = 1.5$
Need to solve for \hat{x}

$$\frac{1.5}{4} = \frac{\ln(1 + 255 \frac{\hat{x}}{4})}{\ln(256)}$$

$$0.375 \times 5.545 = \ln(1 + 255 \frac{\hat{x}}{4}) \Rightarrow 1 + 63.75 \hat{x} = e^{2.079}$$

$$\hat{x} = \frac{e^{2.079} - 1}{63.75} = \frac{7.999 - 1}{63.75} = 0.1097$$

$$\Rightarrow \text{error} = |0.15 - 0.1097| = 0.04$$

Problem 4: 25 Points

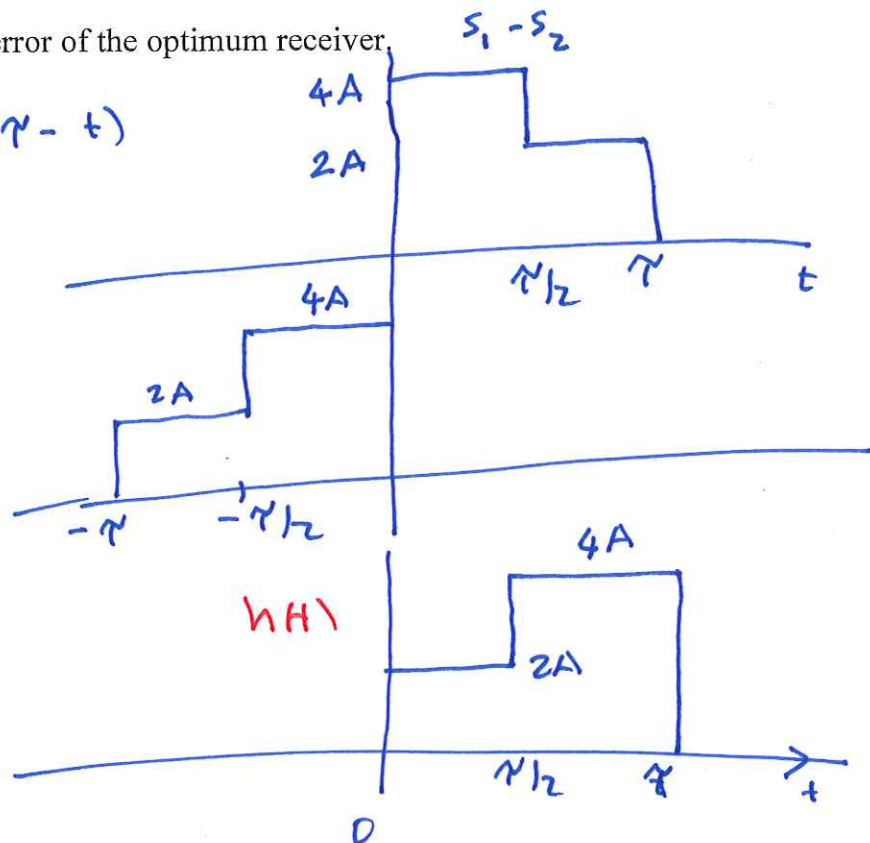
A binary digital signaling scheme employs the signal $s_1(t)$ to represent digit 1 and $s_2(t) = -s_1(t)$ to represent binary digit 0, over a channel corrupted by AWGN with power spectral density $N_0/2$ W/Hz, where

$$s_1(t) = \begin{cases} 2A, & 0 \leq t \leq \tau/2 \\ A, & \tau/2 \leq t \leq \tau \\ 0, & \text{otherwise} \end{cases}$$

- Find and sketch the optimum filter
- Find E_1 , the energy in $s_1(t)$.
- Find the average probability of error of the optimum receiver.

a. $h(t) = s_1(\tau - t) - s_2(\tau - t)$

9



b. $E_1 = \int_0^\tau |s_1(t)|^2 dt$

7

$$= \int_0^{\tau/2} 4A^2 dt + \int_{\tau/2}^\tau A^2 dt$$

$$= 4A^2 \cdot \frac{\tau}{2} + A^2 \cdot \frac{\tau}{2}$$

$$E_1 = \frac{5A^2\tau}{2}$$

c. $P_b = Q\left(\sqrt{\frac{\int_0^\tau (s_1 - s_2)^2 dt}{2N_0}}\right)$

9

$$\int_0^\tau (s_1 - s_2)^2 dt = \int_0^{\tau/2} 16A^2 dt + \int_{\tau/2}^\tau 4A^2 dt = 16A^2 \frac{\tau}{2} + 4A^2 \frac{\tau}{2}$$

$$= 10A^2\tau$$

$$P_b = Q\left(\sqrt{\frac{10A^2\tau}{2N_0}}\right) = Q\left(\sqrt{\frac{5A^2\tau}{N_0}}\right)$$

TABLE 4-2
Fourier Transform Pairs

Pair Number	$x(t)$	$X(f)$	Comments on Derivation
1.	$\Pi\left(\frac{t}{\tau}\right)$	$\tau \operatorname{sinc} \tau f$	Direct evaluation
2.	$2W \operatorname{sinc} 2Wt$	$\Pi\left(\frac{f}{2W}\right)$	Duality with pair 1, Example 4-7
3.	$\Lambda\left(\frac{t}{\tau}\right)$	$\tau \operatorname{sinc}^2 \tau f$	Convolution using pair 1
4.	$\exp(-\alpha t)u(t), \alpha > 0$	$\frac{1}{\alpha + j2\pi f}$	Direct evaluation
5.	$t \exp(-\alpha t)u(t), \alpha > 0$	$\frac{1}{(\alpha + j2\pi f)^2}$	Differentiation of pair 4 with respect to α
6.	$\exp(-\alpha t), \alpha > 0$	$\frac{2\alpha}{\alpha^2 + (2\pi f)^2}$	Direct evaluation
7.	$e^{-\pi t/\tau^2}$	$\tau e^{-\pi f/\tau^2}$	Direct evaluation
8.	$\delta(t)$	1	Example 4-9
9.	1	$\delta(f)$	Duality with pair 7
10.	$\delta(t - t_0)$	$\exp(-j2\pi f t_0)$	Shift and pair 7
11.	$\exp(j2\pi f_0 t)$	$\delta(f - f_0)$	Duality with pair 9
12.	$\cos 2\pi f_0 t$	$\frac{1}{2}\delta(f - f_0) + \frac{1}{2}\delta(f + f_0)$	Exponential representation of cos and sin and pair 10
13.	$\sin 2\pi f_0 t$	$\frac{1}{2j}\delta(f - f_0) - \frac{1}{2j}\delta(f + f_0)$	
14.	$u(t)$	$(j2\pi f)^{-1} + \frac{1}{2}\delta(f)$	Integration and pair 7
15.	$\operatorname{sgn} t$	$(j\pi f)^{-1}$	Pair 8 and pair 13 with superposition
16.	$\frac{1}{\pi t}$	$-j \operatorname{sgn}(f)$	Duality with pair 14
17.	$\hat{x}(t) = \frac{1}{\pi} \int_{-\infty}^{\infty} \frac{x(\lambda)}{t - \lambda} d\lambda$	$-j \operatorname{sgn}(f)X(f)$	Convolution and pair 15
18.	$\sum_{m=-\infty}^{\infty} \delta(t - mT_s)$	$f_s \sum_{m=-\infty}^{\infty} \delta(f - mf_s),$ $f_s = T_s^{-1}$	Example 4-10

TABLE A6.4 Trigonometric Identities

$$\begin{aligned} \exp(\pm j\theta) &= \cos \theta \pm j \sin \theta \\ \cos \theta &= \frac{1}{2}[\exp(j\theta) + \exp(-j\theta)] \\ \sin \theta &= \frac{1}{2j}[\exp(j\theta) - \exp(-j\theta)] \\ \sin^2 \theta + \cos^2 \theta &= 1 \\ \cos^2 \theta - \sin^2 \theta &= \cos(2\theta) \\ \cos^2 \theta &= \frac{1}{2}[1 + \cos(2\theta)] \\ \sin^2 \theta &= \frac{1}{2}[1 - \cos(2\theta)] \\ 2 \sin \theta \cos \theta &= \sin(2\theta) \\ \sin(\alpha \pm \beta) &= \sin \alpha \cos \beta \pm \cos \alpha \sin \beta \\ \cos(\alpha \pm \beta) &= \cos \alpha \cos \beta \mp \sin \alpha \sin \beta \\ \tan(\alpha \pm \beta) &= \frac{\tan \alpha \pm \tan \beta}{1 \mp \tan \alpha \tan \beta} \\ \sin \alpha \sin \beta &= \frac{1}{2}[\cos(\alpha - \beta) - \cos(\alpha + \beta)] \\ \cos \alpha \cos \beta &= \frac{1}{2}[\cos(\alpha - \beta) + \cos(\alpha + \beta)] \\ \sin \alpha \cos \beta &= \frac{1}{2}[\sin(\alpha - \beta) + \sin(\alpha + \beta)] \end{aligned}$$

Table of Standard Integrals

- | | |
|---|--|
| 1. $\int x^n dx = \frac{x^{n+1}}{n+1} + C \quad (n \neq -1)$ | 9. $\int \sec^2 x dx = \tan x + C$ |
| 2. $\int \frac{dx}{x} = \ln x + C$ | 10. $\int \operatorname{cosec}^2 x dx = -\cot x + C$ |
| 3. $\int e^x dx = e^x + C$ | 11. $\int \sec x dx = \ln \sec x + \tan x + C$ |
| 4. $\int \sin x dx = -\cos x + C$ | 12. $\int \operatorname{cosec} x dx = \ln \operatorname{cosec} x - \cot x + C$ |
| 5. $\int \cos x dx = \sin x + C$ | 13. $\int \sinh x dx = \cosh x + C$ |
| 6. $\int \tan x dx = -\ln \cos x + C$ | 14. $\int \cosh x dx = \sinh x + C$ |
| 7. $\int \cot x dx = \ln \sin x + C$ | 15. $\int \tanh x dx = \ln \cosh x + C$ |
| 8. $\int \frac{dx}{a^2 + x^2} = \frac{1}{a} \tan^{-1}\left(\frac{x}{a}\right) + C$ | 16. $\int \frac{dx}{\sqrt{a^2 - x^2}} = \sin^{-1}\left(\frac{x}{a}\right) + C \quad (x < a)$ |
| 17. $\int \frac{dx}{\sqrt{x^2 + a^2}} = \sinh^{-1}\left(\frac{x}{a}\right) + C = \ln\left(x + \sqrt{x^2 + a^2}\right) + C'$ | |
| 18. $\int \frac{dx}{\sqrt{x^2 - a^2}} = \cosh^{-1}\left(\frac{x}{a}\right) + C = \ln\left(x + \sqrt{x^2 - a^2}\right) + C' \quad (x > a)$ | |

Q-Function Table

z	Q(z)	z	Q(z)
0.0	0.50000	2.0	0.02275
0.1	0.46017	2.1	0.01786
0.2	0.42074	2.2	0.01390
0.3	0.38209	2.3	0.01072
0.4	0.34458	2.4	0.00820
0.5	0.30854	2.5	0.00621
0.6	0.27425	2.6	0.00466
0.7	0.24196	2.7	0.00347
0.8	0.21186	2.8	0.00256
0.9	0.18406	2.9	0.00187
1.0	0.15866	3.0	0.00135
1.1	0.13567	3.1	0.00097
1.2	0.11507	3.2	0.00069
1.3	0.09680	3.3	0.00048
1.4	0.08076	3.4	0.00034
1.5	0.06681	3.5	0.00023
1.6	0.05480	3.6	0.00016
1.7	0.04457	3.7	0.00011
1.8	0.03593	3.8	0.00007
1.9	0.02872	3.9	0.00005